

Application note

LSM6DSV32X: 6-axis IMU with 32 *g* accelerometer and embedded sensor fusion, AI, Qvar for high-end applications

Introduction

This document provides usage information and application hints related to ST's [LSM6DSV32X](https://www.st.com/en/product/lsm6dsv32x?ecmp=tt9470_gl_link_feb2019&rt=an&id=AN6016) iNEMO 6-axis IMU (inertial measurement unit).

The LSM6DSV32X is a 3-axis digital accelerometer and 3-axis digital gyroscope system-in-package with a digital I²C, SPI, and MIPI I3C® serial interface standard output, performing at 0.65 mA in combination high-performance mode. Thanks to the ultralow noise performance of both the gyroscope and the accelerometer, the device combines always-on low-power features with superior sensing precision for an optimal motion experience for the consumer. Furthermore, the accelerometer features smart sleep-to-wake-up (activity) and return-to-sleep (inactivity) functions that allow advanced power saving.

The device has a dynamic user-selectable full-scale acceleration range of ±4/±8/±16/±32 *g* and an angular rate range of ±125/±250/±500/±1000/±2000/±4000 dps. It features the capability to enable up to three different cores for UI, EIS, and OIS data processing.

The device can be configured to generate interrupt signals by using hardware recognition of free-fall events, 6D orientation, tap and double-tap sensing, activity or inactivity, and wake-up events.

The availability of different connection modes to external sensors allows implementing additional functionalities such as a sensor hub, auxiliary SPI, and analog hub.

The LSM6DSV32X is compatible with the requirements of the leading OSs, offering real, virtual, and batch-mode sensors. It has been designed to implement in hardware significant motion, relative tilt, pedometer functions, timestamp, and provides an incredible level of customization: up to eight embedded finite state machines can be programmed independently for motion detection or gesture recognition such as glance, absolute wrist tilt, shake, double-shake, or pick-up.

The device also embeds machine learning core logic, which allows identifying if a data pattern matches a user-defined set of classes. A typical example of an application could be activity detection like running, walking, driving, and so on.

The LSM6DSV32X embeds an analog hub sensing functionality, which is able to connect an analog input and convert it to a digital signal for embedded processing. In addition, an embedded Qvar (electric charge variation detection) channel can be used for human presence and motion detection, touch detection, and user interface (UI) applications.

The device has an integrated smart first-in first-out (FIFO) buffer of up to 4.5 KB size, allowing dynamic batching of significant data (that is, external sensors, step counter, timestamp and temperature, MLC exported filters and features).

The LSM6DSV32X is available in a small plastic, land grid array package (LGA-14L) and it is guaranteed to operate over an extended temperature range from -40°C to +85°C.

The ultrasmall size and weight of the SMD package make it an ideal choice for handheld portable applications such as smartphones, IoT connected devices, and wearables or any other application where reduced package size and weight are required.

1 Pin description

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Figure 1. Pin connections

Table 1. Internal pin status

Pin description

1. Mode 3 is enabled when the OIS_XL_EN bit or the OIS_G_EN bit in the UI_CTRL1_OIS (70h) / SPI2_CTRL1_OIS (70h) registers is set to 1.

2. The analog hub and Qvar functions are enabled by setting the AH_QVAR_EN bit to 1 in CTRL7 (16h).

Internal pull-up value is from 30 kΩ to 50 kΩ, depending on Vdd_IO.

All the registers given in the following table are accessible from the primary SPI/I²C/MIPI I3C[®] interface only.

Table 2. Registers

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All the registers given in the following table are accessible from the auxiliary SPI interface only.

2.1 Embedded functions registers

The table given below provides a list of the registers for the embedded functions available in the device and the corresponding addresses. Embedded functions registers are accessible when the EMB_FUNC_REG_ACCESS bit is set to 1 in the FUNC_CFG_ACCESS register.

Table 4. Embedded functions registers

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2.2 Embedded advanced features pages

The table given below provides a list of the registers for the embedded advanced features page 0. These registers are accessible when PAGE_SEL[3:0] are set to 0000 in the PAGE_SEL register.

Table 5. Embedded advanced features registers - page 0

Table 6. Embedded advanced features registers - page 1

The following table provides a list of the registers for the embedded advanced features page 2. These registers are accessible when PAGE_SEL[3:0] are set to 0010 in the PAGE_SEL register.

Table 7. Embedded advanced features registers - page 2

2.3 Sensor hub registers

Table 8. Sensor hub registers

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3 Operating modes

The LSM6DSV32X provides three possible operating configurations:

- Only accelerometer active and gyroscope in power-down mode or sleep mode
- Only gyroscope active and accelerometer in power-down mode
- Both accelerometer and gyroscope active with independent ODR and power mode

The device offers a wide Vdd voltage range from 1.71 V to 3.6 V and a Vdd IO range from 1.08 V to 3.6 V. The power-on sequence is not restricted. The Vdd/Vdd_IO pins can be set to either the power supply level or to ground level (they must not be left floating) and no specific sequence is required for powering them on. In order to avoid potential conflicts, during the power-on sequence it is recommended to set the lines (on the host side) connected to the device IO pins floating or connected to ground, until Vdd IO is set. After Vdd IO is set, the lines connected to the IO pins have to be configured according to their default status described in [Table 1.](#page-2-0) In

order to avoid an unexpected increase in current consumption, the input pins that are not pulled-up/pulled-down must be the polarized by the host. When the Vdd power supply is applied, the device performs a 10 ms (maximum) boot procedure to load the trimming parameters. After the boot is completed, both the accelerometer and the gyroscope are automatically

configured in power-down mode. To guarantee proper power-off of the device, it is recommended to maintain the duration of the Vdd line to GND for at least 100 μs.

The accelerometer and the gyroscope can be configured independently. When both the accelerometer and gyroscope are on, the accelerometer is synchronized with the gyroscope, and the data rates of the two sensors are integer multiples of each other.

The accelerometer can be configured in the one of the following power modes:

- Power-down mode
- Low-power mode (three different modes are available, depending on the number of averaged measurements)
- Normal mode
- High-performance mode
- High-accuracy ODR mode
- ODR-triggered mode

The gyroscope can be configured in one of the following power modes:

- Power-down mode
- Sleep mode
- Low-power mode
- High-performance mode
- High-accuracy ODR mode
- ODR-triggered mode

3.1 Accelerometer power modes and output data rates

The power mode and the output data rate of the accelerometer can be selected using the CTRL1 register.

When the accelerometer is configured in **power-down** mode, almost all internal blocks of the device are switched off to minimize power consumption. Digital interfaces (I²C, MIPI I3C[®], and SPI) are still active to allow communication with the device. The content of the configuration registers is preserved and the output data registers are not updated, keeping the last data sampled in memory before going into power-down mode.

When the accelerometer is configured in **low-power** mode, its reading chain is automatically turned on and off to optimize the current consumption. Three different low-power modes are available, based on the number of measurements that are averaged for the sample generation:

- Low-power mode 1 (LPM1), where two measurements are averaged
- Low-power mode 2 (LPM2), where four measurements are averaged
- Low-power mode 3 (LPM3), where eight measurements are averaged

Increasing the number of averaged measurements allows reducing the noise, while decreasing them allows reducing the current consumption.

In the low-power modes, the antialiasing filter is disabled and accelerometer ODR is selectable up to 240 Hz.

When the accelerometer is configured in **normal** mode, its reading chain is always on. The antialiasing filter is enabled and the accelerometer ODR is selectable up to 1920 Hz. Normal mode provides a balanced trade-off between noise and current consumption.

When the accelerometer is configured in **high-performance** mode, its reading chain is always on. The antialiasing filter is enabled and the accelerometer ODR is selectable up to 7680 Hz. High-performance mode provides the best performance in terms of noise.

When the accelerometer is configured in **high-accuracy ODR** (HAODR) mode, its reading chain is always on. The antialiasing filter is enabled and three sets of accelerometer ODRs are available, based on the value of the HAODR SEL [1:0] bits in the HAODR CFG register, as shown in [Table 11.](#page-17-0) High-accuracy ODR mode provides the best performance in terms of noise (same as high-performance mode) and typically reduces the part-to-part ODR variation.

If high-accuracy ODR mode is intended to be used, the following limitations must be considered:

- When HAODR mode is intended to be used for one sensor (accelerometer or gyroscope), the other sensor has to be configured in HAODR mode too.
- HAODR mode is not applied to the accelerometer and gyroscope OIS channels.
- Both the accelerometer and the gyroscope must be set in power-down mode (ODR_XL = 0000 in the CTRL1 register and ODR_G = 0000 in the CTRL2 register) before enabling or disabling HAODR mode.
- If the accelerometer is on (ODR_XL > 0) in HAODR mode, it is mandatory to wait at least 500 μs between two ODR XL value changes.
- HAODR mode is not compatible with the Qvar functionality.
- HAODR mode is not compatible with the pedometer, relative tilt, or activity/inactivity functionality (only motion/stationary can be used).

When the accelerometer is configured in **ODR-triggered** mode, its reading chain is always on. The antialiasing filter is enabled and the accelerometer ODR can be fine-tuned by the user by means of an external reference signal. ODR-triggered mode provides the best performance in terms of noise (same as high-performance mode) and the additional capability to synchronize the accelerometer ODR with the external reference signal. ODRtriggered mode is described in [Section 3.3: ODR-triggered mode.](#page-20-0)

Table 9 summarizes the available power modes based on the OP_MODE_XL bits of the CTRL1 register. The power-down mode is selected if ODR_XL = 0000, regardless of the configuration of the OP_MODE_XL bits.

Table 9. Accelerometer power modes

Table 10 summarizes the available ODR values based on the ODR_XL bits of the CTRL1 register.

Table 10. Accelerometer ODR

Table 11. Accelerometer ODR selection in high-accuracy ODR mode

3.2 Gyroscope power modes and output data rates

mode is drastically reduced.

The power mode and the output data rate of the gyroscope can be selected using the CTRL2 register. When the gyroscope is configured in **power-down** mode, almost all internal blocks of the device are switched off to minimize power consumption. The digital interfaces (¹²C, MIPI 13C[®], and SPI) are still active to allow communication with the device. The content of the configuration registers is preserved and the output data registers are not updated, keeping the last data sampled in memory before going into power-down mode. When the gyroscope is in **sleep** mode, the circuitry that drives the oscillation of the gyroscope mass is active, but the reading chain is turned off. Compared to power-down mode, the turn-on time from sleep mode to any active

When the gyroscope is configured in **low-power** mode, the driving circuitry is always on, but the reading chain is automatically turned on and off to optimize the current consumption. The gyroscope ODR is selectable up to 240 Hz.

When the gyroscope is configured in **high-performance** mode, its reading chain is always on. The gyroscope ODR is selectable up to 7680 Hz. High-performance mode provides the best performance in terms of noise.

When the gyroscope is configured in **high-accuracy ODR** (HAODR) mode, its reading chain is always on. Three sets of gyroscope ODRs are available, based on the value of the HAODR_SEL_[1:0] bits in the HAODR_CFG register, as shown in [Table 14.](#page-19-0) High-accuracy ODR mode provides the best performance in terms of noise (same as high-performance mode) and typically reduces the part-to-part ODR variation.

If high-accuracy ODR mode is intended to be used, the following limitations must be considered:

- When HAODR mode is intended to be used for one sensor (accelerometer or gyroscope), the other sensor has to be configured in HAODR mode too.
- HAODR mode is not applied to the accelerometer and gyroscope OIS channels.
- Both the accelerometer and the gyroscope must be set in power-down mode (ODR_XL = 0000 in the CTRL1 register and ODR_G = 0000 in the CTRL2 register) before enabling or disabling HAODR mode.
- HAODR mode is not compatible with the Qvar functionality.
- HAODR mode is not compatible with the activity/inactivity functionality (only motion/stationary can be used).

When the gyroscope is configured in **ODR-triggered** mode, its reading chain is always on. The gyroscope ODR can be fine-tuned by the user by means of an external reference signal. ODR-triggered mode provides the best performance in terms of noise (same as high-performance mode) and the additional capability to synchronize the gyroscope ODR with the external reference signal. ODR-triggered mode is described in [Section 3.3: ODR](#page-20-0)[triggered mode.](#page-20-0)

Table 12 summarizes the available power modes based on the OP_MODE_G bits of the CTRL2 register. The power-down mode is selected if ODR_G = 0000, regardless of the configuration of the OP_MODE_G bits.

Table 12. Gyroscope power modes

Table 13 summarizes the available ODR based on the ODR_G bits of the CTRL2 register.

Table 13. Gyroscope ODR

Table 14. Gyroscope ODR selection in high-accuracy ODR mode

3.3 ODR-triggered mode

Every device has its own ODR frequency (due to natural spreads). The LSM6DSV32X device provides a way to synchronize its data generation with an external hardware reference signal provided over the INT2 pin. The device is able to automatically align the frequency and phase to the edges of the reference signal.

ODR-triggered mode supports accelerometer only, gyroscope only, and combo (accelerometer and gyroscope) modes. If either the accelerometer or the gyroscope is intended to be used in ODR-triggered mode, they must be both configured in ODR-triggered mode or power-down mode. When both the accelerometer and gyroscope are enabled, the user must configure the same ODR on both the accelerometer and gyroscope. It is not possible to select different ODRs for the accelerometer and gyroscope; if different ODR values are set, the ODR configured for the gyroscope data is also applied to the accelerometer data.

The full-scale configurations are totally independent between the accelerometer and gyroscope and they can be set in any combination.

Note: ODR-triggered mode has to be enabled / disabled when the device is in power-down mode.

Note: When ODR-triggered mode is enabled, the 0001, 0010, and 1100 configurations of the ODR_XL_[3:0] bits in register CTRL1 and the 0010 and 1100 configurations of the ODR_G_[3:0] bits in register CTRL2 cannot be used.

Note: ODR-triggered mode is not compatible with the analog hub / Qvar functionality or the EIS functionality.

Note: ODR-triggered mode is not compatible with the pedometer, relative tilt, SFLP, DRDY mask, or activity/inactivity functionality (only motion/stationary can be used).

> When using the ODR-triggered mode, the lock is reached (and the filtering chains are settled) after 3 clock periods of the external reference signal.

When using additional digital filters (for example, the accelerometer LPF2 or HP filter, or gyroscope LPF1 filter), their settling time must be considered in addition to the filtering chain settling time indicated above (see [Section 3.6: Accelerometer bandwidth](#page-23-0) and [Section 3.9: Gyroscope bandwidth](#page-28-0) for more details about accelerometer and gyroscope bandwidth and turn-on/off time).

The selectable output date rates (ODR_{sel}) are different with respect to the regular ODRs set. The external reference signal must be provided with a period, which is an even multiple of the desired ODR period. The desired ODR period can have a maximum deviation of ±33% with respect to the selected ODR_{sel}. The desired number of samples in one external reference signal period must be configured through the

ODR_TRIG_N_ODR_[7:0] bits of the ODR_TRIG_CFG register. The value of the ODR_TRIG_N_ODR_[7:0] bits can span from a minimum of 4 to a maximum of 255 and its resolution is 2 samples (therefore, from a minimum of 8 samples to a maximum of 510 samples). The ODR_{sel} set available for ODR-triggered mode, the corresponding minimum number of samples for each ODR, and the corresponding minimum period for the external reference signal are indicated in Table 15.

Table 15. ODR-triggered mode configurability

Figure 2 shows an example of the external reference signal with respect to the internal data-ready signal when configuring ODR_TRIG_N_ODR_[7:0] to 4 (8 samples).

The external reference signal pulse duration must be at least 5 μ s and its polarity can be selected through the INT2_IN_LH bit of the CTRL4 register. If the INT2_IN_LH bit is set to 0, the ODR-triggered mode is sensitive to the falling edge of the external reference signal, otherwise it is sensitive to the rising edge.

The following procedure must be used for configuring the device in ODR-triggered mode.

- 1. Set both the sensors in power-down mode.
- 2. Configure the number of samples for each T_{ref} period by setting the ODR_TRIG_N_ODR_[7:0] bits.
- 3. Set both sensors in ODR-triggered mode.
- 4. Configure the reference signal polarity.
- 5. Start the external reference signal on INT2.
- 6. Set the accelerometer and/or gyroscope ODR_{sel}.

The following example configures the sensors in ODR-triggered combo mode at 100 Hz (10 ms time period).

100 Hz has a 0% deviation with respect to the selected ODR configured through the ODR_XL_[3:0] bits of the CTRL1 register and ODR_G_[3:0] bits of the CTRL2 register. The user could, using the same configuration listed above, achieve an ODR of 100 Hz \pm 33% by fine-tuning the external reference signal period (T_{ref}).

3.4 Current consumption

Table 16 shows the typical values of power consumption for the different operating modes.

3.5 Connection modes

The device offers three different connection modes, described in detail in this document:

- Mode 1: it is the connection mode enabled by default. The I²C slave interface, MIPI I3C[®] slave interface, or SPI (3- / 4-wire) serial interface is available.
- **Mode 2**: it is the sensor hub mode. The I²C slave interface, MIPI I3C® slave interface, or SPI (3- / 4-wire) serial interface and I²C interface master for external sensor connections are available. This connection mode is described in [Section 7: Mode 2 - sensor hub mode.](#page-71-0)
- **Mode 3**: in addition to the primary I²C slave interface, MIPI I3C® slave interface or SPI (3- / 4-wire) serial interface, an auxiliary SPI (3- / 4-wire) serial interface for external device connections (for example, camera module) is available. This connection mode is described in [Section 8: Mode 3 - OIS functionality.](#page-82-0)

3.6 Accelerometer bandwidth

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The accelerometer sampling chain is represented by a cascade of four main blocks: an analog antialiasing lowpass filter, an ADC converter, a digital low-pass filter (LPF1), and the composite group of digital filters.

Figure 3. Accelerometer filtering chain (UI path) shows the accelerometer sampling chain on the UI path. The accelerometer sampling chain active on the OIS path (when using mode 3 configuration) is described in [Section 8: Mode 3 - OIS functionality](#page-82-0).

The analog signal coming from the mechanical parts is filtered by an analog antialiasing low-pass filter before being converted by the ADC. The antialiasing filter is not enabled in the low-power modes. The digital LPF1 filter provides different cutoff values based on the accelerometer mode selected:

- ODR / 2 when the accelerometer is configured in normal, high-performance, or high-accuracy ODR mode
- 2300 Hz when the accelerometer is configured in low-power mode 1
- 912 Hz when the accelerometer is configured in low-power mode 2
- 431 Hz when the accelerometer is configured in low-power mode 3

Figure 3. Accelerometer filtering chain (UI path)

The "Advanced functions" block in the figure above refers to the pedometer, step detector and step counter, significant motion and tilt functions, described in [Section 6: Embedded functions,](#page-61-0) and also includes the finite state machine and the machine learning core.

Finally, the composite group of filters composed of a low-pass digital filter (LPF2), a high-pass digital filter, and a slope filter processes the digital signal.

The HP_LPF2_XL_BW_[2:0] bits in the CTRL8 register and the CTRL9 register can be used to configure the composite filter group and the overall bandwidth of the accelerometer filtering chain, as shown in [Table 17. Accelerometer bandwidth selection in mode 1/2/3.](#page-24-0) Referring to this table, on the low-pass path side, the Bandwidth column refers to the LPF1 bandwidth if LPF2_XL_EN = 0; it refers to the LPF2 bandwidth if LPF2_XL_EN = 1. On the high-pass path side, the Bandwidth column refers to the slope filter bandwidth if HP_LPF2_XL_BW_ $[2:0] = 000$; it refers to the HP filter bandwidth for all the other configurations.

Table 17. Accelerometer bandwidth selection in mode 1/2/3 also provides the maximum (worst case) settling time in terms of samples to be discarded for the various configurations of the accelerometer filtering chain.

1. Settling time @ 99% of the final value, taking into account all output data rates and all operating mode switches

2. This value is ODR / 2 when the accelerometer is in high-performance mode, high-accuracy ODR mode and normal mode. It is equal to 2300 Hz when the accelerometer is in low-power mode 1 (2 mean), 912 Hz in low-power mode 2 (4 mean) and *431 Hz in low-power mode 3 (8 mean).*

Setting the HP_SLOPE_XL_EN bit to 0, the low-pass path of the composite filter block is selected. If the LPF2_XL_EN bit is set to 0, no additional filter is applied. If the LPF2_XL_EN bit is set to 1, the LPF2 filter is applied in addition to LPF1, and the overall bandwidth of the accelerometer chain can be set by configuring the HP_LPF2_XL_BW_[2:0] field of the CTRL8 register.

The LPF2 low-pass filter can also be used in the 6D/4D functionality by setting the LOW_PASS_ON_6D bit of the TAP CFG0 register to 1.

Setting the HP_SLOPE_XL_EN bit to 1, the high-pass path of the composite filter block is selected. The HP_LPF2_XL_BW_[2:0] field is used in order to enable, in addition to the LPF1 filter, either the slope filter usage (when HP_LPF2_XL_BW_[2:0] = 000) or the digital high-pass filter (other HP_LPF2_XL_BW_[2:0] configurations). The HP_LPF2_XL_BW_[2:0] field is also used to select the cutoff frequencies of the HP filter.

The high-pass filter reference mode feature is available for the accelerometer sensor: when this feature is enabled, the current X, Y, Z accelerometer sample is internally stored and subtracted from all subsequent output values. In order to enable the reference mode, both the HP_REF_MODE_XL bit and the HP_SLOPE_XL_EN bit of the CTRL9 register have to be set to 1, and the value of the HP_LPF2_XL_BW_[2:0] field has to be different than 000. When the reference mode feature is enabled, both the LPF2 filter and the HP filter are not available. The first accelerometer output data after enabling the reference mode has to be discarded.

The XL_FASTSETTL_MODE bit of the CTRL9 register enables the accelerometer LPF2 or HPF fast-settling mode: the selected filter sets the first sample after writing this bit. This feature applies only upon device exit from power-down mode.

3.6.1 Accelerometer slope filter

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As shown in [Figure 3. Accelerometer filtering chain \(UI path\),](#page-23-0) the device embeds a digital slope filter, which can also be used for some embedded features such as single/double-tap recognition, wake-up detection and activity/ inactivity.

The slope filter output data is computed using the following formula:

 $slope(t_n) = [acc(t_n) - acc(t_{n-1})]/2$

An example of a slope data signal is illustrated in the following figure.

Figure 4. Accelerometer slope filter

3.7 Accelerometer turn-on/off time

The accelerometer reading chain contains low-pass filtering to improve signal-to-noise performance and to reduce aliasing effects. For this reason, it is necessary to take into account the settling time of the filters when the accelerometer power mode is switched or when the accelerometer ODR is changed.

The accelerometer chain settling time is dependent on the power mode and output data rate selected for the following configurations:

- LPF2 and HP filters disabled
- LPF2 or HP filter enabled with ODR / 4 bandwidth selection

For these two possible configurations, the maximum overall turn-on/off in order to switch accelerometer power modes or accelerometer ODR is the one shown below in [Table 18. Accelerometer turn-on/off time \(LPF2 and HP](#page-26-0) [disabled\)](#page-26-0) and [Table 19. Accelerometer samples to be discarded.](#page-26-0)

Note: Accelerometer ODR timing is not impacted by power mode changes (the new configuration is effective after the completion of the current period).

Table 18. Accelerometer turn-on/off time (LPF2 and HP disabled)

1. Settling time @ 99% of the final value

Table 19. Accelerometer samples to be discarded

Overall settling time if LPF2 or HP digital filters are enabled with bandwidth different from ODR / 4 has been already indicated in [Table 17. Accelerometer bandwidth selection in mode 1/2/3.](#page-24-0) When the device is configured in mode 3, the accelerometer UI path filtering chain is not impacted by the enabling/disabling of the accelerometer/gyroscope OIS path filtering chain.

3.8 Accelerometer dual-channel mode

The LSM6DSV32X accelerometer incorporates a dual-channel architecture capable of simultaneously providing two sets of acceleration data with two different full-scale values. By default, the device operates in single-channel mode supporting full-scale values from ±4 *g* to ±32 *g* and multiple power modes. The dual-channel functionality can be enabled / disabled by configuring the XL_DualC_EN bit to 1 (enable) or to 0 (disable) in the CTRL8 register. It is available for all the accelerometer operating modes.

Enabling / disabling dual-channel mode does not impact the accelerometer channel 1, which keeps working at the full scale set through the FS_XL_[1:0] bits in the CTRL8 register.

Figure 5. Dual-channel mode

As shown in Figure 5, when dual-channel mode is enabled, an additional path (channel 2) of the accelerometer chain generates data at ±32 *g* full scale and at the output data rate configured through the ODR_XL field of the CTRL1_XL register. Accelerometer channel 2 data are available in the output registers from UI_OUTX_L_A_OIS_DualC through UI_OUTZ_H_A_OIS_DualC (34h to 39h) and are expressed in two's complement. The bandwidth and the settling time of channel 2 are imposed by the LPF1 digital low-pass filter. See [Section 3.7: Accelerometer turn-on/off time](#page-25-0) for further details.

The accelerometer channel 2 data at ±32 *g* full scale can also be stored in FIFO by setting the XL_DualC_BATCH_FROM_IF bit to 1 in the EMB_FUNC_CFG register or by using some dedicated commands of the finite state machine (FSM) embedded in the LSM6DSV32X device. The latter functionality can be enabled by setting the XL_DualC_BATCH_FROM_FSM bit to 1 in the FIFO_CTRL2 register. See more details in [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0).

When dual-channel mode is enabled, OIS accelerometer data (mode 3) are not accessible from the primary interface (they are accessible over the auxiliary SPI interface only) unless FSM-triggered FIFO batching of accelerometer channel 2 data has been enabled.

3.9 Gyroscope bandwidth

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The gyroscope filtering chain depends on the connection mode in use.

When mode 1 or mode 2 is selected, the gyroscope filtering chain configuration is the one shown in Figure 6. It is a cascade of two filters: a selectable digital low-pass filter (LPF1) and a digital low-pass filter (LPF2).

The LPF1 filter is available in high-performance mode and in high-accuracy ODR mode. If the gyroscope is configured in low-power mode, the LPF1 filter is bypassed.

Figure 6. Gyroscope digital chain - mode 1 (UI/EIS) and mode 2

Note: When the gyroscope OIS or EIS chain is enabled, the LPF1 filter is not available in the gyroscope UI chain. It is recommended to avoid using the LPF1 filter in the gyroscope UI chain when the gyroscope OIS or EIS is intended to be used.

> The digital LPF1 filter can be enabled by setting the LPF1_G_EN bit of the CTRL7 register to 1 and its bandwidth can be selected through the field LPF1 G BW [2:0] of the CTRL6 register.

The digital LPF2 filter cannot be configured by the user and its cutoff frequency depends on the selected gyroscope ODR. When the gyroscope ODR is equal to 7680 Hz, the LPF2 filter is bypassed.

The overall gyroscope bandwidth for different gyroscope ODR values and for different configurations of the LPF1_G_EN bit of the CTRL7 register and LPF1_G_BW_[2:0] of the CTRL6 register is summarized in the following table.

Table 20. Gyroscope overall bandwidth selection in mode 1/2

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1. For ODR ≥ 3840 Hz the cases LPF1_G_BW_[2:0] = 11x should be avoided due to low LPF1 roll-off at higher frequency.

If the gyroscope is configured in low-power mode, the gyroscope filtering chain presented above is bypassed. The bandwidth in low-power mode is indicated in the following table.

Table 21. Gyroscope low-power mode bandwidth

If mode 3 is enabled, the gyroscope digital chain becomes the one shown in Figure 7. In this configuration, two different data chains are available:

- The user interface (UI) chain, where the gyroscope data are provided to the primary I²C / MIPI I3C[®] / SPI with an ODR selectable from 7.5 Hz up to 7680 Hz.
- The optical image stabilization (OIS) chain, where the gyroscope data are provided to the auxiliary SPI with an ODR fixed at 7680 Hz.

Figure 7. Gyroscope digital chain - mode 3 (OIS)

- *1. When the gyroscope OIS or EIS chain is enabled, the LPF1 filter is not available in the gyroscope UI chain.*
- *2. It is recommended to avoid using the LPF1 filter in mode1/2 when the gyroscope OIS or EIS chain is intended to be used.*

In mode 3, the LPF1 filter is dedicated to the OIS chain only. On the UI side, if the gyroscope is configured in high-performance mode, the total bandwidth depends on the gyroscope ODR value, as shown in Table 22. If the gyroscope is configured in low-power mode, the gyroscope chain bandwidth is still the one indicated in [Table 21.](#page-30-0)

A detailed description of mode 3 connection modes and the gyroscope OIS chain is provided in [Section 8: Mode 3](#page-82-0) [- OIS functionality](#page-82-0).

3.10 Gyroscope turn-on/off time

Turn-on/off time has to be considered also for the gyroscope sensor when switching its modes or when the gyroscope ODR is changed.

When the device is configured in mode 1/2, the maximum overall turn-on/off time in order to switch gyroscope power modes or gyroscope ODR is the one shown in Table 23. Gyroscope turn-on/off time in mode 1/2.

Note: The gyroscope ODR timing is not impacted by power mode changes (the new configuration is effective after the completion of the current period).

Table 23. Gyroscope turn-on/off time in mode 1/2

1. Settling time @ 99% of the final value

2. Only when LPF1 is enabled

Table 24. Gyroscope samples to be discarded in mode 1/2 (LPF1 disabled)

1. Settling time @ 99% of the final value

Table 25. Gyroscope chain settling time in mode 1/2 (LPF1 enabled)

1. Settling time @ 99% of the final value

When the device is configured in mode 3, the gyroscope UI path filtering chain is not impacted by enabling/ disabling the gyroscope OIS path filtering chain.

3.11 Gyroscope EIS channel

The LSM6DSV32X integrates an independent gyroscope channel specifically designed for EIS (electrical image stabilization) applications. It basically provides an additional dedicated gyroscope channel, accessible from the primary I²C / MIPI I3C[®] / SPI interface, with the full scale and filtering chain independently configurable.

The gyroscope EIS channel can be enabled/disabled by configuring the ODR_G_EIS_[1:0] field of the CTRL_EIS register. By default, the EIS channel is off (ODR G EIS [1:0] = 00). When these bits are set to 01 or 10, the EIS channel is enabled and the EIS gyroscope data are generated at a rate of 1920 Hz or 960 Hz, respectively. These are the only two data rates available for the EIS channel.

Note: When the EIS channel is enabled, the gyroscope UI can be configured only in power-down, high-performance, or high-accuracy ODR mode.

Figure 8. Gyroscope EIS channel

The full scale of the EIS gyroscope data can be set using the FS_G_EIS_[2:0] field of the CTRL_EIS register.

Note: If the FS_G_[3:0] bits in CTRL6 are equal to 1100 (±4000 dps), FS_G_EIS_[2:0] must be set to 100 in order to select ±4000 dps full scale on both the UI and EIS channels.

> As shown in Figure 8, a dedicated filtering chain is available for the EIS channel. The bandwidth of the LPF_EIS digital low-pass filters depends on the configuration applied to the ODR_G_EIS_[1:0] field and to the LPF_G_EIS_BW bit in the CTRL_EIS register, as described in Table 26, which indicates also the number of samples to be discarded after the 70 ms wait time if the gyroscope exits from power-down mode.

Table 26. Gyroscope LPF_EIS filter bandwidth selection

1. Settling time @ 99% of the final value

When the EIS channel is enabled, the EIS gyroscope data are available in the output data register from UI_OUTX_L_G_OIS_EIS through UI_OUTZ_H_G_OIS_EIS (2Eh to 33h) by setting the

G_EIS_ON_G_OIS_OUT_REG bit to 1 in the CTRL_EIS register. The EIS channel is compatible with mode 3 selection, but when the G_EIS_ON_G_OIS_OUT_REG bit is set to 1, the gyroscope OIS data cannot be read from the primary interface (they are accessible over the auxiliary SPI interface only).

For the EIS gyroscope channel, the data-ready signal is represented by the GDA_EIS bit of the STATUS_REG register. This signal can be driven to the INT2 pin by setting the INT2_DRDY_G_EIS bit of the INT2_CTRL register to 1.

EIS gyroscope data can also be stored in FIFO by setting the G_EIS_FIFO_EN bit to 1 in the FIFO_CTRL4 register. See [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0) for more details.

4 Mode 1 - reading output data

4.1 Startup sequence

Once the device is powered up, it automatically downloads the calibration coefficients from the embedded flash to the internal registers. When the boot procedure is completed, that is, after 10 ms (maximum), the accelerometer and gyroscope automatically enter power-down mode.

To turn on the accelerometer and gather acceleration data through the primary I²C / MIPI I3C[®] / SPI interface, it is necessary to select one of the operating modes through the CTRL1 register.

The following general-purpose sequence can be used to configure the accelerometer:

To turn on the gyroscope and gather angular rate data through the primary I²C / MIPI I3C[®] / SPI interface, it is necessary to select one of the operating modes through the CTRL2 register.

The following general-purpose sequence can be used to configure the gyroscope:

4.2 Using the status register

The device is provided with a STATUS REG register which can be polled to check when a new set of data is available. The XLDA bit is set to 1 when a new set of data is available in the accelerometer output registers. The GDA bit is set to 1 when a new set of data is available in the gyroscope output registers.

For the accelerometer (the gyroscope is similar), the read of the output registers can be performed as follows:

- 1. Read STATUS REG.
- 2. If $XLDA = 0$, then go to 1.
- 3. Read OUTX L A.
- 4. Read OUTX H A.
- 5. Read OUTY L A.
- 6. Read OUTY H A.
- 7. Read OUTZ L A.
- 8. Read OUTZ H A.
- 9. Data processing

10. Go to 1.

4.3 Using the data-ready signal

The device can be configured to have a hardware signal to determine when a new set of measurement data is available to be read.

For the accelerometer sensor, the data-ready signal is represented by the XLDA bit of the STATUS_REG register. The signal can be driven to the INT1 pin by setting the INT1_DRDY_XL bit of the INT1_CTRL register to 1 and to the INT2 pin by setting the INT2_DRDY_XL bit of the INT2_CTRL register to 1.

For the gyroscope sensor, the data-ready signal is represented by the GDA bit of the STATUS_REG register. The signal can be driven to the INT1 pin by setting the INT1_DRDY_G bit of the INT1_CTRL register to 1 and to the INT2 pin by setting the INT2_DRDY_G bit of the INT2_CTRL register to 1.

The data-ready signal rises to 1 when a new set of data has been generated and it is available to be read. The data-ready signal can be either latched or pulsed. If the DRDY_PULSED bit of the CTRL4 register is set to 0 (default value), then the data-ready signal is latched and the interrupt is reset when the higher part of one axis is read (29h, 2Bh, 2Dh registers for the accelerometer; 23h, 25h, 27h registers for the gyroscope). If the DRDY PULSED bit of the CTRL4 register is set to 1, then the data-ready is pulsed and the duration of the pulse observed on the interrupt pin is 65 μs. If either the accelerometer or gyroscope is configured in HAODR mode, the duration of the pulse observed on the interrupt pin is 43 μs. Pulsed mode is not applied to the XLDA and GDA bits, which are always latched.

Figure 9. Data-ready signal

4.3.1 DRDY mask functionality

Setting the DRDY_MASK bit of the CTRL4 register to 1, the accelerometer and gyroscope data-ready signals are masked until the settling of the sensor filters is completed.

When FIFO is active and the DRDY_MASK bit is set to 1, accelerometer/gyroscope invalid samples stored in FIFO can be equal to 7FFFh, 7FFEh or 7FFDh. In this way, a tag is applied to the invalid samples stored in the FIFO buffer so that they can be easily identified and discarded during data post-processing.

Referring to the accelerometer UI chain, the DRDY mask functionality operates on all the power modes, full scales, and ODRs, considering also runtime changes. It covers the HP or LPF2 filter configuration up to ODR / 20 and it can be combined with the XL_FASTSETTL_MODE bit of the CTRL9 register for managing all the other filter configurations. If both the DRDY_MASK and the XL_FASTSETTL_MODE bits are set to 1, all the data-ready signals are masked until the internal filters are settled.

Referring to the gyroscope UI chain, the DRDY mask functionality operates on all the power modes, full scales, and ODRs, considering also runtime changes.

Referring to the gyroscope EIS chain, the DRDY mask functionality operates only during the settling of the drive circuit.

The DRDY mask feature is not available for HAODR mode.

Note: If the DRDY mask functionality is enabled, in order to guarantee the proper masking of the accelerometer sensor data-ready signal until the settling of the accelerometer filtering chain is completed, the following procedure must *be implemented when the accelerometer ODR is intended to be changed:*

1. Set the ODR_XL_[3:0] bits in the CTRL1 register to 0000 (power-down mode).

2. Wait x time period (x = 300 μs if the target operating mode is high-performance or high-accuracy ODR mode,

x = 1.2 ms if the target operating mode is normal mode).

3. Set the ODR_XL_[3:0] bits in the CTRL1 register to the desired value.

This procedure is not necessary if the accelerometer target operating mode is low-power mode.

4.4 Using the block data update (BDU) feature

If reading the accelerometer/gyroscope data is not synchronized with either the XLDA/GDA bits in the STATUS REG register or with the data-ready signal driven to the INT1/INT2 pins, it is strongly recommended to set the BDU (block data update) bit to 1 in the CTRL3 register.

This feature avoids reading values (most significant and least significant bytes of the output data) related to different samples. In particular, when the BDU is activated, the data registers related to each axis always contain the most recent output data produced by the device, but, in case the read of a given pair (that is, OUTX_H_A(G) and OUTX_L_A(G), OUTY_H_A(G) and OUTY_L_A(G), OUTZ_H_A(G) and OUTZ_L_A(G)) is initiated, the refresh for that pair is blocked until both the MSB and LSB of the data are read.

Note: BDU only guarantees that the LSB and MSB have been sampled at the same moment. For example, if the reading speed is too slow, X and Y can be read at T1 and Z sampled at T2.

> The BDU feature also acts on the FIFO_STATUS1 and FIFO_STATUS2 registers. When the BDU bit is set to 1, it is mandatory to read FIFO_STATUS1 first and then FIFO_STATUS2.

4.5 Understanding output data

The measured acceleration data are sent to the OUTX_H_A, OUTX_L_A, OUTY_H_A, OUTY_L_A, OUTZ_H_A, and OUTZ L_A registers. These registers contain, respectively, the most significant part and the least significant part of the acceleration signals acting on the X, Y, and Z axes.

The measured angular rate data are sent to the OUTX_H_G, OUTX_L_G, OUTY_H_G, OUTY_L_G, OUTZ_H_G, and OUTZ L_G registers. These registers contain, respectively, the most significant part and the least significant part of the angular rate signals acting on the X, Y, and Z axes.

The complete output data for the X, Y, Z channels is given by the concatenation OUTX H A(G) & OUTX L A(G), OUTY_H_A(G) & OUTY_L_A(G) , OUTZ_H_A(G) & OUTZ_L_A(G) and it is expressed as a two's complement number.

Both acceleration data and angular rate data are represented as 16-bit numbers. In order to translate them to their corresponding physical representation, a sensitivity parameter must be applied. This sensitivity value depends on the selected full-scale range (refer to the datasheet). In detail:

- Each acceleration sample must be multiplied by the proper sensitivity parameter LA_So (linear acceleration sensitivity expressed in m*g*/LSB) in order to obtain the corresponding value in m*g*.
- Each angular rate sample must be multiplied by the proper sensitivity parameter G_So (angular rate sensitivity expressed in mdps/LSB) in order to obtain the corresponding value in mdps.

4.5.1 Examples of output data

Table 27. Content of output data registers vs. acceleration (FS_XL = ±4 *g*) provides a few basic examples of the accelerometer data that is read in the data registers when the device is subjected to a given acceleration. Table 28. Content of output data registers vs. angular rate (FS $G = \pm 250$ dps) provides a few basic examples of the gyroscope data that is read in the data registers when the device is subjected to a given angular rate. The values listed in the following tables are given under the hypothesis of perfect device calibration (that is, no offset, no gain error, and so on).

Table 27. Content of output data registers vs. acceleration (FS_XL = ±4 *g***)**

4.6 Accelerometer offset registers

The device provides accelerometer offset registers (X_OFS_USR, Y_OFS_USR, Z_OFS_USR) which can be used for zero-*g* offset correction or, in general, to apply an offset to the accelerometer output data.

The accelerometer offset block can be enabled by setting the USR_OFF_ON_OUT bit of the CTRL9 register. The offset value set in the offset registers is internally subtracted from the measured acceleration value for the respective axis. Internally processed data are then sent to the accelerometer output register and to the FIFO (if enabled). These register values are expressed as an 8-bit word in two's complement and must be in the range [-127, 127].

The weight [*g*/LSB] to be applied to the offset register values is independent of the accelerometer selected full scale and can be configured using the USR_OFF_W bit of the CTRL9 register:

- 2 -10 *g*/LSB if the USR_OFF_W bit is set to 0
- 2 -6 *g*/LSB if the USR_OFF_W bit is set to 1

4.7 DEN (data enable)

The device allows an external trigger level recognition by configuring the LVL1_EN and LVL2_EN bits in the DEN register.

Two different modes can be selected (see Table 29. DEN configurations):

- Level-sensitive trigger mode
- Level-sensitive latched mode

The data enable (DEN) input signal must be driven on the INT2 pin, which is configured as an input pin when one of these modes is enabled.

The DEN functionality is active by default on the gyroscope data only. To extend this feature to the accelerometer data, the bit DEN_XL_EN in the DEN register must be set to 1.

The DEN active level is low by default. It can be changed to active-high by setting the bit INT2_IN_LH in the CTRL4 register to 1.

Table 29. DEN configurations

4.7.1 Level-sensitive trigger mode

Level-sensitive trigger mode can be enabled by setting the LVL1_EN bit in the DEN register to 1, and the LVL2 EN bit in the DEN register to 0.

Once the level-sensitive trigger mode is enabled, the LSB bit of the selected data (in output registers and FIFO) is replaced by 1 if the DEN level is active, or 0 if the DEN level is not active. The selected data can be the X, Y, Z axes of the accelerometer or gyroscope sensor (see [Section 4.7.3: LSB selection for DEN stamping](#page-41-0) for details).

All data can be stored in the FIFO according to the FIFO settings.

Note that the DEN level is read internally just before the update of the data registers: if a level change occurs after the read, DEN is acknowledged in the next ODR.

If the DEN feature is enabled on the accelerometer sensor by asserting the DEN_XL_EN bit of the DEN register, the accelerometer and gyroscope sensors must be configured at the same ODR or the gyroscope must be set in power-down mode.

Figure 10 shows with magenta circles the samples stored in the FIFO with LSB = 0 (DEN not active) and with blue circles the samples stored in the FIFO with LSB = 1 (DEN active).

Figure 10. Level-sensitive trigger mode, DEN active-low

4.7.2 Level-sensitive latched mode

Level-sensitive latched mode can be enabled by setting the LVL1_EN and LVL2_EN bits in the DEN register to 1. When the level-sensitive latched mode is enabled, the LSB bit of the selected data (in output registers and FIFO) is normally set to 0 and becomes 1 only on the first sample after a pulse on the DEN pin.

Note that the DEN level is read internally read before the update of the data registers: if a level change occurs after the read, DEN is acknowledged in the next ODR.

If the DEN feature is enabled on the accelerometer sensor by asserting the DEN_XL_EN bit of the DEN register, the accelerometer and gyroscope sensors must be configured at the same ODR or the gyroscope must be set in power-down mode.

Data can be selected through the DEN_X, DEN_Y, DEN_Z, DEN_XL_G bits in the DEN register (see [Section 4.7.3: LSB selection for DEN stamping](#page-41-0) for details).

Figure 11 shows an example of level-sensitive latched mode with DEN active-low. After the pulse on the DEN pin, the sample with a magenta circle has the value 1 on the LSB bit. All the other samples have LSB bit 0.

Figure 11. Level-sensitive latched mode, DEN active-low

4.7.3 LSB selection for DEN stamping

When level-sensitive modes (trigger or latched) are used, it is possible to select which LSB have to contain the information related to the DEN input signal behavior. This information can be stamped on the accelerometer or gyroscope axes in accordance with bits DEN_X, DEN_Y, DEN_Z and DEN_XL_G of the DEN register. Setting the DEN_X, DEN_Y, DEN_Z bits to 1, DEN information is stamped in the LSB of the corresponding axes of the sensor selected with the DEN_XL_G bit. By setting DEN_XL_G to 0, the DEN information is stamped in the selected gyroscope axes, while by setting DEN_XL_G to 1, the DEN information is stamped in the selected accelerometer axes.

By default, the bits are configured to have information on all the gyroscope axes.

5 Interrupt generation

V

Interrupt generation is based on accelerometer data only, so, for interrupt-generation purposes, the accelerometer sensor has to be set in an active operating mode (not in power-down). The gyroscope sensor can be configured in power-down mode since it is not involved in interrupt generation.

The interrupt generator can be configured to detect:

- **Free-fall**
- Wake-up
- 6D/4D orientation detection
- Single-tap and double-tap sensing
- Activity/inactivity and motion/stationary recognition

The device can also efficiently run the sensor-related features specified in Android, saving power and enabling faster reaction time. The following functions are implemented in the hardware:

- Significant motion
- Relative tilt
- Pedometer functions
- **Timestamp**
- Sensor fusion functions (game rotation vector, gravity vector, gyroscope bias)

Moreover, the device can be configured to generate interrupt signals activated by user-defined motion patterns. To do this, up to 8 embedded finite state machines can be programmed independently for motion detection or gesture recognition such as glance, absolute wrist tilt, shake, double-shake, or pick-up. Furthermore up to 4 decision trees can simultaneously and independently run inside the machine learning core logic.

The embedded finite state machine and the machine learning core features offer very high customization capabilities starting from scratch or importing activity/gesture recognition programs directly provided by STMicroelectronics. Please refer to the finite state machine application note and the machine learning core application note available on www.st.com.

All these interrupt signals, together with the FIFO interrupt signals, can be independently driven to the INT1 and INT2 interrupt pins or checked by reading the dedicated source register bits.

When the MIPI I3C[®] interface is used, information about the feature triggering the interrupt event is contained in the in-band interrupt (IBI) frame as described in the datasheet (default behavior). As an additional feature, by setting the INT_EN_I3C bit of the CTRL5 register to 1, the interrupt pins are activated even if using the MIPI I3C® interface.

The H_LACTIVE bit of the IF_CFG register must be used to select the polarity of the interrupt pins. If this bit is set to 0 (default value), the interrupt pins are active high and they change from low to high level when the related interrupt condition is verified. Otherwise, if the H_LACTIVE bit is set to 1 (active low), the interrupt pins are normally at high level and they change from high to low when interrupt condition is reached.

The PP_OD bit of the IF_CFG register allows changing the behavior of the interrupt pins from push-pull to open drain. If the PP_OD bit is set to 0, the interrupt pins are in push-pull configuration (low-impedance output for both high and low level). When the PP_OD bit is set to 1, only the interrupt active state is a low-impedance output.

5.1 Interrupt pin configuration

The device is provided with two pins that can be activated to generate either data-ready or interrupt signals. The functionality of these pins is selected through the MD1_CFG and INT1_CTRL registers for the INT1 pin, and through the MD2_CFG and INT2_CTRL registers for the INT2 pin.

A brief description of these interrupt control registers is given in the following summary. The default value of their bits is equal to 0, which corresponds to 'disable'. In order to enable routing a specific interrupt signal to the pin, the related bit has to be set to 1.

Table 30. INT1_CTRL register

- INT1_CNT_BDR: FIFO COUNTER_BDR_IA interrupt on INT1
- INT1_FIFO_FULL: FIFO full flag interrupt on INT1
- INT1_FIFO_OVR: FIFO overrun flag interrupt on INT1
- INT1_FIFO_TH: FIFO threshold interrupt on INT1
- INT1_DRDY_G: gyroscope data-ready on INT1
- INT1_DRDY_XL: accelerometer data-ready on INT1

Table 31. MD1_CFG register

- INT1_SLEEP_CHANGE: activity/inactivity recognition event interrupt on INT1
- INT1_SINGLE_TAP: single-tap interrupt on INT1
- INT1_WU: wake-up interrupt on INT1
- INT1_FF: free-fall interrupt on INT1
- INT1_DOUBLE_TAP: double-tap interrupt on INT1
- INT1_6D: 6D detection interrupt on INT1
- INT1_EMB_FUNC: embedded functions interrupt on INT1 (refer to [Section 6: Embedded functions](#page-61-0) for more details)
- INT1_SHUB: sensor hub end operation interrupt on INT1

Table 32. INT2_CTRL register

• INT2_EMB_FUNC_ENDOP: embedded functions end of operations interrupt on INT2. This pin is intended to be used for debug purposes. For this reason, it is not recommended to enable it if other interrupt signals are intended to be routed to the INT2 pin. When it is enabled, the INT2 pin is set to high level if no embedded function is running, otherwise, it is set to low level if any embedded function is running. For this reason, it can be used to measure the execution time of the embedded functions.

- INT2_CNT_BDR: FIFO COUNTER_BDR_IA interrupt on INT2
- INT2_FIFO_FULL: FIFO full flag interrupt on INT2
- INT2_FIFO_OVR: FIFO overrun flag interrupt on INT2
- INT2_FIFO_TH: FIFO threshold interrupt on INT2
- INT2_DRDY_G_EIS: gyroscope EIS data-ready interrupt on INT2

- INT2_DRDY_G: gyroscope data-ready on INT2
- INT2_DRDY_XL: accelerometer data-ready on INT2

The INT2_DRDY_TEMP bit of the CTRL4 register enables the temperature data-ready interrupt on the INT2 pin. The INT2_DRDY_AH_QVAR bit of the CTRL7 register enables the analog hub (or Qvar) data-ready interrupt on the INT2 pin.

Table 33. MD2_CFG register

- INT2_SLEEP_CHANGE: activity/inactivity recognition event interrupt on INT2
- INT2_SINGLE_TAP: single-tap interrupt on INT2
- INT2_WU: wake-up interrupt on INT2
- INT2 FF: free-fall interrupt on INT2
- INT2_DOUBLE_TAP: double-tap interrupt on INT2
- INT2_6D: 6D detection interrupt on INT2
- INT2_EMB_FUNC: embedded functions interrupt on INT2 (refer to [Section 6: Embedded functions](#page-61-0) for more details)
- INT2_TIMESTAMP: timestamp overflow alert interrupt on INT2

If multiple interrupt signals are routed to the same interrupt pin, the logic level of this pin is the "OR" combination of the selected interrupt signals. In order to know which event has generated the interrupt condition, the related source registers have to be read:

- WAKE_UP_SRC, TAP_SRC, D6D_SRC (basic interrupt functions)
- STATUS REG (for data-ready signals)
- EMB_FUNC_STATUS_MAINPAGE / EMB_FUNC_STATUS (for embedded functions)
- FSM_STATUS_MAINPAGE / FSM_STATUS (for finite state machine)
- MLC_STATUS_MAINPAGE / MLC_STATUS (for machine learning core)
- STATUS_MASTER_MAINPAGE / STATUS_MASTER (for sensor hub)
- FIFO STATUS2 (for FIFO)

The ALL_INT_SRC register groups the basic interrupts functions event status (6D/4D, free-fall, wake-up, tap, activity/inactivity) and the embedded functions and sensor hub interrupt status in a single register. It is possible to read this register in order to address a subsequent specific source register read.

The INT2 on INT1 pin of the CTRL4 register allows driving some specific interrupt signals in logic "OR" on the INT1 pin (by setting this bit to 1). When this bit is set to 0, the interrupt signals are divided between the INT1 and INT2 pins. When this bit is set to 1, the movable interrupts are INT2_DRDY_G_EIS and

INT2_EMB_FUNC_ENDOP (enabled through the INT2_CTRL register), INT2_TIMESTAMP (enabled through the MD2_CFG register), INT2_DRDY_TEMP (enabled through the CTRL4 register), and INT2_DRDY_AH_QVAR (enabled through the CTRL7 register).

The basic interrupts have to be enabled by setting the INTERRUPTS_ENABLE bit in the FUNCTIONS_ENABLE register.

The LIR bit of the TAP_CFG0 register enables the latched interrupt for the basic interrupt functions: when this bit is set to 1 and the interrupt flag is sent to the INT1 pin and/or INT2 pin, the interrupt remains active until the ALL_INT_SRC register or the corresponding source register is read. The latched interrupt is enabled on a function only if a function is routed to the INT1 or INT2 pin: if latched mode is enabled but the interrupt signal is not driven to the interrupt pins, the latch feature does not take effect. The DIS_RST_LIR_ALL_INT bit of the FUNCTIONS_ENABLE register can be set to 1 in order to avoid resetting the latched interrupt signals by reading the ALL_INT_SRC register. This feature is useful in order to not reset some status flags before reading the corresponding status register.

5.2 Free-fall interrupt

Free-fall detection refers to a specific register configuration that allows recognizing when the device is in free-fall: the acceleration measured along all the axes goes to zero. In a real case a "free-fall zone" is defined around the zero-*g* level where all the accelerations are small enough to generate the interrupt. Configurable threshold and duration parameters are associated to free-fall event detection. The threshold parameter defines the free-fall zone amplitude; the duration parameter defines the minimum duration of the free-fall interrupt event to be recognized (Figure 12).

The free-fall interrupt signal can be enabled by setting the INTERRUPTS_ENABLE bit in the FUNCTIONS ENABLE register to 1 and can be driven to the two interrupt pins by setting the INT1_FF bit of the MD1_CFG register to 1 or the INT2_FF bit of the MD2_CFG register to 1. It can also be checked by reading the FF_IA bit of the WAKE_UP_SRC or ALL_INT_SRC register.

If latched mode is disabled (LIR bit of TAP_CFG0 is set to 0), the interrupt signal is automatically reset when the free-fall condition is no longer verified. If latched mode is enabled and the free-fall interrupt signal is driven to the interrupt pins, once a free-fall event has occurred and the interrupt pin is asserted, it must be reset by reading the WAKE_UP_SRC or ALL_INT_SRC register. If latched mode is enabled but the interrupt signal is not driven to the interrupt pins, the latch feature does not take effect.

The FREE_FALL register is used to configure the threshold parameter. The unsigned threshold value is related to the value of the FF_THS_[2:0] field value as indicated in Table 34. Free-fall threshold LSB value. The values given in this table are valid for each accelerometer full-scale value.

Table 34. Free-fall threshold LSB value

Duration time is measured in N/ODR_XL, where N is the content of the FF_DUR_[5:0] field of the FREE_FALL / WAKE_UP_DUR registers and ODR_XL is the accelerometer data rate.

A basic software routine for free-fall event recognition is given below.

1. Write 08h to CTRL1 // Turn on the accelerometer (ODR = 480 Hz) 2. Write 01h to TAP_CFG0 // Enable latched mode 3. Write 80h to FUNCTIONS ENABLE // Enable interrupt functions 4. Write 00h to WAKE_UP_DUR // Set event duration (FF_DUR_5 bit) 5. Write 33h to FREE_FALL // Set FF threshold (FF_THS_[2:0] = 011) // Set six samples event duration (FF_DUR_ $[5:0] = 000110$) 6. Write 10h to MD1 CFG // FF interrupt driven to INT1 pin

The sample code sets the threshold to 312 m*g* for free-fall recognition and the event is notified by hardware through the INT1 pin. The FF_DUR_[5:0] field of the FREE_FALL / WAKE_UP_DUR registers is configured to ignore events that are shorter than $6/ODR$ XL = $6/480$ Hz $\overline{~}$ = 12.5 msec in order to avoid false detections.

5.3 Wake-up interrupt

The wake-up feature can be implemented using either the slope filter (see [Section 3.6.1: Accelerometer slope](#page-25-0) [filter](#page-25-0) for more details) or the high-pass digital filter, as illustrated in [Figure 3. Accelerometer filtering chain \(UI](#page-23-0) [path\).](#page-23-0) The filter to be applied can be selected using the SLOPE_FDS bit of the TAP_CFG0 register. If this bit is set to 0 (default value), the slope filter is used; if it is set to 1, the HPF digital filter is used. Moreover, it is possible to configure the wake-up feature as an absolute wake-up with respect to a programmable position. This can be done by setting both the SLOPE_FDS bit of the TAP_CFG0 register and the USR_OFF_ON_WU bit of the WAKE_UP_THS register to 1. Using this configuration, the input data for the wake-up function comes from the low-pass filter path and the programmable position is subtracted as an offset. The programmable position can be configured through the X_OFS_USR, Y_OFS_USR and Z_OFS_USR registers (refer to [Section 4.6: Accelerometer offset registers](#page-39-0) for more details).

The wake-up interrupt signal is generated if a certain number of consecutive filtered data exceed the configured threshold ([Figure 13. Wake-up interrupt \(using the slope filter\)](#page-47-0)).

The unsigned threshold value is defined using the WK_THS_[5:0] bits of the WAKE_UP_THS register. The value of 1 LSB of these 6 bits depends on the value of the WU_INACT_THS_W_[2:0] bits of the INACTIVITY_DUR register as shown in the table below.

Table 35. Wake-up threshold resolution

The threshold is applied to both positive and negative data: for wake-up interrupt generation, the absolute value of the filtered data must be bigger than the threshold.

The duration parameter defines the minimum duration of the wake-up event to be recognized. Its value is set using the WAKE_DUR_[1:0] bits of the WAKE_UP_DUR register: 1 LSB corresponds to 1/ODR_XL time, where ODR_{XL} is the accelerometer output data rate. It is important to appropriately define the duration parameter to avoid unwanted wake-up interrupts due to spurious spikes of the input signal.

This interrupt signal can be enabled by setting the INTERRUPTS_ENABLE bit in the FUNCTIONS_ENABLE register to 1 and can be driven to the two interrupt pins by setting the INT1_WU bit of the MD1_CFG register or the INT2 WU bit of the MD2 CFG register to 1. It can also be checked by reading the WU IA bit of the WAKE_UP_SRC or ALL_INT_SRC register. The X_WU, Y_WU, Z_WU bits of the WAKE_UP_SRC register indicate which axes have triggered the wake-up event.

Figure 13. Wake-up interrupt (using the slope filter)

If latched mode is disabled (LIR bit of TAP_CFG0 is set to 0), the interrupt signal is automatically reset when the filtered data falls below the threshold. If latched mode is enabled and the wake-up interrupt signal is driven to the interrupt pins, once a wake-up event has occurred and the interrupt pin is asserted, it must be reset by reading the WAKE_UP_SRC register or the ALL_INT_SRC register. The X_WU, Y_WU, Z_WU bits are maintained at the state in which the interrupt was generated until the read is performed. In case the WU_X, WU_Y, WU_Z bits have to be evaluated (in addition to the WU_IA bit), it is recommended to directly read the WAKE_UP_SRC register (do not use ALL_INT_SRC register for this specific case). If latched mode is enabled but the interrupt signal is not driven to the interrupt pins, the latch feature does not take effect.

A basic software routine for wake-up event recognition using the high-pass digital filter is given below.

-
-
- 3. Write 01h to WAKE_UP_THS // Set wake-up threshold
- 4. Write 00h to WAKE_UP_DUR // Set duration to 0
-
- 6. Write 80h to FUNCTIONS ENABLE // Enable interrupt functions
-
- 1. Write 34h to INACTIVITY_DUR *// Set wake-up threshold resolution to 62.5 mg*
- 2. Write 11h to TAP_CFG0 // Select HPF path and enable latched mode
	-
	-
- 5. Write 20h to MD1_CFG // Wake-up interrupt driven to INT1 pin
	-
- 7. Write 08h to CTRL1 // Turn on the accelerometer (ODR = 480 Hz)

Since the duration time is set to zero, the wake-up interrupt signal is generated for each X, Y, Z filtered data exceeding the configured threshold. The WK_THS field of the WAKE_UP_THS register is set to 000001 and the resolution of 1 LSB is set to 62.5 m*g* (WU_INACT_THS_W_[2:0] bits of INACTIVITY_DUR register are set to 011), therefore the wake-up threshold is 62.5 m*g*.

Since the wake-up functionality is implemented using the slope/high-pass digital filter, it is necessary to consider the settling time of the filter just after this functionality is enabled. For example, when using the slope filter (but a similar consideration can be done for the high-pass digital filter usage) the wake-up functionality is based on the comparison of the threshold value with half of the difference of the acceleration of the current (X, Y, Z) sample and the previous one (refer to [Section 3.6.1: Accelerometer slope filter](#page-25-0)).

At the very first sample, the slope filter output is calculated as half of the difference of the current sample, for example $(X, Y, Z) = (0, 0, 1)$ *g*, with the previous one, which is $(X, Y, Z) = (0, 0, 0)$ *g* since no sample has been generated yet. For this reason, on the Z-axis the first output value of the slope filter is (1 - 0) / 2 = 0.5 *g* = 500 m*g* and it could be higher than the threshold value in which case a spurious interrupt event is generated. The interrupt signal is kept high for 1 ODR then it goes low.

In order to avoid the spurious interrupt generation due to the settling of the digital slope / high-pass filter, it is possible to mask the execution trigger of the basic interrupt functions during the digital filter settling by configuring to 1 both the XL_FASTSETTL_MODE bit of the CTRL9 register and the HW_FUNC_MASK_XL_SETTL of the TAP_CFG0 register.

The wake-up configuration procedure described above can be easily modified as follows:

- 1. Write 20h to CTRL9 // Set XL_FASTSETTL_MODE = 1
- 2. Write 34h to INACTIVITY_DUR // Set wake-up threshold resolution to 62.5 mg
- 3. Write 31h to TAP_CFG0 // Set HW_FUNC_MASK_XL_SETTL = 1, select HPF path and enable latched mode
- 4. Write 01h to WAKE_UP_THS // Set wake-up threshold
- 5. Write 00h to WAKE_UP_DUR // Set duration to 0
- 6. Write 20h to MD1 CFG // Wake-up interrupt driven to INT1 pin
- 7. Write 80h to FUNCTIONS_ENABLE // Enable interrupt functions
- 8. Write 08h to CTRL1 // Turn on the accelerometer (ODR = 480 Hz)

5.4 6D/4D orientation detection

The accelerometer provides the capability to detect the orientation of the device in space, enabling easy implementation of energy-saving procedures and automatic screen rotation for mobile devices.

5.4.1 6D orientation detection

Six orientations of the device in space can be detected. The interrupt signal is asserted when the device switches from one orientation to another. The interrupt is not reasserted as long as the position is maintained.

6D interrupt is generated when, for two consecutive samples, only one axis exceeds a selected threshold and the acceleration values measured from the other two axes are lower than the threshold: the ZH, ZL, YH, YL, XH, XL bits of the D6D_SRC register indicate which axis has triggered the 6D event. In more detail:

Table 36. D6D_SRC register

D6D IA is set high when the device switches from one orientation to another.

- ZH (YH, XH) is set high when the face perpendicular to the Z (Y, X) axis is almost flat and the acceleration measured on the Z (Y, X) axis is positive and in the absolute value bigger than the threshold.
- ZL (YL, XL) is set high when the face perpendicular to the Z (Y, X) axis is almost flat and the acceleration measured on the Z (Y, X) axis is negative and in the absolute value bigger than the threshold.

The SIXD_THS_[1:0] bits of the TAP_THS_6D register are used to select the threshold value used to detect the change in device orientation. The threshold values given in the following table are valid for each accelerometer full-scale value.

Table 37. Threshold for 4D/6D function

The low-pass filter LPF2 can also be used in 6D functionality by setting the LOW_PASS_ON_6D bit of the TAP CFG0 register to 1.

This interrupt signal can be enabled by setting the INTERRUPTS_ENABLE bit in the FUNCTIONS_ENABLE register to 1 and can be driven to the two interrupt pins by setting the INT1_6D bit of the MD1_CFG register or the INT2_6D bit of the MD2_CFG register to 1. It can also be checked by reading the D6D_IA bit of the D6D_SRC or ALL_INT_SRC register.

If latched mode is disabled (LIR bit of TAP_CFG is set to 0), the interrupt signal is active only for 1/ODR_XL then it is automatically disserted (ODR_XL is the accelerometer output data rate). If latched mode is enabled and the 6D interrupt signal is driven to the interrupt pins, once an orientation change has occurred and the interrupt pin is asserted, a read of the D6D_SRC or ALL_INT_SRC register clears the request and the device is ready to recognize a different orientation. The XL, XH, YL, YH, ZL, ZH bits are not affected by the LIR configuration. They correspond to the current state of the device when the D6D_SRC register is read. If latched mode is enabled but the interrupt signal is not driven to the interrupt pins, the latch feature does not take effect.

Referring to the six possible cases illustrated in [Figure 14. 6D recognized orientations](#page-50-0), the content of the D6D_SRC register for each position is shown in [Table 38. D6D_SRC register in 6D positions.](#page-50-0)

Figure 14. 6D recognized orientations

Table 38. D6D_SRC register in 6D positions

A basic software routine for 6D orientation detection is as follows.

-
-
- 3. Write 04h to MD1 CFG // 6D interrupt driven to INT1 pin
- 4. Write 80h to FUNCTIONS_ENABLE // Enable interrupt functions
-
- 1. Write 41h to TAP_CFG0 // Enable LPF2 filter for 6D functionality and latched mode
- 2. Write 40h to TAP_THS_6D // Set 6D threshold (SIXD_THS_[1:0] = 10 = 60 degrees)
	-
	-
- 5. Write 08h to CTRL1 // Turn on the accelerometer (ODR = 480 Hz)

5.4.2 4D orientation detection

The 4D direction function is a subset of the 6D function especially defined to be implemented in mobile devices for portrait and landscape detection. It can be enabled by setting the D4D_EN bit of the TAP_THS_6D register to 1. In this configuration, the Z-axis position detection is disabled, therefore reducing position recognition to cases (a), (b), (c), and (d) of Table 38. D6D_SRC register in 6D positions.

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5.5 Single-tap and double-tap recognition

The single-tap and double-tap recognition help to create a man-machine interface with little software loading. The device can be configured to output an interrupt signal on a dedicated pin when tapped in any direction.

If the sensor is exposed to a single input stimulus, it generates an interrupt signal on the interrupt pin INT1 and/or INT2. A more advanced feature allows the generation of an interrupt signal when a double input stimulus with programmable time between the two events is recognized, enabling a mouse button-like function.

The single-tap and double-tap recognition functions use the slope between two consecutive acceleration samples to detect the tap events. The slope data is calculated using the following formula:

 $slope(t_n) = [acc(t_n) - acc(t_{n-1})]/2$

This function can be fully programmed by the user in terms of expected amplitude and timing of the slope data by means of a dedicated set of registers.

Single and double-tap recognition work based on the selected output data rate. The recommended minimum accelerometer ODR for these functions is 480 Hz.

In order to enable the single-tap and double-tap recognition functions it is necessary to set the INTERRUPTS_ENABLE bit in the FUNCTIONS_ENABLE register to 1.

5.5.1 Single tap

If the device is configured for single-tap event detection, an interrupt is generated when the slope data of the selected axis exceeds the programmed threshold, and returns below it within the shock time window.

In the single-tap case, if the LIR bit of the TAP_CFG0 register is set to 0, the interrupt is kept active for the duration of the quiet window. If the LIR bit is set to 1, the interrupt is kept active until the TAP_SRC or ALL_INT_SRC register is read.

The SINGLE_DOUBLE_TAP bit of WAKE_UP_THS has to be set to 0 in order to enable single-tap recognition only.

In case (a) of Figure 15. Single-tap event recognition the single-tap event has been recognized, while in case (b) the tap has not been recognized because the slope data falls below the threshold after the shock time window has expired.

Figure 15. Single-tap event recognition

5.5.2 Double tap

If the device is configured for double-tap event detection, an interrupt is generated when, after a first tap, a second tap is recognized. The recognition of the second tap occurs only if the event satisfies the rules defined by the shock, the quiet and the duration time windows.

In particular, after the first tap has been recognized, the second tap detection procedure is delayed for an interval defined by the quiet time. This means that after the first tap has been recognized, the second tap detection procedure starts only if the slope data exceeds the threshold after the quiet window but before the duration window has expired. In case (a) of Figure 16, a double-tap event has been correctly recognized, while in case (b) the interrupt has not been generated because the slope data exceeds the threshold after the window interval has expired.

Once the second tap detection procedure is initiated, the second tap is recognized with the same rule as the first: the slope data must return below the threshold before the shock window has expired.

It is important to appropriately define the quiet window to avoid unwanted taps due to spurious bouncing of the input signal.

In the double-tap case, if the LIR bit of the TAP_CFG0 register is set to 0, the interrupt is kept active for the duration of the quiet window. If the LIR bit is set to 1, the interrupt is kept active until the TAP_SRC or ALL_INT_SRC register is read.

Figure 16. Double-tap event recognition (LIR bit = 0)

5.5.3 Single-tap and double-tap recognition configuration

The device can be configured to output an interrupt signal when tapped (once or twice) in any direction: the TAP_X_EN, TAP_Y_EN and TAP_Z_EN bits of the TAP_CFG0 register must be set to 1 to enable the tap recognition on the X, Y, Z directions, respectively. In addition, the INTERRUPTS_ENABLE bit of the FUNCTIONS_ENABLE register has to be set to 1.

Configurable parameters for tap recognition functionality are the tap thresholds (each axis has a dedicated threshold) and the shock, quiet and duration time windows.

The TAP_THS_X_[4:0] bits of the TAP_CFG1 register, the TAP_THS_Y_[4:0] bits of the TAP_CFG2 register and the TAP_THS_Z_[4:0] bits of the TAP_THS_6D register are used to select the unsigned threshold value used to detect the tap event on the respective axis. The value of 1 LSB of these 5 bits depends on the selected accelerometer full scale: 1 LSB = FS_XL / 2^5 . The unsigned threshold is applied to both positive and negative slope data.

Both single-tap and double-tap recognition functions apply to only one axis. If more than one axis are enabled and they are over the respective threshold, the algorithm continues to evaluate only the axis with highest priority. The priority can be configured through the TAP_PRIORITY_[2:0] bits of TAP_CFG1. The following table shows all the possible configurations.

Table 39. TAP_PRIORITY_[2:0] bits configuration

The shock time window defines the maximum duration of the overcoming threshold event: the acceleration must return below the threshold before the shock window has expired, otherwise the tap event is not detected. The SHOCK [1:0] bits of the TAP_DUR register are used to set the shock time window value: the default value of these bits is 00 and corresponds to 4/ODR_XL time, where ODR_XL is the accelerometer output data rate. If the SHOCK [1:0] bits are set to a different value, 1 LSB corresponds to 8/ODR XL time.

In the double-tap case, the quiet time window defines the time after the first tap recognition in which there must not be any overcoming threshold event. When latched mode is disabled (LIR bit of TAP_CFG is set to 0), the quiet time also defines the length of the interrupt pulse (in both single and double-tap case). The QUIET [1:0] bits of the TAP_DUR register are used to set the quiet time window value: the default value of these bits is 00 and corresponds to 2/ODR_XL time, where ODR_XL is the accelerometer output data rate. If the QUIET_[1:0] bits are set to a different value, 1 LSB corresponds to 4/ODR_XL time.

In the double-tap case, the duration time window defines the maximum time between two consecutive detected taps. The duration time period starts just after the completion of the quiet time of the first tap. The DUR_[3:0] bits of the TAP_DUR register are used to set the duration time window value: the default value of these bits is 0000 and corresponds to 16/ODR_XL time, where ODR_XL is the accelerometer output data rate. If the DUR_[3:0] bits are set to a different value, 1 LSB corresponds to 32/ODR_XL time.

[Figure 17. Single and double-tap recognition \(LIR bit = 0\)](#page-55-0) illustrates a single-tap event (a) and a double-tap event (b). These interrupt signals can be driven to the two interrupt pins by setting the INT1_SINGLE_TAP bit of the MD1_CFG register or the INT2_SINGLE_TAP bit of the MD2_CFG register to 1 for the single-tap case, and setting the INT1_DOUBLE_TAP bit of the MD1_CFG register or the INT2_DOUBLE_TAP bit of the MD2_CFG register to 1 for the double-tap case.

Figure 17. Single and double-tap recognition (LIR bit = 0)

Tap interrupt signals can also be checked by reading the TAP_SRC (1Ch) register, described in the following table.

- TAP IA is set high when a single-tap or double-tap event has been detected.
- SINGLE_TAP is set high when a single tap has been detected.
- DOUBLE_TAP is set high when a double tap has been detected.
- TAP_SIGN indicates the acceleration sign when the tap event is detected. It is set low in case of positive sign and it is set high in case of negative sign.
- X_TAP (Y_TAP, Z_TAP) is set high when the tap event has been detected on the X (Y, Z) axis.

Single and double-tap recognition works independently. Setting the SINGLE_DOUBLE_TAP bit of the WAKE_UP_THS register to 0, only the single-tap recognition is enabled: double-tap recognition is disabled and cannot be detected. When the SINGLE_DOUBLE_TAP is set to 1, both single and double-tap recognition are enabled.

If latched mode is enabled and the interrupt signal is driven to the interrupt pins, the value assigned to SINGLE_DOUBLE_TAP also affects the behavior of the interrupt signal. When it is set to 0, the latched mode is applied to the single-tap interrupt signal; when it is set to 1, the latched mode is applied to the double-tap interrupt signal only. The latched interrupt signal is kept active until the TAP_SRC or ALL_INT_SRC register is read. The TAP_SIGN, X_TAP, Y_TAP, Z_TAP bits are maintained at the state in which the interrupt was generated until the read is performed. In case the TAP_SIGN, X_TAP, Y_TAP, Z_TAP bits have to be evaluated (in addition to the TAP_IA bit), it is recommended to directly read the TAP_SRC register (do not use ALL_INT_SRC register for this specific case). If latched mode is enabled but the interrupt signal is not driven to the interrupt pins, the latch feature does not take effect.

5.5.4 Single-tap example

A basic software routine for single-tap detection is given below.

10. Write 08h to CTRL1 // Turn on the accelerometer (480 Hz)

In this example the TAP_THS_Z_[4:0] bits are set to 00010, therefore the tap threshold for the Z-axis is 500 m*g* $(= 2 * FS$ XL / 2^5).

The SHOCK field of the TAP_DUR register is set to 10. An interrupt is generated when the slope data exceeds the programmed threshold, and returns below it within 33.3 ms (= $2 * 8 /$ ODR_XL) corresponding to the shock time window.

The QUIET field of the TAP_DUR register is set to 01. Since latched mode is disabled, the interrupt is kept high for the duration of the quiet window, therefore 8.3 ms $(= 1 * 4 / ODR XL)$.

5.5.5 Double-tap example

A basic software routine for double-tap detection is given below.

In this example the TAP_THS_Z_[4:0] bits are set to 00011, therefore the tap threshold is 750 mg $(3 * FS$ XL $/$ 2^5).

For interrupt generation, during the first and the second tap the slope data must return below the threshold before the shock window has expired. The SHOCK field of the TAP DUR register is set to 11, therefore the shock time is 50 ms (= 3 * 8 / ODR_XL).

For interrupt generation, after the first tap recognition there must not be any slope data overthreshold during the quiet time window. Furthermore, since latched mode is disabled, the interrupt is kept high for the duration of the quiet window. The QUIET field of the TAP_DUR register is set to 11, therefore the quiet time is 25 ms $(= 3 * 4 / ODR XL)$.

For the maximum time between two consecutive detected taps, the DUR field of the TAP_DUR register is set to 0111, therefore the duration time is 533.3 ms $(= 8 * 32 / ODR XL)$.

5.6 Activity/inactivity and motion/stationary recognition

The working principle of activity/inactivity and motion/stationary embedded functions is similar to wake-up. If no movement condition is detected for a programmable time, an inactivity/stationary condition event is generated. Otherwise, when the accelerometer data exceed the configurable threshold, an activity/motion condition event is generated.

The activity/inactivity recognition function allows reducing system power consumption and developing new smart applications.

When the activity/inactivity recognition function is activated, the device is able to automatically switch the accelerometer power mode to low-power mode 1 and change the sampling rate to a configurable low ODR (available selectable ODRs are 1.875 Hz, 15 Hz, 30 Hz, 60 Hz) when the inactivity state is detected, while it is able to automatically switch back to the power mode and sampling rate selected through the OP_MODE_XL_[2:0] bits and ODR_XL_[3:0] bits of the CTRL1 register when the activity state is detected.

The target accelerometer ODR for the inactivity state can be selected through the XL_INACT_ODR_[1:0] bits of the INACTIVITY_DUR register, with the values indicated in the table below.

Table 41. Target accelerometer ODR configuration for inactivity event

This feature can be extended to the gyroscope, with three possible options:

- Gyroscope configurations do not change.
- Gyroscope enters in sleep mode.
- Gyroscope enters in power-down mode.

With this feature the system may be efficiently switched from low-power consumption to full performance and vice versa depending on user-selectable acceleration events, thus ensuring power saving and flexibility.

The activity/inactivity recognition function is enabled by setting the INTERRUPTS_ENABLE bit to 1 and configuring the INACT_EN_[1:0] bits of the FUNCTIONS_ENABLE register. If the INACT_EN_[1:0] bits of the FUNCTIONS_ENABLE register are equal to 00, the motion/stationary embedded function is enabled. Possible configurations of the inactivity event are summarized in the following table.

Table 42. Inactivity event configuration

The activity/inactivity and motion/stationary recognition functions can be implemented using either the slope filter (see [Section 3.6.1: Accelerometer slope filter](#page-25-0) for more details) or the high-pass digital filter, as illustrated in [Figure 3. Accelerometer filtering chain \(UI path\).](#page-23-0) The filter to be applied can be selected using the SLOPE_FDS bit of the TAP_CFG0 register. If this bit is set to 0 (default value), the slope filter is used; if it is set to 1, the highpass digital filter is used.

This function can be fully programmed by the user in terms of expected amplitude and timing of the filtered data by means of a dedicated set of registers [\(Figure 18. Activity/inactivity recognition \(using the slope filter\)\)](#page-59-0).

The unsigned threshold value is defined using the INACT_THS_[5:0] bits of the INACTIVITY_THS register. The value of 1 LSB of these 6 bits depends on the value of the WU_INACT_THS_W_[2:0] bits of the INACTIVITY_DUR register as shown in the following table.

Table 43. Activity/inactivity threshold resolution

The threshold is applied to both positive and negative filtered data.

When a certain number of consecutive X, Y, Z filtered data is smaller than the configured threshold, the OP_MODE_XL_[2:0] and the ODR_XL_[3:0] bits of the CTRL1 register are bypassed (inactivity) and the accelerometer is internally set in low-power mode 1 at the sampling rate configured through the XL_INACT_ODR_[1:0] bits of the INACTIVITY_DUR register, although the content of the CTRL1 register is left untouched. The gyroscope behavior varies according to the configuration of the INACT_EN_[1:0] bits of the FUNCTIONS_ENABLE register. The duration of the inactivity state to be recognized is defined by the SLEEP_DUR_[3:0] bits of the WAKE_UP_DUR register: 1 LSB corresponds to 512 / ODR_XL time, where ODR_XL is the accelerometer output data rate. If the SLEEP_DUR_[3:0] bits are set to 0000, the duration of the inactivity state to be recognized is equal to 16 / ODR_XL time.

When the inactivity state is detected, the interrupt is set high for 1/ODR_XL[s] period then it is automatically deasserted.

When filtered data on one axis becomes bigger than the threshold for a configurable time, the CTRL1 register settings are immediately restored (activity) and the gyroscope is restored to the previous state. The duration of the activity state to be recognized is defined by the INACT_DUR_[1:0] bits of the INACTIVITY_DUR register. 1 LSB corresponds to 1 / ODR_XL time, where ODR_XL is the accelerometer output data rate.

When the activity state is detected, the interrupt is set high for 1 / ODR_XL[s] period then it is automatically deasserted.

Once the activity/inactivity detection function is enabled, the activity/inactivity event can be driven to the two interrupt pins by setting the INT1_SLEEP_CHANGE bit of the MD1_CFG register or the INT2_SLEEP_CHANGE bit of the MD2_CFG register to 1. The activity/inactivity event can also be checked by reading the SLEEP_CHANGE_IA bit of the WAKE_UP_SRC or ALL_INT_SRC register.

The SLEEP_CHANGE_IA bit is by default in pulsed mode. Latched mode can be selected by setting the LIR bit of the TAP_CFG0 register to 1 and the INT1_SLEEP_CHANGE of the MD1_CFG register or INT2_SLEEP_CHANGE of the MD2_CFG register to 1. The SLEEP_STATE bit of the WAKE_UP_SRC register is

not affected by the LIR configuration. It corresponds to the current state of the device when the WAKE_UP_SRC register is read.

By setting the SLEEP_STATUS_ON_INT bit of the INACTIVITY_DUR register to 1, the signal routed to the INT1 or INT2 pins is configured to be the activity/inactivity state (SLEEP_STATE bit of WAKE_UP_SRC register) instead of the sleep-change signal. It goes high during inactivity state and it goes low during activity state. Latched mode is not supported in this configuration.

A basic software routine for activity/inactivity detection is as follows:

In this example, the INACT_THS_[5:0] bits field of the INACTIVITY_THS register is set to 000001 and the resolution of 1 LSB is set to 62.5 m*g* (WU_INACT_THS_W_[2:0] bits of INACTIVITY_DUR register are set to 011), therefore the activity/inactivity threshold is 62.5 m*g*.

Before inactivity detection, the X, Y, Z slope data must be smaller than the configured threshold for a period of time defined by the SLEEP_DUR field of the WAKE_UP_DUR register: this field is set to 0101, corresponding to 5.33 s (= 5 * 512 / ODR_XL). After this period of time has elapsed, the accelerometer ODR is internally set to 15 Hz (XL_INACT_ODR_[1:0] = 01) and the gyroscope is internally set to power-down mode.

The activity state is detected and the CTRL1 register settings are immediately restored and the gyroscope is turned on as soon as the slope data of (at least) one axis is bigger than the threshold for one sample, since the INACT_DUR_[1:0] bits of the INACTIVITY_DUR register are configured to 00.

5.6.1 Stationary/motion detection

Stationary/motion detection is a particular case of the activity/inactivity functionality in which no ODR / power mode changes occur when a sleep condition (equivalent to stationary condition) is detected. Stationary/motion detection is activated by setting the INACT_EN_[1:0] bits of the FUNCTIONS_ENABLE register to 00.

5.7 Boot status

After the device is powered up, it performs a 10 ms (maximum) boot procedure to load the trimming parameters. After the boot is completed, both the accelerometer and the gyroscope are automatically configured in powerdown mode. During the boot time the registers are not accessible.

Note: If it is required to force a boot by removing and resupplying the Vdd and the time between removing and resupplying the Vdd is less than 20 ms, then the maximum boot time increases to 30 ms.

> After power-up, the trimming parameters can be reloaded by setting the BOOT bit of the CTRL3 register to 1. In this case, it is mandatory to wait 30 ms for the completion of the reboot internal procedure. The BOOT bit of the CTRL3 register automatically returns to 0.

If the reset to the default value of the control registers is required, it can be performed by setting the SW_RESET bit of the CTRL3 register to 1. When this bit is set to 1, the following registers are reset to their default value:

- FUNC_CFG_ACCESS (01h)
- ODR_TRIG_CFG (06h) through ALL_INT_SRC (1Dh)
- TIMESTAMP0 (40h) through TIMESTAMP3 (43h)
- WAKE_UP_SRC (45h) through D6D_SRC (47h)
- FUNCTIONS_ENABLE (50h) through UI_HANDSHAKE_CTRL (64h)
- CTRL_EIS (6Bh) through Z_OFS_USR (75h)
- FIFO DATA OUT TAG (78h)

The software reset procedure takes a maximum of 150 us. The status of reset is signaled by the status of the SW_RESET bit of the CTRL3 register. Once the reset is completed, this bit is automatically set low.

The reboot flow is as follows:

- 1. Set both the accelerometer and gyroscope in power-down mode.
- 2. Set the BOOT bit of the CTRL3 register to 1.
- 3. Wait 30 ms.

The software reset flow is as follows:

- 1. Set both the accelerometer and gyroscope in power-down mode.
- 2. Set the SW_RESET bit of the CTRL3 register to 1.
- 3. Monitor the software reset status. There are two possibilities:
	- a. Wait 150 µs.

b. Poll the SW_RESET bit of the CTRL3 register until it returns to 0.

In order to avoid conflicts, the reboot and the software reset must not be executed at the same time (do not set to 1 at the same time both the BOOT bit and the SW_RESET bit of CTRL3 register). The above flows must be performed serially.

If a complete reset (including the boot, software reset, and a reset of the embedded functions and internal filters) is required, it can be performed by setting the SW_POR bit of the FUNC_CFG_ACCESS register. When this bit is set to 1, the device triggers a complete reset of the device, analogous to a power-on-reset. In this case, it is mandatory to wait 30 ms for the completion of device reset. The SW_POR bit of the FUNC_CFG_ACCESS register automatically returns to 0. The complete reset flow is as follows:

- 1. Set the SW_POR bit of the FUNC_CFG_ACCESS register to 1.
- 2. Wait 30 ms.

6 Embedded functions

The device implements in the hardware many embedded functions. Specific IP blocks with negligible power consumption and high-level performance implement the following functions:

- Pedometer functions (step detector and step counter)
- Significant motion
- Relative tilt
- **Timestamp**
- Sensor fusion functions (game rotation vector, gravity vector, gyroscope bias)

6.1 Pedometer functions: step detector and step counter

A specific IP block is dedicated to pedometer functions: the step detector and the step counter. Pedometer functions work at 30 Hz and are based on the accelerometer sensor only. Consequently, the accelerometer ODR must be set at a value of 30 Hz or higher when using them.

In order to enable the pedometer functions, it is necessary to set the PEDO_EN bit of the EMB_FUNC_EN_A embedded functions register to 1. The algorithm internal state can be reinitialized by asserting the STEP_DET_INIT bit of the EMB_FUNC_INIT_A embedded functions register.

The step counter indicates the number of steps detected by the step detector algorithm after the pedometer function has been enabled. The step count is given by the concatenation of the STEP_COUNTER_H and STEP_COUNTER_L embedded functions registers and it is represented as a 16-bit unsigned number.

The step count is not reset to zero when the accelerometer is configured in power-down or the pedometer is disabled or reinitialized. It can be reset to zero by setting the PEDO_RST_STEP bit of the EMB_FUNC_SRC register to 1. After the counter resets, the PEDO_RST_STEP bit is automatically set back to 0.

The step detector functionality generates an interrupt every time a step is recognized. In the case of interspersed step sessions, 10 consecutive steps (debounce steps) have to be detected before the first interrupt generation in order to avoid false step detections (debounce functionality).

The number of debounce steps can be modified through the DEB_STEP[7:0] bits of the PEDO_DEB_STEPS_CONF register in the embedded advanced features registers: basically, it corresponds to the minimum number of steps to be detected before the first step counter increment. 1 LSB of this field corresponds to 1 step, the default value is 10 steps. The debounce functionality restarts after around 1 s of device inactivity.

An additional false-positive rejection (FPR) block can be enabled to perform the real-time recognition of the walking activity (including running) based on statistical data and to inhibit the step counter if no walking activity is detected. It can be activated as follows:

- Set the FP_REJECTION_EN bit of the PEDO_CMD_REG embedded advanced features register to 1.
- Set either the MLC_EN bit of the EMB_FUNC_EN_B or the MLC_BEFORE_FSM_EN bit of the EMB_FUNC_EN_A to 1.

In the LSM6DSV32X device, the FPR block can be customized by the user. In this case, the MLC must be programmed in order to use the first decision tree for the recognition of two classes: no walk (class with code 0x04) and walk (class with code 0x08). In detail, the step counter is inhibited if the following group of classes is detected by the MLC:

- Classes with code 0x4 through 0x7
- Classes with code 0xC through 0xE

STMicroelectronics provides the tools to generate specific pedometer configurations starting from a set of datalogs with a reference number of steps (Unico GUI on st.com).

The EMB_FUNC_SRC embedded functions register contains some read-only bits related to the pedometer function state.

Table 44. EMB_FUNC_SRC embedded functions register

- PEDO, RST, STEP: pedometer step counter reset. It can be set to 1 to reset the number of steps counted. It is automatically set back to 0 after the counter reset.
- STEP_DETECTED: step detector event status. It signals a step detection (after the debounce).
- STEP_COUNT_DELTA_IA: instead of generating an interrupt signal every time a step is recognized, it is possible to generate it if at least one step is detected within a certain time period, defined by setting a value different from 00h in the PEDO_SC_DELTAT_H and PEDO_SC_DELTAT_L embedded advanced features (page 1) registers. It is necessary to set the TIMESTAMP_EN bit of the FUNCTIONS_ENABLE register to 1 (to enable the timer). The time period is given by the concatenation of PEDO_SC_DELTAT_H and PEDO_SC_DELTAT_L and it is represented as a 16-bit unsigned value with a resolution of 5.6 ms. STEP_COUNT_DELTA_IA goes high (at the end of each time period) if at least one step is counted (after the debounce) within the programmed time period. If the time period is not programmed (PEDO_SC_DELTAT = 0), this bit is kept to 0.
- STEP_OVERFLOW: overflow signal that goes high when the step counter value reaches 2¹⁶.
- STEPCOUNTER_BIT_SET: step counter event status. It signals an increase in the step counter (after the debounce). If a timer period is programmed in the PEDO_SC_DELTAT_H and PEDO_SC_DELTAT_L embedded advanced features (page 1) registers, this bit is kept to 0.

The step detection interrupt signal can also be checked by reading the IS_STEP_DET bit of the EMB_FUNC_STATUS embedded functions register or the IS_STEP_DET bit of the EMB_FUNC_STATUS_MAINPAGE register.

The IS_STEP_DET bit can have different behaviors, as summarized in the table below, depending on the value of the PEDO_SC_DELTAT field (concatenation of PEDO_SC_DELTAT_H and PEDO_SC_DELTAT_L embedded advanced features registers) and the CARRY_COUNT_EN bit in the PEDO_CMD_REG embedded advanced features register.

Table 45. IS_STEP_DET configuration

The IS_STEP_DET interrupt signal can be driven to the INT1/INT2 interrupt pin by setting the INT1_STEP_DETECTOR/INT2_STEP_DETECTOR bit of the EMB_FUNC_INT1/EMB_FUNC_INT2 register to 1. In this case it is mandatory to also enable routing the embedded functions event to the INT1/INT2 interrupt pin by setting the INT1_EMB_FUNC/INT2_EMB_FUNC bit of the MD1_CFG/MD2_CFG register.

The behavior of the interrupt signal is pulsed by default. The duration of the pulse is equal to 1 / MAX_RATE seconds, where MAX_RATE denotes the maximum rate of the enabled embedded functions. If only the pedometer function is enabled, the duration of the pulse is then equal to 1 / 30 seconds. Latched mode can be enabled by setting the EMB_FUNC_LIR bit of the PAGE_RW embedded functions register to 1. In this case, the interrupt signal is reset by reading the IS_STEP_DET bit of the EMB_FUNC_STATUS embedded functions register or the IS_STEP_DET bit of the EMB_FUNC_STATUS_MAINPAGE register.

The step counter can be batched in FIFO (see [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0) for details).

A basic software routine that shows how to enable step counter detection is as follows:

- 1. Write 80h to FUNC_CFG_ACCESS // Enable access to embedded functions registers
- 2. Write 40h to PAGE_RW // Select write operation mode
- 3. Write 11h to PAGE SEL // Select page 1
- 4. Write 83h to PAGE_ADDR // Set embedded advanced features register to be written (PEDO_CMD_REG)

5. Write 04h to PAGE_VALUE // Enable false-positive rejection block link with pedometer

- (FP_REJECTION_EN = 1)
- 6. Write 00h to PAGE_RW // Write operation mode disabled
- 7. Write 08h to EMB_FUNC_EN_A // Enable pedometer
- 8. Write 10h to EMB_FUNC_EN_B // Enable false-positive rejection block (MLC_EN = 1)
- 9. Write 08h to EMB_FUNC_INT1 // Step detection interrupt driven to INT1 pin
- 10. Write 00h to FUNC_CFG_ACCESS // Disable access to embedded functions registers
- 11. Write 02h to MD1_CFG // Enable routing the embedded functions interrupt
- 12. Write 05h to CTRL8 \angle // FS_XL = ± 8 *g*
-
- 13. Write 04h to CTRL1 // Turn on the accelerometer (ODR_XL = 30 Hz)

6.2 Significant motion

The significant motion function generates an interrupt when a 'significant motion', that could be due to a change in user location, is detected. In the device, this function has been implemented in hardware using only the accelerometer.

The significant motion functionality can be used in location-based applications in order to receive a notification indicating when the user is changing location.

The significant motion function works at 30 Hz, so the accelerometer ODR must be set at a value of 30 Hz or higher. It generates an interrupt when the difference between the number of steps counted from its initialization/ reset is higher than 10 steps. After an interrupt generation, the algorithm internal state is reset.

In order to enable significant motion detection it is necessary to set the SIGN_MOTION_EN bit of the EMB_FUNC_EN_A embedded functions register to 1. The algorithm can be reinitialized by asserting the SIG_MOT_INIT bit of the EMB_FUNC_INIT_A embedded functions register.

Note: The significant motion feature automatically enables the internal step counter algorithm.

The significant motion interrupt signal can be driven to the INT1/INT2 interrupt pin by setting the INT1_SIG_MOT/ INT2_SIG_MOT bit of the EMB_FUNC_INT1/EMB_FUNC_INT2 register to 1. In this case it is mandatory to also enable routing the embedded functions event to the INT1/INT2 interrupt pin by setting the INT1_EMB_FUNC/ INT2_EMB_FUNC bit of the MD1_CFG/MD2_CFG register.

The significant motion interrupt signal can also be checked by reading the IS_SIGMOT bit of the EMB_FUNC_STATUS embedded functions register or the IS_SIGMOT bit of the EMB_FUNC_STATUS_MAINPAGE register.

The behavior of the significant motion interrupt signal is pulsed by default. The duration of the pulse is equal to 1 / MAX_RATE seconds, where MAX_RATE denotes the maximum rate of the enabled embedded functions. If only the significant motion function is enabled, the duration of the pulse is then equal to 1 / 30 seconds. Latched mode can be enabled by setting the EMB_FUNC_LIR bit of the PAGE_RW embedded functions register to 1. In this case, the interrupt signal is reset by reading the IS_SIGMOT bit of the EMB_FUNC_STATUS embedded functions register or the IS_SIGMOT bit of the EMB_FUNC_STATUS_MAINPAGE register.

A basic software routine that shows how to enable significant motion detection is as follows:

- 1. Write 80h to FUNC_CFG_ACCESS // Enable access to embedded functions registers
	-
- 2. Write 20h to EMB_FUNC_EN_A // Enable significant motion detection
- 3. Write 20h to EMB_FUNC_INT1 // Significant motion interrupt driven to INT1 pin
- 4. Write 80h to PAGE_RW // Enable latched mode for embedded functions
- 5. Write 00h to FUNC_CFG_ACCESS // Disable access to embedded functions registers
- 6. Write 02h to MD1_CFG // Enable routing the embedded functions interrupt
- 7. Write 04h to CTRL1 // Turn on the accelerometer
	- $\text{/} \text{/}$ ODR $\text{XL} = 30$ Hz

6.3 Relative tilt

The tilt function allows detecting when an activity change occurs (for example, when a phone is in a front pocket and the user goes from sitting to standing or from standing to sitting). In the device it has been implemented in hardware using only the accelerometer.

The tilt function works at 30 Hz, so the accelerometer ODR must be set at a value of 30 Hz or higher.

In order to enable the relative tilt detection function it is necessary to set the TILT_EN bit of the EMB_FUNC_EN_A embedded functions register to 1. The algorithm can be reinitialized by asserting the TILT_INIT bit of the EMB_FUNC_INIT_A embedded functions register.

If the device is configured for tilt event detection, an interrupt is generated when the device is tilted by an angle greater than 35 degrees from the start position. The start position is defined as the position of the device when the tilt detection is enabled/reinitialized or the position of the device when the last tilt interrupt was generated.

After this function is enabled or reinitialized, the tilt logic typically requires a 2-second settling time before being able to generate the first interrupt.

In the example shown in Figure 19. Tilt example tilt detection is enabled when the device orientation corresponds to "start position #0". The first interrupt is generated if the device is rotated by an angle greater than 35 degrees from the start position. After the first tilt detection interrupt is generated, the new start position (#1) corresponds to the position of the device when the previous interrupt was generated (final position #0), and the next interrupt signal is generated as soon as the device is tilted by an angle greater than 35 degrees, entering the blue zone surrounding the start position #1.

The tilt interrupt signal can be driven to the INT1/INT2 interrupt pin by setting the INT1_TILT/INT2_TILT bit of the EMB_FUNC_INT1/EMB_FUNC_INT2 register to 1. In this case it is mandatory to also enable routing the embedded functions event to the INT1/INT2 interrupt pin by setting the INT1_EMB_FUNC/INT2_EMB_FUNC bit of MD1_CFG/MD2_CFG register.

The tilt interrupt signal can also be checked by reading the IS_TILT bit of the EMB_FUNC_STATUS embedded functions register or the IS_TILT bit of the EMB_FUNC_STATUS_MAINPAGE register.

The behavior of the tilt interrupt signal is pulsed by default. The duration of the pulse is equal to 1 / MAX_RATE seconds, where MAX_RATE denotes the maximum rate of the enabled embedded functions. If only the tilt function is enabled, the duration of the pulse is then equal to 1 / 30 seconds. Latched mode can be enabled by setting the EMB_FUNC_LIR bit of the PAGE_RW embedded functions register to 1. In this case, the interrupt signal is reset by reading the IS_TILT bit of the EMB_FUNC_STATUS embedded functions register or the IS_TILT bit of the EMB_FUNC_STATUS_MAINPAGE register.

Hereafter a basic software routine that shows how to enable the tilt detection function:

-
- 2. Write 10h to EMB_FUNC_EN_A // Enable tilt detection
- 3. Write 10h to EMB_FUNC_INT1 // Tilt interrupt driven to INT1 pin
-
-
-
-
- 1. Write 80h to FUNC_CFG_ACCESS // Enable access to embedded functions registers
	-
	-
- 4. Write 80h to PAGE_RW // Enable latched mode for embedded functions
- 5. Write 00h to FUNC_CFG_ACCESS // Disable access to embedded functions registers
- 6. Write 02h to MD1_CFG // // Enable routing the embedded functions interrupt
- 7. Write 04h to CTRL1 // Turn on the accelerometer
	- $\text{/} \text{/}$ ODR $\text{XL} = 30 \text{ Hz}$

6.4 Timestamp

Together with sensor data the device can provide timestamp information.

To enable this functionality the TIMESTAMP_EN bit of the FUNCTIONS_ENABLE register has to be set to 1. The time step count is given by the concatenation of the TIMESTAMP3 & TIMESTAMP2 & TIMESTAMP1 & TIMESTAMP0 registers and is represented as a 32-bit unsigned number.

The nominal timestamp resolution is 21.75 μs. It is possible to get the actual timestamp resolution value through the FREQ_FINE_[7:0] bits of the INTERNAL_FREQ_FINE register, which contains the difference in percentage of the actual ODR (and timestamp rate) with respect to the nominal value.

 $t_{actual}[s] = \frac{1}{46080 \cdot (1 + 0.0013 \cdot FREQ_FINE)}$

Similarly, it is possible to get the actual output data rate by using the following formula:

$$
ODR_{actual}[Hz] = \frac{7680 \cdot (1 + 0.0013 \cdot \text{FREQ_FINE})}{ODR_{coeff}}
$$

where the ODR_{coeff} values are indicated in the table below.

Table 46. ODRcoeff values

If both the accelerometer and the gyroscope are in power-down mode, the timestamp counter does not work and the timestamp value is frozen at the last value.

When the maximum value 4294967295 LSB (equal to FFFFFFFh) is reached corresponding to approximately 26 hours, the counter is automatically reset to 00000000h and continues to count. The timer count can be reset to zero at any time by writing the reset value AAh in the TIMESTAMP2 register.

The TIMESTAMP_ENDCOUNT bit of the ALL_INT_SRC goes high 5.6 ms before the occurrence of a timestamp overrun condition. This flag is reset when the ALL_INT_SRC register is read. It is also possible to route this signal to the INT2 pin (65 μs duration pulse) by setting the INT2_TIMESTAMP bit of MD2_CFG to 1.

The timestamp can be batched in FIFO (see [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0) for details).

6.5 Sensor fusion functions

A dedicated sensor fusion block SFLP (sensor fusion low power) is available for generating the following data based on the accelerometer and gyroscope data processing:

- Game rotation vector, which provides a quaternion representing the attitude of the device
- Gravity vector, which provides a three-dimensional vector representing the direction of gravity
- Gyroscope bias, which provides a three-dimensional vector representing the gyroscope bias

The SFLP block is enabled by setting the SFLP_GAME_EN bit to 1 of the EMB_FUNC_EN_A embedded functions register.

The SFLP block can be reinitialized by setting the SFLP_GAME_INIT bit to 1 of the EMB_FUNC_INIT_A embedded functions register.

The SFLP block works at a configurable output data rate (which must be equal to or less than the selected output data rates of the accelerometer and gyroscope) through the SFLP_GAME_ODR_[2:0] field of the SFLP_ODR embedded functions register according to the following values:

- 000: 15 Hz
- 001: 30 Hz
- 010: 60 Hz
- 011: 120 Hz (default)
- 100: 240 Hz
- 101: 480 Hz

SFLP-generated data can be read from the FIFO only, see [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0) for details. The typical current consumption of the SFLP block is indicated in Table 47.

Table 47. SFLP current consumption (@ Vdd = 1.8 V, T = 25°C)

6.5.1 Gyroscope bias initial value setting

The SFLP embeds a gyroscope bias calibration routine, which is automatically executed when the device is steady. In applications where a steady condition for the gyroscope bias calibration cannot be guaranteed, a specific flow is needed to set a previously computed bias in the SFLP block. This procedure forces a reset of the SFLP algorithm and must be implemented as follows:

- 1. Convert gbias in HFP format in [rad/s] and divide by the k factor according to Table 48.
- 2. Save the current sensor configuration (CTRL1 and CTRL2 registers) and set the high-performance mode (if both the accelerometer and gyroscope are in power-down mode, turn the accelerometer on and wait for the first valid sample).
- 3. [optional] If the sensor hub is enabled, turn the I²C master off as indicated in Section 7.2.1: MASTER CONFIG (14h).
- 4. Disable the embedded functions (save the current values of the EMB_FUNC_EN_A and EMB_FUNC_EN_B registers and set them to 00h).
- 5. Wait until EMB_FUNC_ENDOP = 1.
- 6. Set the EMB_FUNC_DEBUG bit of the CTRL10 register to 1.
- 7. Set the SFLP_GAME_EN bit of the EMB_FUNC_EN_A register to 1.
- 8. Read the current accelerometer output data and write it in the sensor hub registers from SENSOR_HUB_1 to SENSOR_HUB_9. Each axis must be written to the sensor hub registers as a 24-bit signed number in two's complement, left-shifted by the current accelerometer full-scale setting (FS_XL = 00, do not shift; FS_XL = 01, shift by one; FS_XL = 10, shift by two; FS_XL = 11, shift by three). The sensor hub registers from SENSOR_HUB_10 to SENSOR_HUB_18 must be set to 00h.
- 9. Wait 30 us and then wait until EMB_FUNC_ENDOP = 1.
- 10. Write the gbias values computed at step #1 in the embedded advanced features page 0 registers from SFLP_GAME_GBIASX_L to SFLP_GAME_GBIASZ_H.
- 11. Reload the sensor configuration saved at steps #2 and #4 (do not set the SFLP_GAME_EN bit of the EMB_FUNC_EN_A register back to 0).
- 12. Set the EMB_FUNC_DEBUG bit of the CTRL10 register to 0.
- 13. [optional] If the sensor hub is enabled, turn the I²C master on.

Table 48. k factor

6.6 Embedded functions additional configurations and monitoring

The device provides the possibility to enable some additional configurations if needed through the EMB_FUNC_CFG register.

It allows three additional features:

- EMB_FUNC_IRQ_MASK_XL_SETTL bit can be set to 1 to enable the masking of the execution trigger of the embedded functions when accelerometer data are in the settling phase, in order to avoid the processing of accelerometer data during the settling phase.
- EMB_FUNC_IRQ_MASK_G_SETTL bit can be set to 1 to enable the masking of the execution trigger of the embedded functions when gyroscope data are in the settling phase, in order to avoid the processing of gyroscope data during the settling phase.
- EMB_FUNC_DISABLE bit can be set to 1 to stop the execution trigger of the embedded functions. When this bit is set back to 0, all the initialization procedures are forced and the execution trigger is again enabled.

The device provides the capability to monitor the execution of the embedded functions through the EMB_FUNC_EXEC_STATUS embedded functions register.

It contains the following information:

- Execution time overrun: this information is contained in the EMB_FUNC_EXEC_OVR bit. It is asserted if the execution time of the enabled embedded functions exceeds the maximum time, that is, a new set of sensor data to be used as input is generated before the end of the embedded functions execution.
- Execution ongoing: this information is contained in the EMB_FUNC_ENDOP bit. When this bit is set to 1, no embedded function is running, while when this bit is set to 0, embedded functions are running. This information can be routed to the INT2 pin by setting the INT2_EMB_FUNC_ENDOP bit of the INT2_CTRL register.

7 Mode 2 - sensor hub mode

The hardware flexibility of the LSM6DSV32X allows connecting the pins with different mode connections to external sensors to expand functionalities such as adding a sensor hub. When sensor hub mode (mode 2) is enabled, both the primary I²C/MIPI I3C[®]/SPI (3- and 4-wire) slave interface and the I²C master interface for the connection of external sensors are available. Mode 2 connection mode is described in detail in the following paragraphs.

7.1 Sensor hub mode description

In sensor hub mode (mode 2) up to 4 external sensors can be connected to the I²C master interface of the device. The sensor hub trigger signal can be synchronized with the accelerometer/gyroscope data-ready signal (up to 480 Hz). In this configuration, the sensor hub ODR can be configured through the SHUB_ODR_[2:0] bits of the SLV0_CONFIG register. Alternatively, an external signal connected to the INT2 pin can be used as the sensor hub trigger. In this second case, the maximum ODR supported for external sensors depends on the number of read / write operations that can be executed between two consecutive trigger signals.

On the sensor hub trigger signal, all the write and read I²C operations configured through the registers SLVx_ADD, SLVx_SUBADD, SLVx_CONFIG and DATAWRITE_SLV0 are performed sequentially from external sensor 0 to external sensor 3 (depending on the external sensors enabled through the AUX_SENS_ON[1:0] field in the MASTER_CONFIG register).

External sensor data can also be stored in FIFO (see [Section 9: First-in, first out \(FIFO\) buffer](#page-90-0) for details).

If both the accelerometer and the gyroscope are in power-down mode, the sensor hub does not work.

All external sensors have to be connected in parallel to the SDx/SCx pins of the device, as illustrated in Figure 20. External sensor connections in mode 2 for a single external sensor. External pull-up resistors and the external trigger signal connection are optional and depend on the configuration of the registers.

The SHUB PU EN bit of the IF CFG register can be used to enable or disable the internal pull-up on the I²C master line. When this bit is set to 0, the internal pull-up is disabled and the external pull-up resistors on the SDx/SCx pins are required, as shown in Figure 20. External sensor connections in mode 2. When this bit is set to 1, the internal pull-up is enabled (regardless of the configuration of the MASTER_ON bit) and the external pull-up resistors on the SDx/SCx pins are not required.

External trigger is optional

The sensor hub configuration registers and output registers are accessible when the bit SHUB_REG_ACCESS of the FUNC_CFG_ACCESS register is set to 1. After setting the SHUB_REG_ACCESS bit to 1, only sensor hub registers are available. In order to guarantee the correct register mapping for other operations, after the sensor hub configuration or output data reading, the SHUB_REG_ACCESS bit of the FUNC_CFG_ACCESS register must be set to 0.

The MASTER_CONFIG register has to be used for the configuration of the I²C master interface.

A set of registers SLVx_ADD, SLVx_SUBADD, SLVx_CONFIG is dedicated to the configuration of the 4 slave interfaces associated to the 4 connectable external sensors. An additional register, DATAWRITE_SLV0, is associated to slave #0 only. It has to be used to implement the write operations.

Finally, 18 registers (from SENSOR_HUB_1 to SENSOR_HUB_18) are available to store the data read from the external sensors.

7.2.1 MASTER_CONFIG (14h)

This register is used to configure the I²C master interface.

Table 49. MASTER_CONFIG register

- RST_MASTER_REGS bit is used to reset the I²C master interface, configuration and output registers. It must be manually asserted and de-asserted.
- WRITE_ONCE bit is used to limit the write operations on slave 0 to only one occurrence (avoiding to repeat the same write operation multiple times). If this bit is not asserted, a write operation is triggered at each ODR.

Note: The WRITE_ONCE bit must be set to 1 if slave 0 is used for read transactions.

- START_CONFIG bit selects the sensor hub trigger signal.
	- When this bit is set to 0, the accelerometer/gyroscope sensor has to be active (not in power-down mode) and the sensor hub trigger signal is the accelerometer/gyroscope data-ready signal, with a frequency defined by the SHUB_ODR_[2:0] bits of the SLV0_CONFIG register (up to 480 Hz).
	- When this bit is set to 1, at least one sensor between the accelerometer and the gyroscope has to be active and the sensor hub trigger signal is the INT2 pin. In fact, when both the MASTER ON bit and START_CONFIG bit are set to 1, the INT2 pin is configured as an input signal. In this case, the INT2 pin has to be connected to the data-ready pin of the external sensor [\(Figure 20. External sensor](#page-71-0) [connections in mode 2](#page-71-0)) in order to trigger the read/write operations on the external sensor registers. The sensor hub interrupt from INT2 polarity can be selected through the INT2_IN_LH bit of the CTRL4 register: if it is set to 0, the pin is active low, otherwise, it is active high.
- *Note: In case of external trigger signal usage (START_CONFIG=1), if the INT2 pin is connected to the dataready pin of the external sensor [\(Figure 20. External sensor connections in mode 2](#page-71-0)) and the latter is in power-down mode, then no data-ready signal can be generated by the external sensor. For this reason, the initial configuration of the external sensor's register has to be performed using the internal trigger signal (START_CONFIG=0). After the external sensor is activated and the data-ready signal is available, the external trigger signal can be used by switching the START_CONFIG bit to 1.*
	- PASS_THROUGH_MODE bit is used to enable/disable the I²C interface pass-through. When this bit is set to 1, the main I²C line (for example, connected to an external microcontroller) is short-circuited with the auxiliary one, in order to implement a direct access to the external sensor registers. See [Section 7.3: Sensor hub pass-through feature](#page-79-0) for details.

- MASTER_ON bit has to be set to 1 to enable the auxiliary I²C master of the device (sensor hub mode). In order to change the sensor hub configuration at runtime or when setting the accelerometer and gyroscope sensor in power-down mode, or when applying the software reset procedure, the I²C master must be disabled, followed by a 300 µs wait. The following procedure must be implemented:
	- 1. Turn off I^2C master by setting MASTER ON = 0.
	- 2. Wait 300 µs.
	- 3. Change the configuration of the sensor hub registers or set the accelerometer/gyroscope in powerdown mode or apply the software reset procedure.
- AUX_SENS_ON[1:0] bits have to be set accordingly to the number of slaves to be used. I²C transactions are performed sequentially from slave 0 to slave 3. The possible values are:
	- 00: one slave
	- 01: two slaves
	- 10: three slaves
	- 11: four slaves

7.2.2 STATUS_MASTER (22h)

The STATUS MASTER register, similarly to the other sensor hub configurations and output registers, can be read only after setting the SHUB_REG_ACCESS bit of the FUNC_CFG_ACCESS register to 1. The STATUS MASTER register is also mapped to the STATUS MASTER MAINPAGE register, which can be directly read without enabling access to the sensor hub registers.

Table 50. STATUS_MASTER / STATUS_MASTER_MAINPAGE register

- WR_ONCE_DONE bit is set to 1 after a write operation performed with the WRITE_ONCE bit configured to 1 in the MASTER_CONFIG register. This bit can be polled in order to check if the single write transaction has been completed.
- SLAVEx NACK bits are set to 1 if a "not acknowledge" event happens during the communication with the corresponding slave x.
- SENS HUB ENDOP bit reports the end of an I²C master transaction. It is set to 1 when the transaction is concluded; it is reset to 0 when the STATUS_MASTER / STATUS_MASTER_MAINPAGE register is read. When a sensor hub routine is completed, this bit automatically goes to 1 and the external sensor data are available to be read from the SENSOR_HUB_x registers (depending on the configuration of the SLVx ADD, SLVx SUBADD, SLVx CONFIG registers). Information about the status of the I²C master can be driven to the INT1 interrupt pin by setting the INT1_SHUB bit of the MD1_CFG register to 1. This signal goes high on a rising edge of the SENS_HUB_ENDOP signal and it is cleared only if the STATUS_MASTER / STATUS_MASTER_MAINPAGE register is read.

7.2.3 SLV0_ADD (15h), SLV0_SUBADD (16h), SLV0_CONFIG (17h)

The sensor hub registers used to configure the I²C slave interface associated to the first external sensor are described hereafter.

Table 51. SLV0_ADD register

- The slave0 add[6:0] bits are used to indicate the I²C slave address of the first external sensor.
- The rw_0 bit configures the read/write operation to be performed on the first external sensor (0: write operation; 1: read operation). The read/write operation is executed when the next sensor hub trigger event occurs.

Table 52. SLV0_SUBADD register

The slave0 reg[7:0] bits are used to indicate the address of the register of the first external sensor to be written (if the rw 0 bit of the SLV0_ADD register is set to 0) or the address of the first register to be read (if the rw_0 bit is set to 1).

Table 53. SLV0_CONFIG register

- The SHUB ODR [2:0] bits are used to configure the sensor hub output data rate when using an internal trigger (accelerometer/gyroscope data-ready signals). The sensor hub output data rate can be configured to six possible values, limited by the ODR of the accelerometer and gyroscope sensors:
	- 000: 1.875 Hz
	- 001: 15 Hz
	- 010: 30 Hz
	- 011: 60 Hz
	- 100: 120 Hz (default)
	- 101: 240 Hz
	- 110: 480 Hz

The maximum allowed value for the SHUB_ODR_[2:0] bits corresponds to the maximum ODR between the accelerometer and gyroscope sensors.

- The BATCH_EXT_SENS_0_EN bit is used to enable batching the external sensor data associated to slave0 in FIFO.
- The Slave0_numop[2:0] bits are dedicated to define the number of consecutive read operations to be performed on the first external sensor starting from the register address indicated in the SLV0_SUBADD register.

7.2.4 SLV1_ADD (18h), SLV1_SUBADD (19h), SLV1_CONFIG (1Ah)

The sensor hub registers used to configure the I²C slave interface associated to the second external sensor are described hereafter.

Table 54. SLV1_ADD register

- The slave1 add[6:0] bits are used to indicate the I²C slave address of the second external sensor.
- The r_1 bit enables/disables the read operation to be performed on the second external sensor (0: read operation disabled; 1: read operation enabled). The read operation is executed when the next sensor hub trigger event occurs.

The slave1 reg[7:0] bits are used to indicate the address of the register of the second external sensor to be read when the r¹ bit of SLV1 ADD register is set to 1.

Table 56. SLV1_CONFIG register

- The BATCH_EXT_SENS_1_EN bit is used to enable batching the external sensor data associated to slave1 in FIFO.
- The Slave1_numop[2:0] bits are dedicated to define the number of consecutive read operations to be performed on the second external sensor starting from the register address indicated in the SLV1_SUBADD register.

7.2.5 SLV2_ADD (1Bh), SLV2_SUBADD (1Ch), SLV2_CONFIG (1Dh)

The sensor hub registers used to configure the I²C slave interface associated to the third external sensor are described hereafter.

Table 57. SLV2_ADD register

- The slave2 add[6:0] bits are used to indicate the I²C slave address of the third external sensor.
- The r_2 bit enables/disables the read operation to be performed on the third external sensor (0: read operation disabled; 1: read operation enabled). The read operation is executed when the next sensor hub trigger event occurs.

The slave2_reg[7:0] bits are used to indicate the address of the register of the third external sensor to be read when the r_2 bit of the SLV2_ADD register is set to 1.

Table 59. SLV2_CONFIG register

- The BATCH_EXT_SENS_2_EN bit is used to enable batching the external sensor data associated to slave2 in FIFO.
- The Slave2_numop[2:0] bits are dedicated to define the number of consecutive read operations to be performed on the third external sensor starting from the register address indicated in the SLV2_SUBADD register.

7.2.6 SLV3_ADD (1Eh), SLV3_SUBADD (1Fh), SLV3_CONFIG (20h)

The sensor hub registers used to configure the I²C slave interface associated to the fourth external sensor are described hereafter.

- The slave3 add[6:0] bits are used to indicate the I²C slave address of the fourth external sensor.
- The r_3 bit enables/disables the read operation to be performed on the fourth external sensor (0: read operation disabled; 1: read operation enabled). The read operation is executed when the next sensor hub trigger event occurs.

• The slave3_reg[7:0] bits are used to indicate the address of the register of the fourth external sensor to be read when the r_3 bit of the SLV3_ADD register is set to 1.

Table 62. SLV3_CONFIG register

- The BATCH_EXT_SENS_3_EN bit is used to enable batching the external sensor data associated to slave3 in FIFO.
- The Slave3_numop[2:0] bits are dedicated to define the number of consecutive read operations to be performed on the fourth external sensor starting from the register address indicated in the SLV3_SUBADD register.

7.2.7 DATAWRITE_SLV0 (21h)

Table 63. DATAWRITE_SLV0 register

The Slave0_dataw[7:0] bits are dedicated, when the rw_0 bit of SLV0_ADD register is set to 0 (write operation), to indicate the data to be written to the first external sensor at the address specified in the SLV0_SUBADD register.

7.2.8 SENSOR_HUB_x registers

Once the auxiliary I²C master is enabled, for each of the external sensors it reads a number of registers equal to the value of the Slavex_numop ($x = 0, 1, 2, 3$) field, starting from the register address specified in the SLVx_SUBADD (x = 0, 1, 2, 3) register. The number of external sensors to be managed is specified in the AUX_SENS_ON[1:0] bits of the MASTER_CONFIG register.

Read data are consecutively stored (in the same order they are read) in the device registers starting from the SENSOR_HUB_1 register, as in the example in Figure 21. SENSOR_HUB_X allocation example; 18 registers, from SENSOR_HUB_1 to SENSOR_HUB_18, are available to store the data read from the external sensors.

Figure 21. SENSOR_HUB_X allocation example

SENSOR_HUB_18

Value of reg 44h

7.3 Sensor hub pass-through feature

The PASS_THROUGH_MODE bit of the MASTER_CONFIG register is used to enable/disable the I²C interface pass-through. When it is set to 1, the main I²C line (for example, connected to an external microcontroller) is short-circuited with the auxiliary one in order to implement a direct access to the external sensor registers. The pass-through feature for external device configuration can be used only if the I²C protocol is used on the primary interface. This feature can be used to configure the external sensors.

Figure 22. Pass-through feature

The following procedure can be implemented to enable the pass-through mode:

- 1. If the I²C master is enabled (MASTER ON = 1), turn it off (set the MASTER ON bit to 0) and wait 300 μs.
- 2. If the pull-up on the I²C master line is enabled, disable it (set the SHUB_PU_EN bit of the IF_CFG register to 0).
- 3. Enable the pass-through mode by setting the PASS_THROUGH_MODE bit to 1.

7.4 Sensor hub mode example

The configuration of the external sensors can be performed using the pass-through feature. This feature can be enabled by setting the PASS_THROUGH_MODE bit of the MASTER_CONFIG register to 1 and implements a direct access to the external sensor registers, allowing quick configuration.

The code provided below gives basic routines to configure a device in sensor hub mode. Three different snippets of code are provided here, in order to present how to easily perform a one-shot write or read operation, using slave 0, and how to set up slave 0 for continuously reading external sensor data.

The PASS_THROUGH_MODE bit is disabled in all these routines, in order to be as generic as possible.

One-shot read routine (using internal trigger) is described below. For simplicity, the routine uses the accelerometer configured at 120 Hz, with external pull-ups on the I²C auxiliary bus.

- 1. Write 40h to FUNC_CFG_ACCESS // Enable access to sensor hub registers 2. Write EXT_SENS_ADDR | 01h to SLV0_ADD // Configure external device address (EXT_SENS_ADDR) $//$ Enable read operation (rw_0 = 1) 3. Write REG to SLV0_SUBADD // Configure address (REG) of the register to be read 4. Write 81h to SLV0_CONFIG // Read one byte, SHUB_ODR = 120 Hz 5. Write 44h to MASTER_CONFIG // WRITE_ONCE is mandatory for read // I²C master enabled, using slave 0 only 6. Write 00h to FUNC_CFG_ACCESS // Disable access to sensor hub registers 7. Read OUTX H A register // Clear accelerometer data-ready XLDA 8. Poll STATUS_REG, until XLDA = 1 // Wait for sensor hub trigger
- 9. Poll STATUS_MASTER_MAINPAGE, until SENS_HUB_ENDOP = 1
- // Wait for sensor hub read transaction

The one-shot routine can be easily changed to setup the device for **continuous reading** of external sensor data:

After the execution of step 6, external sensor data are available to be read in sensor hub output registers. The **One-shot write routine** (using internal trigger) is described below. For simplicity, the routine uses the accelerometer configured at 120 Hz, with external pull-ups on the I²C auxiliary bus.

- 1. Write 40h to FUNC_CFG_ACCESS // Enable access to sensor hub registers
-
-
- 4. Write 80h to SLV0 CONFIG // SHUB_ODR = 120 Hz
-
-
- 7. Poll STATUS_MASTER, until WR_ONCE_DONE = 1
- 8. Write 00h to MASTER_CONFIG // I²C master disabled
- 9. Wait 300 µs
- 10. Write 00h to FUNC_CFG_ACCESS // Disable access to sensor hub registers
-
- 2. Write EXT_SENS_ADDR to SLV0_ADD // Configure external device address (EXT_SENS_ADDR)
	- // Enable write operation (rw_0 = 0)
- 3. Write REG to SLV0_SUBADD // Configure address (REG) of the register to be written
	-
- 5. Write VAL to DATAWRITE_SLV0 // Configure value (VAL) to be written in REG
- 6. Write 44h to MASTER_CONFIG // WRITE_ONCE enabled for single write
	- // I²C master enabled, using slave 0 only
	- // Wait for sensor hub write transaction
	-
	-

The following sequence configures the LIS2MDL external magnetometer sensor (refer to the datasheet for additional details) in continuous-conversion mode at 100 Hz (enabling temperature compensation, BDU and offset cancellation features) and reads the magnetometer output registers, saving their values in the SENSOR_HUB_1 to SENSOR_HUB_6 registers.

-
- 2. Perform **one-shot read** with SLV0_ADD = 3Dh SLV0_SUBADD = 4Fh
- 3. Perform **one-shot write** with SLV0_ADD = 3Ch SLV0_SUBADD = 60h DATAWRITE_SLV0 = 8Ch
- 4. Perform **one-shot write** with SLV0_ADD = 3Ch SLV0_SUBADD = 61h DATAWRITE_SLV0 = 02h
- 5. Perform **one-shot write** with $SLVO$ $ADD = 3Ch$ SLV0_SUBADD = 62h DATAWRITE_SLV0 = 10h
- 6. Set up **continuous read** with SLV0_ADD = 3Dh SLV0_SUBADD = 68h SLV0_CONFIG = 80h | 06h
- 1. Write 06h to CTRL1 // Turn on the accelerometer (for trigger signal) at 120 Hz
	- // Check LIS2MDL WHO_AM_I register
	- // LIS2MDL slave address is 3Ch and rw_0=1
	- // WHO_AM_I register address is 4Fh
	- // Write LIS2MDL register CFG_REG_A (60h) = 8Ch
	- // LIS2MDL slave address is 3Ch and rw_0=0
	- // Enable temperature compensation
	- // Enable magnetometer at 100 Hz ODR in continuous mode
	- // Write LIS2MDL register CFG_REG_B (61h) = 02h
	- // LIS2MDL slave address is 3Ch and rw_0=0
	- // Enable magnetometer offset-cancellation

// Write LIS2MDL register CFG_REG_B (62h) = 10h // LIS2MDL slave address is 3Ch and rw_0=0 // Enable magnetometer BDU

// LIS2MDL slave address is 3Ch and rw_0=1 // Magnetometer output registers start from 68h // Set up a continuous 6-byte read from I²C master interface

8 Mode 3 - OIS functionality

The LSM6DSV32X embeds a dedicated gyroscope and accelerometer DSP for OIS applications. Enabling the OIS functionality (mode 3) it is possible to access the device from multiple external devices. Both an I²C/SPI (3/4 wire)/MIPI I3C[®] slave interface and an auxiliary SPI (3/4-wire) slave interface are available for connecting external master devices. It can be used, for example, in optical image stabilization (OIS) applications to access the device from both the application processor and the camera module at the same time. The camera module can continuously get the sensor data at a high rate for its image stabilization algorithms.

There are two different ways in order to enable and configure OIS functionality, which can be selected by using the OIS_CTRL_FROM_UI bit in the FUNC_CFG_ACCESS register (to be configured from the primary interface):

- **Auxiliary SPI full control (OIS_CTRL_FROM_UI = 0)**: enabling and configuration done from the auxiliary SPI (see Section 8.1: Auxiliary SPI full control)
- **Primary interface full control (OIS_CTRL_FROM_UI = 1): enabling and configuration done from the** primary interface (see [Section 8.5: Primary interface full control\)](#page-89-0)

Note: The OIS_CTRL_FROM_UI bit is reset by the software reset procedure.

These two modes offer complete flexibility for the management of the OIS chain and provide the same functionalities and configuration options for both the primary interface and the SPI auxiliary interface.

8.1 Auxiliary SPI full control

The auxiliary SPI full control has been designed for the case where the camera module connected to the SPI auxiliary interface is completely independent from the application processor (connected to the device through the primary interface). The auxiliary SPI mode can be configured by accessing and setting the registers SPI2_INT_OIS, SPI2_CTRL1_OIS, SPI2_CTRL2_OIS and SPI2_CTRL3_OIS from the auxiliary SPI.

Setting the OIS_G_EN bit to 1 of the SPI2_CTRL1_OIS register enables the gyroscope OIS chain. When it is enabled, the gyroscope output values are available through the auxiliary SPI interface (3/4-wire) with the full scale selected through the FS_G_OIS_[2:0] bits of the SPI2_CTRL2_OIS register and ODR at 7680 Hz.

Setting the OIS_XL_EN bit to 1 of the SPI2_CTRL1_OIS register enables the accelerometer OIS chain. When it is enabled, the accelerometer output values are available through the auxiliary SPI interface (3/4-wire) with the full scale selected through the FS_XL_OIS_[1:0] bits of the SPI2_CTRL3_OIS register and ODR at 7680 Hz.

The function of the device pins after mode 3 is enabled is indicated in the following table.

Table 64. Mode 3 pin description

The external devices have to be connected to the LSM6DSV32X as illustrated in [Figure 23](#page-83-0)), if using the SPI 3 wire interface (SIM_OIS bit in SPI2_CTRL1_OIS = 1). The setup has to be changed accordingly when using the SPI 4-wire interface (connect the SDO_Aux pin too).

Figure 23. External controller connections in mode 3 (SPI 3-wire)

The gyroscope filtering chain is shown in the following figure. The digital low-pass filter LPF1 is dedicated to the OIS chain and it is possible to configure the bandwidth of the gyroscope OIS chain through the LPF1_G_OIS_BW_[1:0] bits of the SPI2_CTRL2_OIS register.

Note: The ±4000 dps gyroscope full-scale configuration is not compatible with mode 3. When ±4000 dps full-scale is selected, the gyroscope OIS chain must be disabled (the OIS_G_EN bit of the UI_CTRL1_OIS / SPI2_CTRL1_OIS registers must be set to 0).

Note: When the gyroscope UI chain is on (ODR_G > 0) in low-power mode, and the gyroscope OIS chain is intended to be turned on or off, it is mandatory to apply the following procedure:

1. Set OIS_CTRL_FROM_UI = 1 to take control of the OIS chain settings from the UI interface.

2a. When the gyroscope OIS chain is intended to be turned on: set ODR_G_EIS > 0 and set OIS_G_EN = 1 from the UI interface.

2b. When the gyroscope OIS chain is intended to be turned off: set OIS_G_EN = 0 and set ODR_G_EIS = 0 from the UI interface.

The accelerometer filtering chain is shown in the following figure. A digital low-pass filter LPF_OIS is dedicated to the OIS chain and it is possible to configure the bandwidth of the accelerometer OIS chain through the LPF_XL_OIS_BW_[2:0] bits of the SPI2_CTRL3_OIS register.

- *Note: When the accelerometer OIS chain is enabled, the accelerometer UI chain cannot be configured in normal mode.*
- *Note: If the accelerometer OIS chain is turned on (OIS_XL_EN = 1) while the accelerometer UI chain is on (ODR_XL > 0) in low-power mode, it is mandatory to wait at least 1 / ODR_XL time period before turning the accelerometer OIS off (OIS_XL_EN = 0).*
- *Note: If the accelerometer OIS chain is turned on (OIS_XL_EN = 1) while the accelerometer UI chain is on (ODR_XL > 0) in low-power mode and the accelerometer UI chain is intended to be configured in high-performance mode, it is mandatory to wait at least 1 / ODR_XL time period before setting the accelerometer UI chain in highperformance mode.*

Note: If the accelerometer OIS chain is on and the accelerometer UI chain is intended to be used in low-power, it is mandatory to apply the following procedure to turn the accelerometer UI chain on:

1. Turn the accelerometer UI on in high-performance mode at 7680 Hz data rate.

- *2. Wait at least 300 μs.*
- *3. Set the desired ODR_XL and the low-power mode.*

8.2 SPI2 registers

The primary interface is always available and the gyroscope output values can be read from registers OUTX_L_G through OUTZ_H_G (22h to 27h) with full scale and ODR selectable through the CTRL2 register. Similarly, the accelerometer output values can be read over the primary interface from registers OUTX_L_A through OUTZ H A (28h to 2Dh) with full scale and ODR selectable through the CTRL1 register. The accelerometer/ gyroscope data stored in FIFO can be accessed over the primary interface only.

The value of the bits of the SPI2_INT_OIS, SPI2_CTRL1_OIS, SPI2_CTRL2_OIS, SPI2_CTRL3_OIS registers can be modified over the auxiliary SPI interface only (these registers are read-only when accessed over the primary interface). These are the only registers that can be written over the auxiliary SPI interface. All the other read/write registers can be written over the primary interface only and can be only read by the auxiliary SPI. Furthermore, the primary interface can access the OIS control registers (UI_INT_OIS, UI_CTRL1_OIS, UI CTRL2 OIS, UI CTRL3 OIS) in read mode.

Reading OIS data from the auxiliary SPI is enabled only when the SPI2_READ_EN bit in the SPI2_CTRL1_OIS register is set to 1.

When the gyroscope OIS chain is enabled, the gyroscope output values can be read from registers SPI2_OUTX_L_G_OIS through SPI2_OUTZ_H_G_OIS (22h to 27h) over the auxiliary SPI interface. When new gyroscope data is available on the OIS chain, the GDA bit of the SPI2_STATUS_REG_OIS register is set to 1. It is reset when one of the high parts of the output data registers (23h, 25h, 27h) is read. The GYRO_SETTLING bit in the SPI2_STATUS_REG_OIS register is equal to 1 when the gyroscope OIS chain is in settling phase. The data read during this settling phase are not valid. It is recommended to check the status of this bit to understand when valid data are available.

When the accelerometer OIS chain is enabled, the accelerometer output values can also be read from registers SPI2_OUTX_L_A_OIS through SPI2_OUTZ_H_A_OIS (28h to 2Dh) over the auxiliary SPI interface. When new accelerometer data is available on the OIS chain, the XLDA bit of the SPI2_STATUS_REG_OIS register is set to 1. It is reset when one of the high parts of the output data registers (29h, 2Bh, 2Dh) is read.

In mode 3 the gyroscope OIS data and the accelerometer OIS data with ODR at 7680 Hz can also be read from the primary interface. If the gyroscope OIS is enabled and the bit G_EIS_ON_G_OIS_OUT_REG of the CTRL EIS register is set to 0, the gyroscope output values can be read over the primary interface from registers UI_OUTX_L_G_OIS_EIS through UI_OUTZ_H_G_OIS_EIS (2Eh to 33h). If the accelerometer OIS is enabled and the accelerometer dual-channel mode is disabled (XL_DualC_EN bit of the CTRL8 register is set to 0), the accelerometer output values can be read over the primary interface from registers UI_OUTX_L_A_OIS_DualC through UI_OUTZ_H_A_OIS_DualC (34h to 39h). When new gyroscope/accelerometer data is available on the OIS chain, the OIS_DRDY bit of the STATUS_REG register (accessible from primary interface) is set to 1.

Also the temperature sensor output data can be read over the auxiliary SPI interface by reading the SPI2_OUT_TEMP_L and SPI2_OUT_TEMP_H registers.

All the registers of the device can be read at the same time from both the external master devices.

8.2.1 SPI2_INT_OIS (6Fh)

Table 65. SPI2_INT_OIS register

- INT2_DRDY_OIS bit can be used to drive the DRDY signal of the OIS chain to the INT2 pin. The DRDY signal of the OIS chain is always pulsed; latched mode is not available.
- DRDY_MASK_OIS can be used to mask the interrupt signal routed to the INT2 pin until the OIS accelerometer/gyroscope filter settling ends. The OIS accelerometer and OIS gyroscope are independently masked.
- ST_OIS_CLAMPDIS bit can be used to enable/disable the OIS chain clamp in the gyroscope and accelerometer self-test. If the ST_OIS_CLAMPDIS bit is set to 1, once the gyroscope/accelerometer selftest functionality is enabled, the output values read from the OIS chain show the same variation observed while reading the data from the UI chain. If the ST_OIS_CLAMPDIS bit is set to 0, when the gyroscope/ accelerometer self-test functionality is enabled, the output values read from the OIS chain are always clamped to the value 8000h. For example, this feature allows the host device connected to the auxiliary interface to detect when the self-test functionality has been enabled from the UI side. By design, the valid range of the output values is between 8004h (-32764 LSB) and 7FFCh (+32764 LSB), and 8000h (-32768 LSB) is a special value which can be used from the auxiliary SPI in order detect that the self-test feature is enabled from the UI side.
- ST_G_OIS_[1:0] can be set in order to select the self-test on the gyroscope OIS chain (see [Section 12: Self-test](#page-123-0) for further details).
- ST_XL_OIS_11:0] can be set in order to select the self-test on the accelerometer OIS chain (see [Section 12: Self-test](#page-123-0) for further details).

8.2.2 SPI2_CTRL1_OIS (70h)

Table 66. SPI2_CTRL1_OIS register

- SIM_OIS bit has to be set to 1 in order to enable the 3-wire auxiliary SPI interface, otherwise the 4-wire auxiliary SPI interface is used.
- OIS XL EN bit allows enabling the accelerometer OIS chain.
- OIS G EN bit allows enabling the gyroscope OIS chain.
- SPI2_READ_EN bit has to be set to 1 to enable the auxiliary SPI to read OIS output data from registers SPI2_OUTX_L_G_OIS (22h) through SPI2_OUTZ_H_A_OIS (2Dh).

8.2.3 SPI2_CTRL2_OIS (71h)

Table 67. SPI2_CTRL2_OIS register

LPF1_G_OIS_BW_[1:0] _bits can be used to select the gyroscope digital LPF1 filter bandwidth. Table 68 shows the cutoff and phase delay values obtained with all the configurations.

• FS_G_OIS_[2:0] bits can be used to select the gyroscope OIS full scale (000: ±125 dps (default); 001: ±250 dps; 010: ±500 dps; 011: ±1000 dps; 100: ±2000 dps).

Table 68. LPF1 filter configuration

1. Settling time @ 99% of the final value

2. If the UI gyroscope is in power-down mode, wait 70 ms before discarding the samples indicated in this column.

8.2.4 SPI2_CTRL3_OIS (72h)

Table 69. SPI2_CTRL3_OIS register

LPF_XL_OIS_BW_[2:0] bits can be used to select the accelerometer digital LPF_OIS filter bandwidth. Table 70 shows the cutoff and phase delay values obtained with all configurations.

• FS_XL_OIS_[1:0] bits can be used to select the accelerometer OIS full scale (00: ±4 *g* (default); 01: ±8 *g*; 10: ±16 *g*; 11: ±32 *g*)

Table 70. LPF_OIS filter configuration

1. Settling time @ 99% of the final value

8.2.5 SPI2_STATUS_REG_OIS (1Eh)

Table 71. SPI2_STATUS_REG_OIS register

- GYRO_SETTLING bit is set to 1 during the initial settling phase of the gyroscope output. The gyroscope output data generated when this bit is equal to 1 have to be discarded.
- GDA bit is set to 1 when new gyroscope data is available in registers from SPI2_OUTX_L_G_OIS through SPI2_OUTZ_H_G_OIS (22h to 27h) on the OIS chain. It is reset when one of the high parts of the output data registers is read.
- XLDA bit is set to 1 when new accelerometer data is available in registers from SPI2_OUTX_L_A_OIS through SPI2_OUTZ_H_A_OIS (28h to 2Dh) on the OIS chain. It is reset when one of the high parts of the output data register is read.

8.3 Reading OIS gyroscope data over the auxiliary SPI

The procedure to be applied after device power-up to read the OIS gyroscope output data over the auxiliary SPI 4-wire interface is as follows:

8.4 Reading OIS accelerometer data over the auxiliary SPI

The procedure to be applied after device power-up to read the OIS accelerometer output data over the auxiliary SPI 4-wire interface is as follows:

8.5 Primary interface full control

In addition to the auxiliary SPI full control described in [Section 8.1: Auxiliary SPI full control,](#page-82-0) the LSM6DSV32X offers another way to manage the OIS data chain: the primary interface full control. It allows enabling the OIS chain and getting both UI and OIS data directly over the primary interface connected to the application processor. The primary interface full control is enabled by setting the OIS_CTRL_FROM_UI bit in the FUNC_CFG_ACCESS register to 1 from the primary interface. After this bit is set to 1, the OIS functionalities can be directly configured from the primary interface using the UI_INT_OIS, UI_CTRL1_OIS, UI_CTRL2_OIS, UI_CTRL3_OIS registers. The functionalities associated to the bits of these registers are the same as the ones implemented by the bits contained in the corresponding SPI2_xxx register (see [Table 3. SPI registers](#page-8-0)).

When the primary interface full control is enabled, OIS gyroscope data (at 7680 Hz rate) are available in registers from UI_OUTX_L_G_OIS_EIS through UI_OUTZ_H_G_OIS_EIS (2Eh to 33h), whereas OIS accelerometer data (at 7680 Hz rate) are available in registers from UI_OUTX_L_A_OIS_DualC through UI_OUTZ_H_A_OIS_DualC (34h to 39h). The UI_STATUS_REG_OIS register (at address 44h) acts as a status register for such data. In addition, by setting the SPI2 READ EN bit of the UI CTRL1 OIS register to 1, it is possible to read the OIS chain data also over the auxiliary SPI interface. The auxiliary SPI can also access the SPI2_INT_OIS, SPI2_CTRL1_OIS, SPI2_CTRL2_OIS, and SPI2_CTRL3_OIS registers in read-only mode.

8.6 UI / SPI2 shared registers

The LSM6DSV32X provides six registers, from UI_SPI2_SHARED_0 (65h) through UI_SPI2_SHARED_5 (6Ah), which are shared between the primary interface and the auxiliary SPI interface and can be used as the contact point between the primary and secondary interface host. These shared registers are accessible only by one interface at a time and access is managed through the UI_SHARED_REQ and UI_SHARED_ACK bits of the UI_HANDSHAKE_CTRL register and the SPI2_SHARED_REQ and SPI2_SHARED_ACK bits of SPI2_HANDSHAKE_CTRL register.

When one interface (UI is the primary, SPI2 is the secondary) wants to write to the shared registers, it must set the corresponding request bit to 1 (UI_SHARED_REQ for the primary interface, SPI2_SHARED_REQ for the secondary interface) and poll the corresponding ack bit (UI_SHARED_ACK for the primary interface, SPI2_SHARED_ACK for the secondary interface) until it goes to 1. When the corresponding ack bit is set to 1, the interface has write access to the shared registers. Once the interface has finished writing to the shared registers, it must set back to 0 the corresponding request bit and poll the corresponding ack bit until it goes back to 0. Both the interfaces always have read access to the shared registers.

9 First-in, first-out (FIFO) buffer

In order to limit intervention by the host processor and facilitate post-processing data for event recognition, the LSM6DSV32X embeds a 1.5 KB (up to 4.5 KB with the compression feature enabled) first-in, first-out buffer (FIFO).

The FIFO can be configured to store the following data:

- Gyroscope sensor data (either UI or EIS channel)
- Accelerometer sensor data (either channel)
- Timestamp data
- Temperature sensor data
- External sensor (connected to sensor hub interface) data
- Step counter (and associated timestamp) data
- SFLP game rotation vector, gravity vector, gyroscope bias
- Machine learning core filters, features, and results

Saving the data in FIFO is based on FIFO words. A FIFO word is composed of:

- Tag, 1 byte
- Data, 6 bytes

Data can be retrieved from the FIFO through six dedicated registers, from address 79h to 7Eh: FIFO_DATA_OUT_X_L, FIFO_DATA_OUT_X_H, FIFO_DATA_OUT_Y_L, FIFO_DATA_OUT_Y_H, FIFO_DATA_OUT_Z_L, FIFO_DATA_OUT_Z_H.

The reconstruction of a FIFO stream is a simple task thanks to the FIFO_TAG field of the FIFO_DATA_OUT_TAG register that allows recognizing the meaning of a word in FIFO. The applications have maximum flexibility in choosing the rate of batching for sensors with dedicated FIFO configurations.

Seven different FIFO operating modes can be chosen through the FIFO_MODE_[2:0] bits of the FIFO_CTRL4 register:

- Bypass mode
- **FIFO** mode
- Continuous mode
- Continuous-to-FIFO mode
- Bypass-to-continuous mode
- Bypass-to-FIFO mode
- ContinuousWTM-to-full mode

To monitor the FIFO status (full, overrun, number of samples stored, and so forth), two dedicated registers are available: FIFO_STATUS1 and FIFO_STATUS2.

Programmable FIFO threshold can be set in FIFO_CTRL1 using the WTM_[7:0] bits.

FIFO full, FIFO threshold, and FIFO overrun events can be enabled to generate dedicated interrupts on the two interrupt pins (INT1 and INT2) through the INT1_FIFO_FULL, INT1_FIFO_TH and INT1_FIFO_OVR bits of the INT1_CTRL register, and through the INT2_FIFO_FULL, INT2_FIFO_TH and INT2_FIFO_OVR bits of the INT2_CTRL register.

Finally, FIFO embeds a compression algorithm that the user can enable in order to have up to 4.5 KB data stored in FIFO and take advantage in terms of interface communication length for FIFO flushing and communication power consumption.

9.1 FIFO description and batched sensors

FIFO is divided into 256 words of 7 bytes each. A FIFO word contains one byte with TAG information and 6 bytes of data: the overall FIFO buffer dimension is equal to 1792 bytes and can contain 1536 bytes of data. The TAG byte contains the information indicating which data is stored in the FIFO data field and other useful information.

FIFO is runtime configurable: a meta-information tag can be enabled in order to notify the user if batched sensor configurations have changed.

Moreover, in order to increase its capability, the FIFO embeds a compression algorithm for accelerometer and gyroscope data (refer to [Section 9.10: FIFO compression](#page-114-0) for further details).

Batched sensors can be classified in three different categories:

- 1. Main sensors, which are physical sensors:
	- a. Accelerometer sensor (either channel)
	- b. Gyroscope sensor (either UI or EIS channel)
- 2. Auxiliary sensors, which contain information of the status of the device:
	- a. Timestamp sensor
	- b. Configuration-change sensor (CFG-Change)
	- c. Temperature sensor
- 3. Virtual sensors:
	- a. External sensors read from sensor hub interface
	- b. Step counter sensor
	- c. SFLP game rotation vector, gravity vector, and gyroscope bias
	- d. Machine learning core filters, features, and results

Data can be retrieved from the FIFO through six dedicated registers: FIFO_DATA_OUT_X_L, FIFO_DATA_OUT_X_H, FIFO_DATA_OUT_Y_L, FIFO_DATA_OUT_Y_H, FIFO_DATA_OUT_Z_L, FIFO_DATA_OUT_Z_H.

A write to FIFO can be triggered by the following different events:

- Internal data-ready signal (fastest sensor between the accelerometer and gyroscope)
- Sensor hub data-ready
- Step detection event
- Virtual sensor new data available

9.2 FIFO registers

The FIFO buffer is managed by:

- Six control registers: FIFO_CTRL1, FIFO_CTRL2, FIFO_CTRL3, FIFO_CTRL4, COUNTER_BDR_REG1, COUNTER_BDR_REG2
- Two status registers: FIFO_STATUS1 and FIFO_STATUS2
- Seven output registers (tag + data): FIFO_DATA_OUT_TAG, FIFO_DATA_OUT_X_L, FIFO_DATA_OUT_X_H, FIFO_DATA_OUT_Y_L, FIFO_DATA_OUT_Y_H, FIFO_DATA_OUT_Z_L, FIFO_DATA_OUT_Z_H
- Some additional bits to route FIFO events to the two interrupt lines: INT1_CNT_BDR, INT1_FIFO_FULL, INT1_FIFO_OVR, INT1_FIFO_TH bits of the INT1_CTRL register and INT2_CNT_BDR, INT2_FIFO_FULL, INT2_FIFO_OVR, INT2_FIFO_TH bits of the INT2_CTRL register
- Some additional bits for other features:
	- FIFO_COMPR_EN bit of the EMB_FUNC_EN_B embedded function register in order to enable the FIFO compression algorithm
	- STEP_COUNTER_FIFO_EN bit of the EMB_FUNC_FIFO_EN_A register in order to enable batching the step counter data in FIFO
	- MLC_FIFO_EN bit of the EMB_FUNC_FIFO_EN_A register in order to enable batching the machine learning core results in FIFO
	- MLC_FILTER_FEATURE_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register in order to enable batching the machine learning core filters and features in FIFO
	- SFLP_GBIAS_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register in order to enable batching the gyroscope bias data in FIFO (the SFLP embedded function must be enabled)
	- SFLP_GRAVITY_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register in order to enable batching the gravity vector data in FIFO (the SFLP embedded function must be enabled)
	- SFLP_GAME_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register in order to enable batching the game rotation vector data in FIFO (the SFLP embedded function must be enabled)
	- FIFO_COMPR_INIT bit of the EMB_FUNC_INIT_B embedded function register in order to request a reinitialization of the FIFO compression algorithm
	- BATCH_EXT_SENS_0_EN, BATCH_EXT_SENS_1_EN, BATCH_EXT_SENS_2_EN, BATCH_EXT_SENS_3_EN bits of the SLV0_CONFIG, SLV1_CONFIG, SLV2_CONFIG, SLV3_CONFIG sensor hub registers, which enable batching the related external sensor data in FIFO

9.2.1 FIFO_CTRL1

The FIFO_CTRL1 register contains the FIFO watermark threshold level. The value of 1 LSB of the FIFO threshold level is referred to as a FIFO word (7 bytes).

The FIFO watermark flag (FIFO_WTM_IA bit in the FIFO_STATUS2 register) rises when the number of bytes stored in the FIFO is equal to or higher than the watermark threshold level.

In order to limit the FIFO depth to the watermark level, the STOP_ON_WTM bit must be set to 1 in the FIFO_CTRL2 register.

Table 72. FIFO_CTRL1 register

9.2.2 FIFO_CTRL2

Table 73. FIFO_CTRL2 register

The FIFO_CTRL2 register contains the bit STOP_ON_WTM which allows limiting the FIFO depth to the watermark level.

The FIFO_CTRL2 register also contains the bits to manage the FIFO compression algorithm for the accelerometer and gyroscope sensors:

- FIFO_COMPR_RT_EN bit allows runtime enabling / disabling of the compression algorithm: if the bit is set to 1, the compression is enabled, otherwise it is disabled.
- UNCOMPR_RATE_[1:0] configures the compression algorithm to write noncompressed data at a specific rate. The following table summarizes possible configurations.

Table 74. Forced noncompressed data write configurations

Moreover, the FIFO_CTRL2 register contains the ODR_CHG_EN bit which can be set to 1 in order to enable the CFG-Change auxiliary sensor to be batched in FIFO (described in the next sections) and the XL_DualC_BATCH_FROM_FSM bit which, in combination with a specific FSM configuration, can be set to 1 in order to enable batching the accelerometer channel 2 in the FIFO buffer. In this case, it is necessary to enable the accelerometer dual-channel mode by setting the XL_DualC_EN bit of the CTRL8 register to 1 and to configure one FSM to actually enable the batch operation. Refer to the finite state machine application note available on www.st.com.

9.2.3 FIFO_CTRL3

Table 75. FIFO_CTRL3 register

The FIFO_CTRL3 register contains the fields to select the writing frequency in FIFO for accelerometer and gyroscope sensor data. The selected batch data rate must be equal to or lower than the output data rate configured through the ODR_XL and ODR_G fields of the CTRL1_XL and CTRL2_G registers. The following tables indicate all the selectable batch data rates.

Table 76. Accelerometer batch data rate

Table 77. Gyroscope batch data rate

9.2.4 FIFO_CTRL4

The FIFO_CTRL4 register contains the fields to select the decimation factor for timestamp batching in FIFO and the batch data rate for the temperature sensor.

The timestamp write rate is configured as the maximum batch data rate (BDR_MAX) divided by the decimation factor specified in the DEC_TS_BATCH_[1:0] field. BDR_MAX is the maximum batch data rate among the following batch data rates:

- Accelerometer batch data rate (BDR_XL)
- Accelerometer channel 2 batch data rate (equal to ODR XL), if batching the accelerometer channel 2 data in FIFO is enabled
- Gyroscope batch data rate (BDR_GY)
- Gyroscope EIS batch data rate (equal to ODR_G_EIS), if batching the gyroscope EIS channel data in FIFO is enabled
- Sensor hub batch data rate (BDR_SHUB)

The programmable decimation factors are indicated in the table below.

Table 78. Timestamp batch data rate

The temperature batch data rate is configurable through the ODR_T_BATCH_[1:0] field as shown in the table below.

Table 79. Temperature sensor batch data rate

The FIFO_CTRL4 register also contains the FIFO operating modes bits. FIFO operating modes are described in [Section 9.7: FIFO modes](#page-105-0). Moreover, the FIFO_CTRL4 register contains the G_EIS_FIFO_EN bit, which can be set to 1 in order to enable batching the gyroscope EIS channel data.

Table 80. FIFO_CTRL4 register

9.2.5 COUNTER_BDR_REG1

Since the FIFO might contain meta-information (that is, CFG-Change sensor) and accelerometer and gyroscope data might be compressed, the FIFO provides a way to synchronize the FIFO reading on the basis of the accelerometer or gyroscope actual number of samples stored in FIFO: the BDR counter.

The BDR counter can be configured through the COUNTER_BDR_REG1 and COUNTER_BDR_REG2 registers.

Table 81. COUNTER_BDR_REG1 register

The TRIG_COUNTER_BDR_[1:0] field selects the trigger for the BDR counter:

- 00: accelerometer sensor is selected as the trigger
- 01: gyroscope sensor (UI channel) is selected as the trigger
- 10 or 11: gyroscope sensor (EIS channel) is selected as the trigger

The user can select the threshold which generates the COUNTER_BDR_IA event in the FIFO_STATUS2 register. Once the internal BDR counter reaches the threshold, the COUNTER_BDR_IA bit is set to 1. The threshold is configurable through the CNT_BDR_TH_[9:0] bits. The upper part of the field is contained in register COUNTER_BDR_REG1. 1 LSB value of the CNT_BDR_TH threshold level is referred to as one accelerometer/ gyroscope sample (X, Y and Z data).

The BDR counter is automatically reset when the FIFO is empty.

9.2.6 COUNTER_BDR_REG2

The COUNTER_BDR_REG2 register contains the lower part of the BDR-counter threshold.

Table 82. COUNTER_BDR_REG2 register

9.2.7 FIFO_STATUS1

The FIFO_STATUS1 register, together with the FIFO_STATUS2 register, provides information about the number of samples stored in the FIFO. 1 LSB value of the DIFF_FIFO level is referred to as a FIFO word (7 bytes).

Table 83. FIFO_STATUS1 register

9.2.8 FIFO_STATUS2

The FIFO_STATUS2 register, together with the FIFO_STATUS1 register, provides information about the number of samples stored in the FIFO and about the current status (watermark, overrun, full, BDR counter) of the FIFO buffer.

Table 84. FIFO_STATUS2 register

- FIFO WTM IA represents the watermark status. This bit goes high when the number of FIFO words (7 bytes each) already stored in the FIFO is equal to or higher than the watermark threshold level. The watermark status signal can be driven to the two interrupt pins by setting the INT1_FIFO_TH bit of the INT1_CTRL register or the INT2_FIFO_TH bit of the INT2_CTRL register to 1.
- FIFO OVR IA goes high when the FIFO is completely filled and at least one sample has already been overwritten to store the new data. This signal can be driven to the two interrupt pins by setting the INT1_FIFO_OVR bit of the INT1_CTRL register or the INT2_FIFO_OVR bit of the INT2_CTRL register to 1.
- FIFO, FULL, IA goes high when the next set of data that is stored in FIFO makes the FIFO completely full (that is, DIFF FIFO $8 = 1$) or generate a FIFO overrun. This signal can be driven to the two interrupt pins by setting the INT1_FIFO_FULL bit of the INT1_CTRL register or the INT2_FIFO_FULL bit of the INT2_CTRL register to 1.
- COUNTER_BDR_IA represents the BDR-counter status. This bit goes high when the number of accelerometer or gyroscope batched samples (on the base of the selected sensor trigger) reaches the BDR-counter threshold level configured through the CNT_BDR_TH_[9:0] bits of the COUNTER_BDR_REG1 and COUNTER_BDR_REG2 registers. The COUNTER_BDR_IA bit is automatically reset when the FIFO_STATUS2 register is read. The BDR-counter status can be driven to the two interrupt pins by setting the INT1_CNT_BDR bit of the INT1_CTRL register or the INT2_CNT_BDR bit of the INT2_CTRL register to 1.
- FIFO OVR LATCHED, as FIFO_OVR_IA, goes high when the FIFO is completely filled and at least one sample has already been overwritten to store the new data. The difference between the two flags is that FIFO_OVR_LATCHED is reset when the FIFO_STATUS2 register is read, whereas the FIFO_OVR_IA is reset when at least one FIFO word is read. This allows detecting a FIFO overrun condition during reading data from FIFO.
- DIFF_FIFO_8 contains the upper part of the number of unread words stored in the FIFO. The lower part is represented by the DIFF_FIFO_[7:0] bits in FIFO_STATUS1. The value of the DIFF_FIFO_[8:0] field corresponds to the number of 7-byte words in the FIFO.

Register content is updated synchronously to the FIFO write and read operations.

Note: The BDU feature also acts on the FIFO_STATUS1 and FIFO_STATUS2 registers. When the BDU bit is set to 1, it is mandatory to read FIFO_STATUS1 first and then FIFO_STATUS2.

9.2.9 FIFO_DATA_OUT_TAG

By reading the FIFO_DATA_OUT_TAG register, it is possible to understand to which sensor the data of the current reading belongs and to check if data are consistent.

Table 85. FIFO_DATA_OUT_TAG register

TAG SENSOR [4:0] field identifies the sensors stored in the 6 data bytes [\(Table 86](#page-98-0)).

TAG CNT [1:0] field identifies the FIFO time slot (described in the next sections).

The table below contains all the possible values and associated type of sensor for the TAG_SENSOR_[4:0] field.

Table 86. TAG_SENSOR field and associated sensor

9.2.10 FIFO_DATA_OUT

Data can be retrieved from the FIFO through six dedicated registers, from address 79h to address 7Eh: FIFO_DATA_OUT_X_L, FIFO_DATA_OUT_X_H, FIFO_DATA_OUT_Y_L, FIFO_DATA_OUT_Y_H, FIFO_DATA_OUT_Z_L, FIFO_DATA_OUT_Z_H.

The FIFO output registers content depends on the sensor category and type, as described in the next section.

9.3 FIFO batched sensors

As previously described, batched sensors can be classified in three different categories:

- 1. Main sensors
- 2. Auxiliary sensors
- 3. Virtual sensors

In this section, all the details about each category are presented.

9.4 Main sensors

The main sensors are the physical sensors of the LSM6DSV32X device: accelerometer and gyroscope. The batch data rate can be configured through the BDR_XL_[3:0] and BDR_GY_[3:0] fields of the FIFO_CTRL3 register. The batch data rate must be equal to or lower than the related sensor output data rate configured through the ODR_XL_[3:0] and ODR_G_[3:0] fields of the CTRL1 and CTRL2 registers.

Batching the accelerometer channel 2 data can be enabled by setting the XL_DualC_BATCH_FROM_IF bit to 1 in the EMB_FUNC_CFG register; alternatively, it can be triggered by the FSM (bit XL_DualC_BATCH_FROM_FSM must be set to 1 and one FSM must be configured to actually enable the batch operation). In both cases, accelerometer channel 2 data are stored in FIFO according to the ODR_XL_[3:0] field of the CTRL1 register.

Batching the gyroscope EIS channel data is enabled by setting the G_EIS_FIFO_EN bit of the FIFO_CTRL4 register to 1. Gyroscope EIS channel data are stored in FIFO according to the ODR_G_EIS_[1:0] field of the CTRL_EIS register.

The main sensors define the FIFO time base. This means that each one of the other sensors can be associated to a time base slot defined by the main sensors. A batch event of the fastest main sensor also increments the TAG counter (TAG_CNT field of FIFO_DATA_OUT_TAG register). This counter is composed of two bits and its value is continuously incremented (from 00 to 11) to identify different time slots.

An example of a batch data rate event is shown in Figure 26. Main sensors and time slot definitions. The BDR_GY event and BDR_XL event identify the time in which the corresponding sensor data is written to the FIFO. The evolution of the TAG counter identifies different time slots and its frequency is equivalent to the maximum value between BDR_XL and BDR_GY, since accelerometer channel 2 and gyroscope EIS channel data are not batched in this example.

In the general case, the frequency of the TAG counter is equivalent to the maximum batch data rate of either the accelerometer or gyroscope (considering also the accelerometer channel 2 and gyroscope EIS channel), whichever is faster.

Figure 26. Main sensors and time slot definitions

The FIFO word format of the main sensors is presented in the table below, representing the device addresses from 78h to 7Eh.

Table 87. Main sensors output data format in FIFO

Time Slot frequency = max(BDR_GY, BDR_XL) = 120 Hz

9.5 Auxiliary sensors

Auxiliary sensors are considered as service sensors for the main sensors. Auxiliary sensors include the:

- Temperature sensor (ODR_T_BATCH_I1:0) bits of the FIFO_CTRL4 register must be configured properly).
- Timestamp sensor: it stores the timestamp corresponding to a FIFO time slot (the TIMESTAMP_EN bit of the FUNCTIONS_ENABLE register must be set to 1 and the DEC_TS_BATCH_[1:0] bits of the FIFO_CTRL4 register must be configured properly).
- CFG-Change sensor: it identifies a change in some configuration of the device (ODR_CHG_EN bit of the FIFO_CTRL2 register must be set to 1).

Auxiliary sensors cannot trigger a write in FIFO. Their registers are written when the first main sensor or the external sensor event occurs (even if they are configured at a higher batch data rate).

The temperature output data format in FIFO is presented in the following table.

Table 88. Temperature output data format in FIFO

The timestamp output data format in FIFO is presented in the following table.

Table 89. Timestamp output data format in FIFO

1. Internal signal that is set to 1 when the accelerometer channel 2 data is stored in FIFO. When enabling the accelerometer channel 2 batching through the XL_DualC_BATCH_FROM_IF bit, one sample of accelerometer channel 2 data might be written in FIFO before the corresponding write of the CFG-Change sensor. When disabling the accelerometer channel 2 *batching through the XL_DualC_BATCH_FROM_IF bit, the latest sample of accelerometer channel 2 data might not be written in FIFO before the corresponding write of the CFG-Change sensor.*

As shown in Table 89, timestamp data contain also some meta-information, which can be used to detect a BDR change if the CFG-Change sensor is not batched in FIFO: the batch data rate of both the main sensors and the sensor hub. BDR_SHUB cannot be configured through a dedicated register. It is the result of the configured sensor hub ODR through the SHUB_ODR_[2:0] bits of the SLV0_CONFIG sensor hub register and the effective trigger sensor output data rate (the fastest between th accelerometer or gyroscope if the internal trigger is used). For the complete description of BDR_SHUB, refer to the next section about virtual sensors. Moreover, the timestamp data contain also meta-information about the current batch state for the accelerometer channel 2 and gyroscope EIS channel.

CFG-Change identifies a runtime change in the output data rate, the batch data rate, or other configurations of the main or virtual sensors. When a supported runtime change is applied, this sensor is written at the first new main sensor or virtual sensor event followed by a timestamp sensor (also if the timestamp sensor is not batched).

This sensor can be used to correlate data from the sensors to the device timestamp without storing the timestamp each time. It could be used also to notify the user to discard data due to embedded filters settling or to other configuration changes (that is, switching mode, output data rate, and so forth).

CFG-Change output data format in FIFO is presented in the following table.

Table 90. CFG-Change output data format in FIFO

1. Internal signal that is set to 1 when the accelerometer channel 2 data is stored in FIFO. When enabling the accelerometer channel 2 batching through the XL_DualC_BATCH_FROM_IF bit, one sample of accelerometer channel 2 data might be written in FIFO before the corresponding write of the CFG-Change sensor. When disabling the accelerometer channel 2 *batching through the XL_DualC_BATCH_FROM_IF bit, the latest sample of accelerometer channel 2 data might not be written in FIFO before the corresponding write of the CFG-Change sensor.*

2. Internal signal that is set to 0 when the gyroscope finishes the startup phase (maximum startup time is 70 ms).

9.6 Virtual sensors

Virtual sensors are divided into the following categories:

- 1. External sensors, read from the sensor hub interface
- 2. Step counter sensor
- 3. SFLP-generated sensors
- 4. MLC-generated sensors

9.6.1 External sensors and NACK sensor

Data of up to four external sensors read from the sensor hub (for a maximum of 18 bytes) can be stored in FIFO. They are continuous virtual sensors with the batch data rate (BDR_SHUB) corresponding to the current value of the SHUB_ODR_[2:0] field in the SLV0_CONFIG register, if an internal trigger is used (sensor hub read triggered by the accelerometer or gyroscope data-ready signal). This value is limited by the effective trigger sensor output data rate (the fastest between the accelerometer or gyroscope). If external sensors are not batched or an external trigger is used, BDR_SHUB is set to 0. The following table shows the possible values of the BDR_SHUB field.

Table 91. BDR_SHUB

1. This value can be obtained by selecting SHUB_ODR_[2:0] different from 000 and using an internal trigger (accelerometer or gyroscope) with ODR equal to 7.5 Hz.

As main sensors, external sensors define the FIFO time base and they can trigger the writing of auxiliary sensors in FIFO (only if they are batched and an external trigger is not used).

It is possible to selectively enable batching the data of the different external sensors using the BATCH_EXT_SENS_0_EN, BATCH_EXT_SENS_1_EN, BATCH_EXT_SENS_2_EN, BATCH_EXT_SENS_3_EN bits of the SLV0_CONFIG, SLV1_CONFIG, SLV2_CONFIG, SLV3_CONFIG sensor hub registers.

Each external sensor has a dedicated TAG value and 6 bytes reserved for data. External sensors are written in FIFO in the same order of the sensor hub output registers and if the number of bytes read from an external sensor is less than 6 bytes, then free bytes are filled with zeros.

If the communication with one external sensor batched in FIFO fails, the sensor hub writes a NACK sensor instead of the corresponding sensor data in FIFO. A NACK sensor contains the index (numbered from 0 to 3) of the failing slave and has the following output data format.

Table 92. Nack sensor output data format in FIFO

9.6.2 Step counter sensor

Step counter data, with associated timestamp, can be stored in FIFO. It is not a continuous rate sensor: the step detection event triggers its writing in FIFO.

In order to enable the step counter sensor in FIFO, the user should:

- 1. Enable the step counter sensor (set the PEDO_EN bit to 1 in the EMB_FUNC_EN_A embedded functions register)
- 2. Enable batching step counter data (set the STEP_COUNTER_FIFO_EN bit to 1 in the EMB_FUNC_FIFO_EN_A embedded functions register)

The format of the step counter data read from FIFO is shown in the table below.

Table 93. Step counter output data format in FIFO

9.6.3 SFLP-generated sensors

A dedicated sensor fusion block (SFLP) is available for generating the following virtual sensors based on processing the accelerometer and gyroscope data:

- Game rotation vector, which provides a quaternion representing the attitude of the device
- Gravity vector, which provides a three dimensional vector representing the direction of gravity
- Gyroscope bias, which provides a three dimensional vector representing the gyroscope bias

SFLP-generated sensors are read only from FIFO and they are selectively enabled:

- Game rotation vector is batched by setting the SFLP_GAME_FIFO_EN bit of the EMB_FUNC_FIFO_EN_A register to 1.
- Gravity vector is batched by setting the SFLP_GRAVITY_FIFO_EN bit of EMB_FUNC_FIFO_EN_A register to 1.
- Gyroscope bias is batched by setting the SFLP_GBIAS_FIFO_EN bit of the EMB_FUNC_FIFO_EN_A register to 1.

If batching in FIFO is enabled, the SFLP-generated sensors are stored in FIFO according to the SFLP output data rate.

The format for the SFLP-generated sensors in FIFO is listed below:

- Game rotation vector: X, Y, and Z axes (vector part of the quaternion) are stored in half-precision floatingpoint format, where w (scalar part of the quaternion) is computed in software after reading the data from the FIFO, since the game rotation vector is a unit quaternion.
- Gravity vector: X, Y, and Z axes are stored as 16-bit two's complement number with a sensitivity of 0.061 m*g* / LSB.
- Gyroscope bias: X, Y, and Z axes are stored as 16-bit two's complement number with a sensitivity of 4.375 mdps / LSB.

9.6.4 MLC-generated sensors

The following machine learning core (MLC-generated) virtual sensors can be stored in FIFO:

- **Results**
- **Filters**
- Features, including windowed and recursive features

In order to store MLC-generated sensors in FIFO, the MLC block must be enabled by setting either the MLC_BEFORE_FSM_EN bit of the EMB_FUNC_EN_A register or the MLC_EN bit of the EMB_FUNC_EN_B register.

Batching MLC results is enabled by setting the MLC_FIFO_EN bit of the EMB_FUNC_FIFO_EN_A register to 1.

An MLC result contains the information of the corresponding MLCx_SRC register and it is stored in FIFO when a change in the corresponding MLCx_SRC occurs.

Batching MLC filters is selectively enabled using one of the tools for configuring the MLC provided by STMicroelectronics. In addition, the MLC_FILTER_FEATURE_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register must be set to 1 to globally enable storing MLC filters or features in FIFO.

MLC filters are stored in FIFO at a rate equivalent to the MLC output data rate (MLC_ODR bits). If the filter is applied to the X, Y, Z axes of the desired sensor, one word is stored in FIFO for every axis. If the filter is applied to the norm of the desired sensor, one word is stored in FIFO.

Batching MLC features is selectively enabled using one of the tools for configuring the MLC provided by STMicroelectronics. In addition, the MLC_FILTER_FEATURE_FIFO_EN bit of the EMB_FUNC_FIFO_EN_B register must be set to 1 to globally enable storing MLC filters or features in FIFO.

MLC-windowed features are stored in FIFO at the end of every window.

MLC recursive features (like MLC filters) are stored in FIFO at a rate equivalent to the MLC output data rate (MLC_ODR).

The format of MLC results, features, and filters in FIFO is indicated in the following tables.

Table 94. MLC results in FIFO

1. MLCx_SRC registers are indexed from 0 to 3 (for example, MLC1_SRC is indexed as 0).

Table 95. MLC filters or features in FIFO

1. This value is represented as a half-precision floating-point number.

2. Filter and feature identifiers are indicated in the configuration file generated by STMicroelectronics tools for configuring the MLC.

9.7 FIFO modes

The LSM6DSV32X FIFO buffer can be configured to operate in seven different modes, selectable through the FIFO_MODE_[2:0] field of the FIFO_CTRL4 register. The available configurations ensure a high level of flexibility and extend the number of functions usable in application development.

Bypass, FIFO, continuous, continuous-to-FIFO, bypass-to-continuous, bypass-to-FIFO, and continuousWTM-tofull modes are described in the following paragraphs.

9.7.1 Bypass mode

When bypass mode is enabled, the FIFO is not used, the buffer content is cleared, and it remains empty until another mode is selected. Bypass mode is selected when the FIFO_MODE_[2:0] bits are set to 000. Bypass mode must be used in order to stop and reset the FIFO buffer when a different mode is intended to be used. Note that by placing the FIFO buffer into bypass mode, the whole buffer content is cleared.

9.7.2 FIFO mode

In FIFO mode, the buffer continues filling until it becomes full. Then it stops collecting data and the FIFO content remains unchanged until a different mode is selected.

Follow these steps for FIFO mode configuration:

- 1. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).
- 2. Set the FIFO_MODE_[2:0] bits in the FIFO_CTRL4 register to 001 to enable FIFO mode.

When this mode is selected, the FIFO starts collecting data. The FIFO_STATUS1 and FIFO_STATUS2 registers are updated according to the number of samples stored.

When the FIFO is full, the DIFF_FIFO_8 bit of the FIFO_STATUS2 register is set to 1 and no more data are stored in the FIFO buffer. Data can be retrieved by reading all the FIFO_DATA_OUT (from 78h to 7Eh) registers for the number of times specified by the DIFF_FIFO_[8:0] bits of the FIFO_STATUS1 and FIFO_STATUS2 registers.

Using the FIFO WTM. IA bit of the FIFO STATUS2 register, data can also be retrieved when a threshold level (WTM_[7:0] in the FIFO_CTRL1 register) is reached if the application requires a lower number of samples in the FIFO.

If the STOP_ON_WTM bit of the FIFO_CTRL2 register is set to 1, the FIFO size is limited to the value of the WTM_[7:0] bits in the FIFO_CTRL1 register. In this case, the FIFO_FULL_IA bit of the FIFO_STATUS2 register is set high when the number of samples in FIFO reaches or exceeds the WTM [7:0] value on the next FIFO write operation.

Communication speed is not very important in FIFO mode because the data collection is stopped and there is no risk of overwriting data already acquired. Before restarting the FIFO mode, it is necessary to set to bypass mode first in order to completely clear the FIFO content.

Figure 27. FIFO mode (STOP ON WTM = 0) shows an example of FIFO mode usage; the data from just one sensor are stored in the FIFO. In these conditions, the number of samples that can be stored in the FIFO buffer is 256 (with compression algorithm disabled). The FIFO_FULL_IA bit of the FIFO_STATUS2 register goes high just after the level labeled as 254 to notify that the FIFO buffer will be completely filled at the next FIFO write operation. After the FIFO is full (FIFO_DIFF_8 = 1), the data collection stops.

Figure 27. FIFO mode (STOP_ON_WTM = 0)

9.7.3 Continuous mode

In continuous mode, the FIFO continues filling. When the buffer is full, the FIFO index restarts from the beginning, and older data are replaced by the new data. The oldest values continue to be overwritten until a read operation frees FIFO slots. The host processor reading speed is important in order to free slots faster than new data is made available. To stop this configuration, bypass mode must be selected.

Follow these steps for continuous mode configuration (if the accelerometer/gyroscope data-ready is used as the FIFO trigger):

- 1. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).
- 2. Set the FIFO_MODE_[2:0] field in the FIFO_CTRL4 register to 110 to enable FIFO mode.

When this mode is selected, the FIFO collects data continuously. The FIFO STATUS1 and FIFO STATUS2 registers are updated according to the number of samples stored. When the next FIFO write operation makes the FIFO completely full or generates a FIFO overrun, the FIFO_FULL_IA bit of the FIFO_STATUS2 register goes to 1. The FIFO_OVR_ IA and FIFO_OVR_LATCHED bits in the FIFO_STATUS2 register indicates when at least one FIFO word has been overwritten to store the new data. Data can be retrieved after the FIFO_FULL_IA event by reading the FIFO_DATA_OUT (from 78h to 7Eh) registers for the number of times specified by the DIFF_FIFO_[8:0] bits in the FIFO_STATUS1 and FIFO_STATUS2 registers. Using the FIFO_WTM_IA bit of the FIFO_STATUS2 register, data can also be retrieved when a threshold level (WTM_[7:0] in the FIFO_CTRL1 register) is reached. If the STOP_ON_WTM bit of the FIFO_CTRL2 register is set to 1, the FIFO size is limited to the value of the WTM_[7:0] bits in the FIFO_CTRL1 register. In this case, the FIFO_FULL_IA bit of the FIFO_STATUS2 register goes high when the number of samples in FIFO reaches or overcomes the WTM_[7:0] value at the next FIFO write operation.

Figure 28. Continuous mode shows an example of the continuous mode usage. In the example, data from just one sensor are stored in the FIFO and the FIFO samples are read on the FIFO_FULL_IA event and faster than 1 * ODR so that no data is lost. In these conditions, the number of samples stored is 255.

Figure 28. Continuous mode

9.7.4 Continuous-to-FIFO mode

This mode is a combination of the continuous and FIFO modes previously described. In continuous-to-FIFO mode, the FIFO buffer starts operating in continuous mode and switches to FIFO mode when an event condition occurs.

The event condition can be one of the following:

- Single tap: event detection has to be configured and the INT2_SINGLE_TAP bit of the MD2_CFG register has to be set to 1.
- Double tap: event detection has to be configured and the INT2_DOUBLE_TAP bit of the MD2_CFG register has to be set to 1.
- Free-fall: event detection has to be configured and the INT2_FF bit of the MD2_CFG register has to be set to 1.
- Wake-up: event detection has to be configured and the INT2_WU bit of the MD2_CFG register has to be set to 1.
- 6D: event detection has to be configured and the INT2_6D bit of the MD2_CFG register has to be set to 1.

Continuous-to-FIFO mode is sensitive to the edge of the interrupt signal. At the first interrupt event, FIFO changes from continuous mode to FIFO mode and maintains it until bypass mode is set.

Figure 29. Continuous-to-FIFO mode

Follow these steps for continuous-to-FIFO mode configuration (if the accelerometer/gyroscope data-ready is used as the FIFO trigger):

- 1. Configure one of the events as previously described.
- 2. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).
- 3. Set the FIFO_MODE_[2:0] bits in the FIFO_CTRL4 register to 011 to enable FIFO continuous-to-FIFO mode.

In continuous-to-FIFO mode the FIFO buffer continues filling. When the FIFO is full or overrun at the next FIFO write operation, the FIFO_FULL_IA bit goes high.

If the STOP_ON_WTM bit of the FIFO_CTRL2 register is set to 1, the FIFO size is limited to the value of the WTM_[7:0] bits in the FIFO_CTRL1 register. In this case, the FIFO_FULL_IA bit of the FIFO_STATUS2 register goes high when the number of samples in FIFO reaches or exceeds the WTM_[7:0] value at the next FIFO write operation.

When the trigger event occurs, two different cases can be observed:

- 1. If the FIFO buffer is already full, it stops collecting data at the first sample after the event trigger. The FIFO content is composed of the samples collected before the event.
- 2. If FIFO buffer is not full yet, it continues filling until it becomes full and then it stops collecting data.

Continuous-to-FIFO can be used in order to analyze the history of the samples which have generated an interrupt. The standard operation is to read the FIFO content when the FIFO mode is triggered and the FIFO buffer is full and stopped.

9.7.5 Bypass-to-continuous mode

This mode is a combination of the bypass and continuous modes previously described. In bypass-to-continuous mode, the FIFO buffer starts operating in bypass mode and switches to continuous mode when an event condition occurs.

The event condition can be one of the following:

- Single tap: event detection has to be configured and the INT2_SINGLE_TAP bit of the MD2_CFG register has to be set to 1.
- Double tap: event detection has to be configured and the INT2_DOUBLE_TAP bit of the MD2_CFG register has to be set to 1.
- Free-fall: event detection has to be configured and the INT2_FF bit of the MD2_CFG register has to be set to 1.
- Wake-up: event detection has to be configured and the INT2_WU bit of the MD2_CFG register has to be set to 1.
- 6D: event detection has to be configured and the INT2_6D bit of the MD2_CFG register has to be set to 1.

Bypass-to-continuous mode is sensitive to the edge of the interrupt signal. At the first interrupt event, FIFO changes from bypass mode to continuous mode and maintains it until bypass mode is set.

Follow these steps for bypass-to-continuous mode configuration (if the accelerometer / gyroscope data-ready is used as the FIFO trigger):

- 1. Configure one of the events as previously described.
- 2. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).
- 3. Set the FIFO_MODE_[2:0] bits in the FIFO_CTRL4 register to 100 to enable FIFO bypass-to-continuous mode.

Once the trigger condition appears and the buffer switches to continuous mode, the FIFO buffer continues filling. When the next stored set of data makes the FIFO full or overrun, the FIFO_FULL_IA bit is set high.

Bypass-to-continuous can be used in order to start the acquisition when the configured interrupt is generated.

Figure 30. Bypass-to-continuous mode

9.7.6 Bypass-to-FIFO mode

This mode is a combination of the bypass and FIFO modes previously described. In bypass-to-FIFO mode, the FIFO buffer starts operating in bypass mode and switches to FIFO mode when an event condition occurs. The event condition can be one of the following:

- Single tap: event detection has to be configured and the INT2_SINGLE_TAP bit of the MD2_CFG register has to be set to 1.
- Double tap: event detection has to be configured and the INT2_DOUBLE_TAP bit of the MD2_CFG register has to be set to 1.
- Free-fall: event detection has to be configured and the INT2_FF bit of the MD2_CFG register has to be set to 1.
- Wake-up: event detection has to be configured and the INT2_WU bit of the MD2_CFG register has to be set to 1.
- 6D: event detection has to be configured and the INT2_6D bit of the MD2_CFG register has to be set to 1.

Bypass-to-FIFO mode is sensitive to the edge of the interrupt signal. At the first interrupt event, FIFO changes from bypass mode to FIFO mode and maintains it until ypass mode is set.

Follow these steps for ypass-to-FIFO mode configuration (if the accelerometer / gyroscope data-ready is used as the FIFO trigger):

- 1. Configure one of the events as previously described.
- 2. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).

3. Set the FIFO_MODE_[2:0] bits in the FIFO_CTRL4 register to 111 to enable FIFO bypass-to-FIFO mode. Once the trigger condition appears and the buffer switches to FIFO mode, the FIFO buffer starts filling. When the next stored set of data makes the FIFO full or overrun, the FIFO_FULL_IA bit is set high and the FIFO stops.

Bypass-to-FIFO can be used in order to analyze the history of the samples which have generated an interrupt.

Figure 31. Bypass-to-FIFO mode

9.7.7 ContinuousWTM-to-full mode

This mode is similar to continuous-to-FIFO mode previously described, with the following additional behaviors:

- When in continuous mode, the FIFO size is automatically limited according to the selected FIFO threshold level (WTM_[7:0] field of the FIFO_CTRL1 register), and for this reason it is referred to as "continuousWTM" mode. When in this mode, the FIFO full event is internally masked.
- When in FIFO mode, the FIFO size is no longer limited to the selected FIFO threshold level, and for this reason it is referred to as "full" mode. When in this mode, the FIFO full event is no longer internally masked.

In continuousWTM-to-full mode, the FIFO buffer starts operating in continuousWTM mode and switches to full mode when an event condition occurs.

The event condition can be one of the following:

- Single tap: event detection has to be configured and the INT2_SINGLE_TAP bit of the MD2_CFG register has to be set to 1.
- Double tap: event detection has to be configured and the INT2_DOUBLE_TAP bit of the MD2_CFG register has to be set to 1.
- Free-fall: event detection has to be configured and the INT2_FF bit of the MD2_CFG register has to be set to 1.
- Wake-up: event detection has to be configured and the INT2_WU bit of the MD2_CFG register has to be set to 1.
- 6D: event detection has to be configured and the INT2_6D bit of the MD2_CFG register has to be set to 1.

ContinuousWTM-to-full mode is sensitive to the edge of the interrupt signal. At the first interrupt event, FIFO changes from continousWTM mode to full mode and maintains it until bypass mode is set.

Follow these steps for continuousWTM-to-full mode configuration (if the accelerometer/gyroscope data-ready is used as the FIFO trigger):

- 1. Configure one of the events as previously described.
- 2. Enable the sensor data to be stored in FIFO with the corresponding batch data rate (if configurable).
- 3. Set the FIFO_MODE_[2:0] field in the FIFO_CTRL4 register to 010 to enable FIFO continuousWTM-to-full mode.

In continuousWTM-to-full mode the FIFO buffer continues filling. When the FIFO is full or overrun at the next FIFO write operation (as indicated above, the FIFO size is automatically limited to the value of the WTM_[7:0] field in the FIFO_CTRL1 register), the FIFO_FULL_IA bit does not go high, since it is internally masked. When the trigger event occurs, the FIFO buffer size is no longer limited to the value of the WTM_[7:0] field in the FIFO_CTRL1 register and it continues filling until it becomes full and then it stops collecting data.

ContinuousWTM-to-full mode can be used in order to analyze the history of both the samples which have generated an interrupt and the samples right after the interrupt generation. The standard operation is to read the FIFO content when the FIFO mode is triggered and the FIFO buffer is full and stopped.

Figure 32. ContinuousWTM-to-full mode

9.8 Retrieving data from the FIFO

When FIFO is enabled and the mode is different from bypass, reading the FIFO output registers return the oldest FIFO sample set. Whenever these registers are read, their content is moved to the SPI/I²C/MIPI I3CSM output buffer.

FIFO slots are ideally shifted up one level in order to release room for a new sample, and the FIFO output registers load the current oldest value stored in the FIFO buffer.

One way to retrieve data from the FIFO is the following:

- 1. Read the FIFO_STATUS1 and FIFO_STATUS2 registers to check how many words are stored in the FIFO. This information is contained in the DIFF_FIFO_[8:0] field.
- 2. For each word in FIFO, read the FIFO word (tag and output data) and interpret it on the basis of the FIFO tag.
- 3. Go to step 1.

The entire FIFO content is retrieved by performing a certain number of read operations from the FIFO output registers until the buffer becomes empty (DIFF_FIFO_[8:0] bits of the FIFO_STATUS1 and FIFO_STATUS2 register are equal to 0).

FIFO can be read when it is empty. In this case, the FIFO word is marked by the specific empty tag.

FIFO output data must be read with multiple of 7 bytes reads starting from the FIFO_DATA_OUT_TAG register. The rounding function from address FIFO_DATA_OUT_Z_H to FIFO_DATA_OUT_TAG is done automatically in the device, in order to allow reading many words with a unique multiple read operation. In this case, it is recommended to retrieve the data from the FIFO as follows:

- 1. Read the FIFO_STATUS1 and FIFO_STATUS2 registers to check how many words are stored in the FIFO. This information is contained in the DIFF_FIFO_[8:0] field.
- 2. Read DIFF_FIFO + N words with a multiple operation (that is, (DIFF_FIFO + N) * 7 bytes), where N is chosen in order to make sure that the FIFO has been emptied.
- 3. If the data read from the FIFO do not contain data marked with the empty tag, then read N additional samples in order to empty the FIFO.

9.9 FIFO watermark threshold

The FIFO threshold is a functionality of the LSM6DSV32X FIFO which can be used to check when the number of samples in the FIFO reaches a defined watermark threshold level.

The bits WTM_[7:0] in the FIFO_CTRL1 register contain the watermark threshold level. The resolution of the WTM $[7:0]$ field is 7 bytes, corresponding to a complete FIFO word. So, the user can select the desired level in a range between 0 and 255.

The bit FIFO_WTM_IA in the FIFO_STATUS2 register represents the watermark status. This bit is set high if the number of words in the FIFO reaches or exceeds the watermark level. FIFO size can be limited to the threshold level by setting the STOP_ON_WTM bit in the FIFO_CTRL2 register to 1.

Figure 33. FIFO threshold (STOP_ON_WTM = 0)

Figure 33. FIFO threshold (STOP ON WTM = 0) shows an example of FIFO threshold level usage when just accelerometer (or gyroscope) data are stored. The STOP_ON_WTM bit set to 0 in the FIFO_CTRL2 register. The threshold level is set to 21 through the WTM_[7:0] bits. The FIFO_WTM_IA bit of the FIFO_STATUS2 register rises after the 21st level has been reached (21 words in the FIFO). Since the STOP_ON_WTM bit is set to 0, the FIFO does not stop at the 21st set of data, but keeps storing data until the FIFO_FULL_IA flag is set high.

Figure 34. FIFO threshold (STOP_ON_WTM = 1) in FIFO mode shows an example of FIFO threshold level usage in FIFO mode with the STOP_ON_WTM bit set to 1 in the FIFO_CTRL2 register. Just accelerometer (or gyroscope) data are stored in this example. The threshold level is set to 21 through the WTM_[7:0] bits and defines the current FIFO size. In FIFO mode, data are stored in the FIFO buffer until the FIFO is full. The FIFO_FULL_IA bit of the FIFO_STATUS2 register rises when the next data stored in the FIFO generates the FIFO full or overrun condition. The FIFO_WTM_IA bit of the FIFO_STATUS2 register goes high when the FIFO is full.

[Figure 35. FIFO threshold \(STOP_ON_WTM = 1\) in continuous mode](#page-113-0) shows an example of FIFO threshold level usage in continuous mode with the STOP_ON_WTM bit set to 1 in the FIFO_CTRL2 register. Just accelerometer (or gyroscope) data are stored in this example. The threshold level is set to 21 through the WTM_[7:0] bits. The FIFO_FULL_IA bit of the FIFO_STATUS2 register rises when the next data stored in the FIFO makes the FIFO full. The FIFO_WTM_IA bit of the FIFO_STATUS2 goes high when the FIFO is full. If data are not retrieved from FIFO, new data (labeled as sample 21) overrides the older data stored in FIFO (labeled as sample F0).

9.10 FIFO compression

FIFO compression is an embedded algorithm that allows storing up to 3 times the number of accelerometer and gyroscope data in FIFO. The compression algorithm automatically analyzes the slope of the sensor waveform and applies the compression of data in FIFO on the basis of the slope (difference between two consecutive samples).

FIFO compression can be enabled on accelerometer and gyroscope data in FIFO by setting both the FIFO_COMPR_EN bit in the EMB_FUNC_EN_B embedded function register and the FIFO_COMPR_RT_EN bit in the FIFO_CTRL2 register. When active, the compression affects both accelerometer and gyroscope data and the level of compression is independent.

The accelerometer and gyroscope batch data rate (BDR) can be configured independently, but the compression algorithm is not supported if the accelerometer and/or the gyroscope are batched at a rate greater than 1920 Hz. FIFO compression supports three different levels of compression:

- NC, not compressed. If the difference between the actual and previous data is higher than 128 LSB, then one sensor sample is stored in one FIFO word.
- 2xC, low compression. If the difference between the actual and previous data between 16 and 128 LSB, then two sensor samples are stored in one FIFO word.
- 3xC, high compression. If the difference between the actual and previous data is less than 16 LSB, then three sensor samples are stored in one FIFO word.

9.10.1 Time correlation

There are five different tags (for each main sensor) depending on the degree of compression:

- NC, noncompressed, associated to the actual time slot
- NC_T_2, noncompressed, associated to two times the previous time slot
- NC_T_1, noncompressed, associated to the previous time slot
- 2xC, low compression
- 3xC, high compression

All NC tags are useful in understanding the time slot correlation. By decoding the sensor tag, it is possible to understand the time frame in which the data was generated.

At the first batch event, the compression algorithm writes a noncompressed word (NC) in FIFO. After that, the algorithm analyzes the slope of the waveforms and three FIFO entries are possible:

- 3xC data written, which contains diff(i), diff(i -1) and diff(i -2)
- 2xC data written, which contains diff($i 1$) and diff($i 2$)
- NC T 2 data written, which contains data($i 2$)

Noncompressed tag sensor NC_T_1 could be written when a configuration change occurs or when the user wants to temporarily disable the runtime FIFO compression by deasserting the FIFO_COMPR_RT_EN bit in the FIFO_CTRL2 register.

The table below summarizes the data and time slot associated for each tag.

Table 96. FIFO compression tags and associated data

As shown in Table 96, using FIFO compression introduces a latency of 2 / BDR, since the compression acts on a window of three BDR.

9.10.2 Data format

A FIFO word of a compressed data contains the information of its slope with respect to its previous data:

 $data(i) = diff(i) + data(i - 1)$

Thus, the last decoded data, data(i-1) in the formula above, must be saved when performing the decompression task.

The following table summarizes the output data format in FIFO for 2xC compressed data.

Table 97. 2xC compressed data output data format in FIFO

The following table summarizes the output data format in FIFO for 3xC compressed data.

Table 98. 3xC compressed data output data format in FIFO

In the table above:

- FIFO_DATA_OUT_X[15:0] = FIFO_DATA_OUT_X_L + FIFO_DATA_OUT_X_H << 8
- FIFO DATA_OUT_Y[15:0] = FIFO_DATA_OUT_Y_L + FIFO_DATA_OUT_Y_H << 8
- FIFO DATA OUT $Z[15:0] =$ FIFO DATA OUT Z L + FIFO DATA OUT Z H << 8

9.10.3 Disabling FIFO compression at runtime

The FIFO compression introduces a latency of 2 / BDR in the writing of the sensor in FIFO. Using FIFO compression is not indicated when user want to flush FIFO with low latency.

In case both high latency and low latency can be used, FIFO can be configured in the more convenient way also at runtime.

The FIFO_COMPR_RT_EN bit can be changed at runtime in order to move from an enabled compression algorithm to a disabled compression algorithm (without latency). The switching is managed as a device configuration change. FIFO writes the CFG-Change sensor at the first BDR event after the change. In that case, all data not yet stored are written at the same time slot with tag NC, NC_T_2 or NC_T_1.

The table below shows an example of a runtime disabled compression algorithm. In this case, a main sensor, CFG-Change sensor and timestamp sensor are supposed to be batched in FIFO. FIFO compression is runtime disabled between time instant t(i-1) and time instant t(i). As explained above, all data that are not yet stored are written to the same slot preceded by CFG-Change and timestamp sensors.

Table 99. Example of disabled runtime compression

9.10.4 CFG-Change sensor with FIFO compression enabled

When a change of configuration is applied to the device, the application processor must discriminate the data of previous configurations with the data of the new configuration. For this task, the same approach as the FIFO_COMPR_RT_EN change is applied as shown in the table below. In this case, a main sensor, CFG-Change sensor and timestamp sensor are supposed to be batched in FIFO. A new device configuration is applied between time instant t(i-1) and time instant t(i). As explained, all data that are not yet stored are written to the same slot preceded by the CFG-Change and timestamp sensors. After that, the FIFO compression algorithm restarts to operate as expected.

Table 100. Example of device configuration change with FIFO compression enabled

9.10.5 Noncompressed data rate

A compression algorithm can be configured in order to guarantee writing of noncompressed data with a certain periodicity (8, 16, 32 BDR events) through the UNCOMPR_RATE_[1:0] field in FIFO_CTRL2.

The usage of the noncompressed data rate in FIFO can be useful for data reconstruction when there is a possibility of FIFO overrun events: if an overrun occurs and the reference noncompressed data is overwritten, it is not possible to reconstruct the current data until new noncompressed data is written in FIFO. The UNCOMPR_RATE_[1:0] field configures the compression algorithm to write noncompressed data at a specific rate, in order to be sure to have at least one noncompressed data every 8, 16, or 32 samples.

Table 101. UNCOPTR_RATE configuration

9.10.6 FIFO compression initialization

When FIFO is set in bypass mode, the compression algorithm must be reinitialized by asserting the FIFO_COMPR_INIT bit in the EMB_FUNC_INIT_B embedded functions register.

9.10.7 FIFO compression example

The following table provides a basic numerical example of the data that could be read from the FIFO when the compression feature is enabled. In this example, the accelerometer sensor only is stored in FIFO and it is configured with a full scale of ±4 *g*.

Table 102. FIFO compression example

At the first batch event, the compression algorithm writes a noncompressed word (NC) without latency in FIFO. After that, the algorithm analyzes the slope of the waveforms and three possible FIFO entries are possible: 3xC, 2xC, NC_T_2. Noncompressed words with the NC_T_1 tag are not present in this example since there is no runtime configuration change.

The second sample stored in FIFO is a noncompressed word with latency of 2 samples (NC_T_2): this FIFO entry contains the entire accelerometer data (without any compression).

Then, since the accelerometer data slope is low, the compression algorithm starts to compress accelerometer data: accelerometer data should be reconstructed starting from the latest sample just before the current one (the first compressed data is expressed as the difference from the NC_T_2 data, the second compressed data is expressed as the difference from the first compressed data, and so on).

As shown in the example, the compression algorithm works with a three-level depth buffer: if a 2xC compression level is written in FIFO, only the previous data (latency 1) and two times the previous data (latency 2) are stored in the FIFO word.

From the example, the benefit of FIFO compression is also shown: the samples are written in FIFO at interlaced ODR, thus limiting intervention by the host processor even more than normal FIFO usage.

9.11 Timestamp correlation

It is possible to reconstruct the timestamp of FIFO stream with three different approaches:

- 1. Basic, using only timestamp sensor information
- 2. Memory-saving, based on the TAG_CNT field in FIFO_DATA_OUT_TAG
- 3. Hybrid, based on combined usage of the TAG_CNT field and decimated timestamp sensor

The basic approach guarantees the highest precision in timestamp reconstruction but wastes a lot of memory space available in FIFO. The timestamp sensor is written in FIFO at each time slot. If the overrun condition occurs, the correct procedure to retrieve the data from FIFO is to discard each data read before a new timestamp sensor.

The memory-saving approach uses only the TAG_CNT information and, when the TAG_CNT value increases, the timestamp stored at the software layer should be updated as follows:

timestamp = timestamp $(i-1)$ + $\frac{1}{BDR_MAX}$

The memory-saving approach allows the user to maximize the data stored in FIFO. With this method all the timestamp correlation is forwarded to the application processor.

This approach is not recommended when the overrun condition can occur.

The hybrid approach is a trade-off and a combination of the two previous solutions. The timestamp is configured to be written in FIFO with decimation. When the TAG_CNT value increases, the timestamp stored at the software layer should be updated as in the memory-saving approach, while when the timestamp sensor is read, the timestamp stored at the software layer should be realigned with the correct value from the sensor.

10 Temperature sensor

The device is provided with an internal temperature sensor that is suitable for ambient temperature measurement. If both the accelerometer and the gyroscope sensors are in power-down mode, the temperature sensor is off.

The maximum output data rate of the temperature sensor is 60 Hz and its value depends on how the accelerometer and gyroscope sensors are configured:

- If the gyroscope is in power-down mode:
	- If the accelerometer is configured in low-power mode and its ODR is lower than 60 Hz, the temperature data rate is equal to the configured accelerometer ODR.
	- The temperature data rate is equal to 60 Hz for all other accelerometer configurations.
- If the gyroscope is not in power-down mode, the temperature data rate is equal to 60 Hz, regardless of the accelerometer and gyroscope configuration.

For the temperature sensor, the data-ready signal is represented by the TDA bit of the STATUS_REG register. The signal can be driven to the INT2 pin by setting the INT2_DRDY_TEMP bit of the CTRL4 register to 1.

The temperature data is given by the concatenation of the OUT_TEMP_H and OUT_TEMP_L registers and it is represented as a number of 16 bits in two's complement format with a sensitivity of 256 LSB/°C. The output zero level corresponds to 25°C.

Temperature sensor data can also be stored in FIFO with a configurable batch data rate (see [Section 9: First-in,](#page-90-0) [first out \(FIFO\) buffer](#page-90-0) for details).

10.1 Example of temperature data calculation

The following table provides a few basic examples of the data that is read from the temperature data registers at different ambient temperature values. The values listed in this table are given under the hypothesis of perfect device calibration (that is, no offset, no gain error, and so forth).

Table 103. Content of output data registers vs. temperature

11 Analog hub and Qvar functionality

The LSM6DSV32X embeds an analog hub (AH) chain which provides the capability to connect an external analog input and convert it to a digital signal for processing. The same analog chain can be used to implement the Qvar (electric charge variation detection) functionality by connecting external electrodes and an electronic signal conditioning circuit. The external electrodes have to be connected to pin 2 (SDx/AH1/Qvar1) and/or pin 3 (SCx/AH2/Qvar2), so the I²C-master interface (mode 2) and the auxiliary SPI (mode 3) are not available when Qvar is used. Figure 36 provides an example of a test circuit.

Note: The Qvar functionality is available in mode 1 connection mode for the I²C interface only.

Figure 36. Qvar external connections

(1) ST ESDALCL5-1BM2 is referenced as an ST catalog product but similar features of other ESD diodes also can be used.

In the LSM6DSV32X, the analog hub / Qvar has a dedicated channel that can be activated by setting the AH_QVAR_EN bit to 1 in the CTRL7 register.

Note: When the analog hub / Qvar channel is enabled, the accelerometer sensor must be set in high-performance mode or in normal mode.

> The analog hub / Qvar data-ready signal is represented by the AH_QVARDA bit of the STATUS_REG register. This signal can be driven to the INT2 pin by setting the INT2_DRDY_AH_QVAR bit to 1 in the CTRL7 register. Analog hub / Qvar data are available as a 16-bit word in two's complement in the AH_QVAR_OUT_L and AH_QVAR_OUT_H registers at a fixed rate of 240 Hz (typical). They can also be processed by MLC/FSM logic. The equivalent input impedance of the analog hub / Qvar buffers can be selected by properly setting the AH_QVAR_C_ZIN_[1:0] bits in the CTRL7 register (00: 2.4 GΩ (default); 01: 730 MΩ; 10: 300 MΩ; 11: 235 MΩ). Additional details about Qvar are available in the application note AN5755 on [www.st.com.](http://www.st.com)

12 Self-test

The embedded self-test functions allow checking the device functionality without moving it.

12.1 Accelerometer self-test (UI) – mode 1, 2

When the accelerometer self-test is enabled, an actuation force is applied to the sensor, simulating a definite input acceleration. In this case, the sensor outputs exhibit a change in their DC levels which are related to the selected full scale through the sensitivity value.

When the device is configured in mode 1 or mode 2, the accelerometer self-test function can be configured from the primary interface only. It is off when the ST_XL_[1:0] bits of the CTRL10 register are programmed to 00. It is enabled when the ST_XL_[1:0] bits are set to 01 (positive sign self-test) or 10 (negative sign self-test).

When the accelerometer self-test is activated, the sensor output level is given by the algebraic sum of the signals produced by the acceleration acting on the sensor and by the electrostatic test-force.

The complete accelerometer self-test procedure in mode 1 or mode 2 is indicated in [Figure 37. Accelerometer](#page-124-0) [self-test procedure \(UI\)](#page-124-0).

Note: All the read/write operations in this procedure must be performed **over the pr i m a ry I 2 C / SPI / I 3 C i n t e rfac e**

Write 05h to CTRL1 (10h) Write 00h to CTRL2 (11h) Write 44h to CTRL3 (12h) Write 00h to CTRL4 (13h) Write 00h to CTRL5 (14h) Write 00h to CTRL6 (15h) Write 00h to CTRL7 (16h) Write 04h to CTRL8 (17h) Write 00h to CTRL9 (18h) Write 00h to CTRL10 (19h)

 \rightarrow Initialize and turn on the sensor → Set BDU = 1, ODR = 60 Hz, FS = ±4 *g*

Power up, wait 100 ms for stable output

Check XLDA in STATUS_REG (1Eh) – accelerometer data -ready bit → Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, wait for the first sample **Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh)**

Discard data

For 5 times, after checking the XLDA bit, read the output registers Read OUTX L_A (28h), OUTX H_A (29h): Store data in OUTX_NOST **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_NOST **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_NOST *The 16 -bit data is expressed in two's complement.*

Average the stored data on each axis

Write 01h to CTRL10 (19h) Enable accelerometer self-test **Wait 100 ms for stable output**

Accelerometer UI self- test mode 1 and mode 2

Check XLDA in STATUS_REG (1Eh) – accelerometer data -ready bit → Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, wait for the first sample **Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh) → Discard data**

For 5 times, after checking the XLDA bit, read the output registers Read OUTX_L_A (28h), OUTX_H_A (29h): Store data in OUTX_ST **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_ST **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_ST *The 16 -bit data is expressed in two's complement.*

Average the stored data on each axis

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12.2 Accelerometer self-test (UI) with OIS chain on - mode 3

If the auxiliary SPI is used and both the UI chain and the OIS chain are on, the accelerometer self-test function has to be enabled from the primary interface through the ST_XL_[1:0] bits of the CTRL10 register. It cannot be enabled from both interfaces at the same time (forbidden condition).

The recommended accelerometer self-test procedure on the UI chain with the OIS chain on is indicated in [Figure 38. Accelerometer self-test procedure \(UI\) with OIS chain on.](#page-126-0)

- **All the black colored read/write operations must be performed**
- **over the primary I²C / SPI interface** • **All the blue colored read/write operations must be performed**
- **over the auxiliary SPI interface**

Accelerometer UI self-test mode 3 with both UI and OIS chains ON

Write 05h to CTRL1 (10h) Write 00h to CTRL2 (11h) Write 44h to CTRL3 (12h) Write 00h to CTRL4 (13h) Write 00h to CTRL5 (14h) **Write 00h to CTRL6 (15h) Write 00h to CTRL7 (16h) Write 04h to CTRL8 (17h) Write 00h to CTRL9 (18h) Write 00h to CTRL10 (19h)**

 \rightarrow Initialize and turn on the sensor (UI chain) \rightarrow **Set BDU =1, ODR = 60 Hz, FS = ±4** *g*

Write 00h to SPI2_INT_OIS (6Fh) Write 05h to SPI2_CTRL1_OIS (70h) (write 25h if using SPI 3 -wire) Write 00h to SPI2_CTRL2_OIS (71h) Write 00h to SPI2_CTRL3_OIS (72h)

 \rightarrow Turn on the sensor (OIS chain)

Power up, wait 100 ms for stable output

Check XLDA in STATUS_REG (1Eh) – accelerometer data -ready bit A Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, wait for the first sample

Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh)

Discard data

For 5 times, after checking the XLDA bit, read the output registers Read OUTX L_A (28h), OUTX_H_A (29h): Store data in OUTX_NOST **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_NOST **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_NOST *The 16 -bit data is expressed in two's complement.*

Average the stored data on each axis

Write 01h to CTRL10 (19h) Enable accelerometer self-test **Wait 100 ms for stable output**

Check XLDA in STATUS_REG (1Eh) – accelerometer data -ready bit → Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, Wait for the first sample **Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh) → Discard data**

For 5 times, after checking the XLDA bit, read the output registers Read OUTX L_A (28h), OUTX_H_A (29h): Store data in OUTX_ST **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_ST **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_ST **Average the stored data on each axis**

Read OUTX L_A (28h), OUTX_H_A (29h): Store data in OUTX_ST_OIS_ **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_ST_OIS **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_ST_OIS

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12.3 Accelerometer self-test (OIS) – mode 3

If the auxiliary SPI is used and the UI chain is off, the accelerometer self-test function on the OIS chain can be enabled through the auxiliary SPI interface by setting the ST_XL_OIS_[1:0] bits of the SPI2_INT_OIS register. The self-test function is off when the ST_XL_OIS_[1:0] bits are programmed to 00. It is enabled when the ST_XL_OIS_[1:0] bits are set to 01 (positive sign self-test) or 10 (negative sign self-test).

The complete accelerometer self-test procedure on the OIS chain with the UI chain off is indicated in [Figure 39. Accelerometer self-test procedure \(OIS\)](#page-128-0). This procedure can be performed only if the UI chain is off $(ODR_XL_{13:0}=0000$ in the CTRL1 register and $ODR_{16:13:0}=0000$ in the CTRL2 register).

Accelerometer OIS self-test mode 3 with OIS chain ON and UI chain OFF

Notes:

- All the read/write operations in this procedure must be performed **over the auxiliary SPI interface**
- This procedure can be performed only if the **UI chain is off** (ODR XL [3:0] = 0000 in CTRL1)

Write 00h to SPI2_INT_OIS (6Fh)

Write 05h to SPI2_CTRL1_OIS (70h) (write 25h if using SPI 3-wire) Write 00h to SPI2_CTRL2_OIS (71h) Write 00h to SPI2_CTRL3_OIS (72h)

 \rightarrow Initialize and turn on the sensor \rightarrow FS = \pm 4 *g* (ODR at 7680 Hz)

Power up, wait 100 ms for stable output

Check XLDA in SPI2_STATUS_REG_OIS (1Eh) – acc data-ready bit → Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, wait for the first sample **Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh)**

→ Discard data

For 5 times, after checking the XLDA bit, read the output registers Read OUTX_L_A (28h), OUTX_H_A (29h): Store data in OUTX_NOST_OIS **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_NOST_OIS **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_NOST_OIS *The 16-bit data is expressed in two's complement.*

Average the stored data on each axis

Write 01h to INT_OIS (6Fh) → Enable accelerometer self-test Wait 100 ms

Check XLDA in SPI2_STATUS_REG_OIS (1Eh) – acc data-ready bit A Reading OUTX_A / OUTY_A / OUTZ_A clears XLDA, wait for the first sample **Read OUTX_A (28h / 29h), OUTY_A (2Ah / 2Bh), OUTZ_A (2Ch / 2Dh) → Discard data**

For 5 times, after checking the XLDA bit, read the output registers Read OUTX_L_A (28h), OUTX_H_A (29h): Store data in OUTX_ST_OIS **Read OUTY_L_A (2Ah), OUTY_H_A (2Bh):** Store data in OUTY_ST_OIS **Read OUTZ_L_A (2Ch), OUTZ_H_A (2Dh):** Store data in OUTZ_ST_OIS *The 16-bit data is expressed in two's complement.*

Average the stored data on each axis

12.4 Gyroscope self-test (UI) – mode 1, 2

The gyroscope self-test allows testing the mechanical and electrical parts of the gyroscope sensor. When it is activated, an equivalent Coriolis signal is emulated at the input of the ASIC front-end and the sensor output exhibits an output change.

When the device is configured in mode 1 or mode 2, the gyroscope self-test function can be configured from the primary interface only. It is off when the ST_G_[1:0] bits of the CTRL10 register are programmed to 00. It is enabled when the ST_G_[1:0] bits are set to 01 (positive sign self-test) or 10 (negative sign self-test).

When the gyroscope self-test is active, the sensor output level is given by the algebraic sum of the signals produced by the angular rate acting on the sensor and by the electrostatic test-force.

The complete gyroscope self-test procedure in mode 1 or mode 2 is indicated in [Figure 40. Gyroscope self-test](#page-130-0) [procedure \(UI\)](#page-130-0).

Note: All the read/write operations in this procedure must be performed **over the primary I 2C / SPI / I3C interface**

Write 00h to CTRL1 (10h) Write 07h to CTRL2 (11h) Write 44h to CTRL3 (12h) Write 00h to CTRL4 (13h) Write 00h to CTRL5 (14h) Write 04h to CTRL6 (15h) Write 00h to CTRL7 (16h) Write 04h to CTRL8 (17h) Write 00h to CTRL9 (18h) Write 00h to CTRL10 (19h)

 \rightarrow Initialize and turn on the sensor \rightarrow Set BDU = 1, ODR = 240 Hz, FS = \pm 2000 dps

Power up, wait 100 ms for stable output Check GDA in STATUS_REG (1Eh) – Gyroscope data-ready bit A Reading OUTX_G / OUTY_G / OUTZ_G clears GDA, wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) → Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX L_G (22h), OUTX H_G (23h): Store data in OUTX_NOST **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_NOST **Read OUTZ L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_NOST *The 16-bit data is expressed in two's complement.* **Average the stored data on each axis**

Gyroscope UI self-test mode 1 and mode 2

Write 04h to CTRL10 (19h) → Enable gyroscope self-test Wait 100 ms

Check GDA in STATUS_REG (1Eh) – Gyroscope data-ready bit → Reading OUTX / OUTY / OUTZ clears GDA, wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) → Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX_L_G (22h), OUTX_H_G (23h): Store data in OUTX_ST **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_ST **Read OUTZ_L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_ST *The 16-bit data is expressed in two's complement.* **Average the stored data on each axis**

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12.5 Gyroscope self-test (UI) with OIS chain on - mode 3

If the auxiliary SPI is used and both the UI chain and the OIS chain are on, the gyroscope self-test function has to be enabled from the primary interface through the ST_G_[1:0] bits of the CTRL10 register. It cannot be enabled from both interfaces at the same time (forbidden condition).

The recommended gyroscope self-test procedure on the UI chain with the OIS chain on is indicated in [Figure 41. Gyroscope self-test procedure \(UI\) with OIS chain on](#page-132-0).

Figure 41. Gyroscope self-test procedure (UI) with OIS chain on

Notes:

- **All the black colored read/write operations must be performed**
- **over the primary I²C / SPI interface**
- **All the blue colored read/write operations must be performed**
- **over the auxiliary SPI interface**

Gyroscope UI Self-test mode 3 with both UI and OIS chains ON

Write 00h to CTRL1 (10h) Write 07h to CTRL2 (11h) Write 44h to CTRL3 (12h) Write 04h to CTRL4 (13h) Write 00h to CTRL5 (14h)

Write 04h to CTRL6 (15h) Write 00h to CTRL7 (16h) Write 04h to CTRL8 (17h) Write 00h to CTRL9 (18h) Write 00h to CTRL10 (19h)

 \rightarrow Initialize and turn on the sensor (UI chain) \rightarrow Set BDU = 1, ODR = 240 Hz, FS = \pm 2000 dps

Write 00h to SPI2_INT_OIS (6Fh) Write 03h to SPI2_CTRL1_OIS (70h) (write 23h if using SPI 3 -wire) Write 00h to SPI2_CTRL2_OIS (71h) Write 00h to SPI2_CTRL3_OIS (72h)

 \rightarrow Turn on the sensor (OIS chain)

Power up, wait 100 ms for stable output

Check GDA in STATUS_REG (1Eh) – Gyroscope data -ready A Reading OUTX_G / OUTY_G / OUTZ_G clears GDA, wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX L_G (22h), OUTX_H_G (23h): Store data in OUTX_NOST **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_NOST **Read OUTZ L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_NOST *The 16 -bit data is expressed in two's complement.*

Average the stored data on each axis

Write 04h to CTRL10 (19h) Enable gyroscope self-test **Wait 100 ms**

Check GDA in STATUS_REG (1Eh) – Gyroscope data-ready bit → Reading OUTX/OUTY/OUTZ clears GDA, Wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX L G (22h), OUTX H G (23h): Store data in OUTX ST Read OUTY_L_G (24h), OUTY_H_G (25h): Store data in OUTY_ST **Read OUTZ_L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_ST **Average the stored data on each axis**

Read OUTX L_G (22h), OUTX_H_G (23h): Store data in OUTX_ST_OIS **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_ST_OIS **Read OUTZ_L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_ST_OIS

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12.6 Gyroscope self-test (OIS) – mode 3

If the auxiliary SPI is used and the UI chain is off, the gyroscope self-test function on the OIS chain can be enabled through the auxiliary SPI interface by setting the ST_G_OIS_[1:0] bits of the SPI2_INT_OIS register. The self-test function is off when the ST_G_OIS_1:0] bits are programmed to 00. It is enabled when the ST_G_OIS_[1:0] bits are set to 01 (positive sign self-test) or 11 (negative sign self-test).

The complete gyroscope self-test procedure on the OIS chain with the UI chain off is indicated in [Figure 42. Gyroscope self-test procedure \(OIS\).](#page-134-0) This procedure can be performed only if the UI chain is off $(ODR_XL_{13:0}] = 0000$ in the CTRL1 register and ODR_{$_G13:0$} = 0000 in the CTRL2 register).

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Gyroscope OIS self-test mode 3 with OIS chain ON and UI chain OFF

Notes:

- All the read/write operations in this procedure must be performed **over the auxiliary SPI interface**
- This procedure can be performed only if the **UI chain is off** $(ODR \ G \ [3:0] = 0000$ in CTRL2)

Write 00h to SPI2_INT_OIS (6Fh)

Write 03h to SPI2_CTRL1_OIS (70h) (write 23h if using SPI 3-wire) Write 04h to SPI2_CTRL2_OIS (71h) Write 00h to SPI2_CTRL3_OIS (72h)

 \rightarrow Initialize and turn on the sensor \rightarrow FS = \pm 2000 dps (ODR at 7680 Hz)

Power up, wait 100 ms for stable output

Check GDA in SPI2_STATUS_REG_OIS (1Eh) – gyro data-ready bit → Reading OUTX_G / OUTY_G / OUTZ_G clears GDA, wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX L_G (22h), OUTX_H_G (23h): Store data in OUTX_NOST_OIS **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_NOST_OIS **Read OUTZ L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_NOST_OIS *The 16-bit data is expressed in two's complement.*

Average the stored data on each axis

Write 04h to SPI2_INT_OIS (6Fh) → Enable gyroscope self-test Wait 100 ms

Check GDA in SPI2_STATUS_REG_OIS (1Eh) – gyro data-ready bit → Reading OUTX / OUTY / OUTZ clears GDA, wait for the first sample **Read OUTX_G (22h / 23h), OUTY_G (24h / 25h), OUTZ_G (26h / 27h) → Discard data**

For 5 times, after checking the GDA bit, read the output registers Read OUTX L_G (22h), OUTX_H_G (23h): Store data in OUTX_ST_OIS **Read OUTY_L_G (24h), OUTY_H_G (25h):** Store data in OUTY_ST_OIS **Read OUTZ_L_G (26h), OUTZ_H_G (27h):** Store data in OUTZ_ST_OIS *The 16-bit data is expressed in two's complement.*

Average the stored data on each axis

Revision history

Table 104. Document revision history

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